

SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

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1

TLC271, TLC271A, TLC271B

LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

DEVICE FEATURES

PARAMETER†	BIAS-SELECT MODE			UNIT
	HIGH	MEDIUM	LOW	
P_D	3375	525	50	μW
SR	3.6	0.4	0.03	V/ μs
V_n	25	32	68	nV/ $\sqrt{Hz}$
B_1	1.7	0.5	0.09	MHz
A_{VD}	23	170	480	V/mV

† Typical at $V_{DD} = 5 V$, $T_A = 25^\circ C$

description (continued)

Using the bias-select option, these cost-effective devices can be programmed to span a wide range of applications that previously required BiFET, NFET, or bipolar technology. Three offset voltage grades are available (C-suffix and I-suffix types), ranging from the low-cost TLC271 (10 mV) to the TLC271B (2 mV) low-offset version. The extremely high input impedance and low bias currents, in conjunction with good common-mode rejection and supply voltage rejection, make these devices a good choice for new state-of-the-art designs as well as for upgrading existing designs.

In general, many features associated with bipolar technology are available in LinCMOS™ operational amplifiers, without the power penalties of bipolar technology. General applications such as transducer interfacing, analog calculations, amplifier blocks, active filters, and signal buffering are all easily designed with the TLC271. The devices also exhibit low-voltage single-supply operation, making them ideally suited for remote and inaccessible battery-powered applications. The common-mode input voltage range includes the negative rail.

A wide range of packaging options is available, including small-outline and chip-carrier versions for high-density system applications.

The device inputs and output are designed to withstand –100-mA surge currents without sustaining latch-up.

The TLC271 incorporates internal ESD-protection circuits that prevent functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2; however, care should be exercised in handling these devices as exposure to ESD may result in the degradation of the device parametric performance.

The C-suffix devices are characterized for operation from $0^\circ C$ to $70^\circ C$. The I-suffix devices are characterized for operation from $-40^\circ C$ to $85^\circ C$. The M-suffix devices are characterized for operation over the full military temperature range of $-55^\circ C$ to $125^\circ C$.

bias-select feature

The TLC271 offers a bias-select feature that allows the user to select any one of three bias levels depending on the level of performance desired. The tradeoffs between bias levels involve ac performance and power dissipation (see Table 1).

Table 1. Effect of Bias Selection on Performance

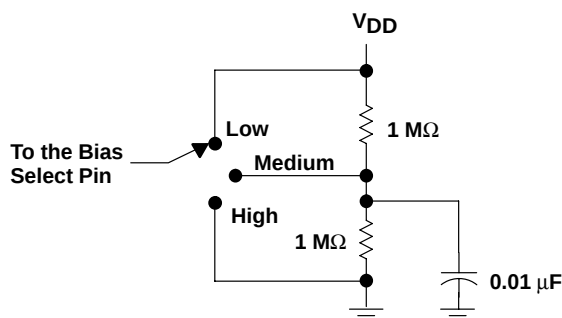
TYPICAL PARAMETER VALUES $T_A = 25^\circ C$, $V_{DD} = 5 V$		MODE			UNIT
		HIGH BIAS $R_L = 10 k\Omega$	MEDIUM BIAS $R_L = 100 k\Omega$	LOW BIAS $R_L = 1 M\Omega$	
P_D	Power dissipation	3.4	0.5	0.05	mW
SR	Slew rate	3.6	0.4	0.03	V/ μs
V_n	Equivalent input noise voltage at $f = 1 kHz$	25	32	68	nV/ $\sqrt{Hz}$
B_1	Unity-gain bandwidth	1.7	0.5	0.09	MHz
ϕ_m	Phase margin	46°	40°	34°	
A_{VD}	Large-signal differential voltage amplification	23	170	480	V/mV



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bias selection

Bias selection is achieved by connecting the bias select pin to one of three voltage levels (see Figure 1). For medium-bias applications, it is recommended that the bias select pin be connected to the midpoint between the supply rails. This procedure is simple in split-supply applications, since this point is ground. In single-supply applications, the medium-bias mode necessitates using a voltage divider as indicated in Figure 1. The use of large-value resistors in the voltage divider reduces the current drain of the divider from the supply line. However, large-value resistors used in conjunction with a large-value capacitor require significant time to charge up to the supply midpoint after the supply is switched on. A voltage other than the midpoint can be used if it is within the voltages specified in Figure 1.



BIAS MODE	BIAS-SELECT VOLTAGE (single supply)
Low	V_{DD}
Medium	1 V to $V_{DD} - 1$ V
High	GND

Figure 1. Bias Selection for Single-Supply Applications

high-bias mode

In the high-bias mode, the TLC271 series features low offset voltage drift, high input impedance, and low noise. Speed in this mode approaches that of BiFET devices but at only a fraction of the power dissipation. Unity-gain bandwidth is typically greater than 1 MHz.

medium-bias mode

The TLC271 in the medium-bias mode features low offset voltage drift, high input impedance, and low noise. Speed in this mode is similar to general-purpose bipolar devices but power dissipation is only a fraction of that consumed by bipolar devices.

SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

In the low-bias mode, the TLC271 features low offset voltage drift, high input impedance, extremely low power consumption, and high differential voltage gain.

TOPIC	BIAS MODE
schematic	all
absolute maximum ratings	all
recommended operating conditions	all
electrical characteristics operating characteristics typical characteristics	high (Figures 2 – 33)
electrical characteristics operating characteristics typical characteristics	medium (Figures 34 – 65)
electrical characteristics operating characteristics typical characteristics	low (Figures 66 – 97)
parameter measurement information	all
application information	all

The schematic diagram illustrates a 12-bit Successive Approximation Register (SAR) ADC. The circuit is powered by a 1.5V reference voltage (V_{DD}) and a ground (GND) rail. The input signals are labeled IN- and IN+.

Input Stage: The input signals IN- and IN+ are connected to a differential pair of NMOS transistors (N1, N2) and PMOS transistors (P1, P2). The gates of N1 and N2 are connected to the IN- and IN+ inputs, respectively. The sources of N1 and N2 are connected to ground. The drains of N1 and N2 are connected to the gates of P1 and P2, respectively. The sources of P1 and P2 are connected to V_{DD}. The output of the input stage is connected to the gates of N3 and N4.

Offset Compensation: The circuit includes two offset compensation stages, labeled OFFSET N1 and OFFSET N2. These stages use NMOS transistors (N1, N2) and PMOS transistors (P1, P2) to compensate for offset errors. The output of the offset compensation stage is connected to the gates of N3 and N4.

Reference Voltage: A 1.5V reference voltage (V_{DD}) is applied to the gates of the PMOS transistors (P1, P2, P3, P4, P5, P6A, P6B, P7A, P7B, P8, P9A, P9B, P10, P11, P12) and the drains of the NMOS transistors (N1, N2, N3, N4, N5, N6, N7, N8, N9, N10, N11, N12, N13).

Successive Approximation Register (SAR): The SAR is implemented using a series of PMOS transistors (P3, P4, P5, P6A, P6B, P7A, P7B, P8, P9A, P9B, P10, P11, P12) and NMOS transistors (N3, N4, N5, N6, N7, N8, N9, N10, N11, N12, N13). The gates of the PMOS transistors are connected to the output of the input stage and the output of the offset compensation stage. The gates of the NMOS transistors are connected to the output of the input stage and the output of the offset compensation stage. The sources of the NMOS transistors are connected to ground.

Output: The output of the SAR is connected to the gates of the PMOS transistors (P3, P4, P5, P6A, P6B, P7A, P7B, P8, P9A, P9B, P10, P11, P12) and the drains of the NMOS transistors (N3, N4, N5, N6, N7, N8, N9, N10, N11, N12, N13). The output signal is labeled OUT.

Labels: The circuit includes labels for various components: IN-, IN+, V_{DD}, GND, OUT, BIAS SELECT, and OFFSET N1, OFFSET N2.

absolute maximum ratings over operating free-air temperature (unless otherwise noted)[†]

Supply voltage, V_{DD} (see Note 1)	18 V
Differential input voltage, V_{ID} (see Note 2)	$\pm V_{DD}$
Input voltage range, V_I (any input)	–0.3 V to V_{DD}
Input current, I_I	± 5 mA
Output current, I_O	± 30 mA
Duration of short-circuit current at (or below) 25°C (see Note 3)	Unlimited
Continuous total dissipation	See Dissipation Rating Table
Operating free-air temperature, T_A : C suffix	0°C to 70°C
I suffix	–40°C to 85°C
M suffix	–55°C to 125°C
Storage temperature range	–65°C to 150°C
Case temperature for 60 seconds: FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds: D or P package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: JG package	300°C

[†] Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES:
1. All voltage values, except differential voltages, are with respect to network ground.
 2. Differential voltages are at $IN+$ with respect to $IN-$.
 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded (see application section).

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	145 mW
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
JG	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
P	1000 mW	8.0 mW/°C	640 mW	520 mW	200 mW

recommended operating conditions

		C SUFFIX		I SUFFIX		M SUFFIX		UNIT
		MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, V_{DD}		3	16	4	16	5	16	V
Common-mode input voltage, V_{IC}	$V_{DD} = 5$ V	–0.2	3.5	–0.2	3.5	0	3.5	V
	$V_{DD} = 10$ V	–0.2	8.5	–0.2	8.5	0	8.5	
Operating free-air temperature, T_A		0	70	–40	85	–55	125	°C

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SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

HIGH-BIAS MODE

electrical characteristics at specified free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T _A [†]	TLC271C, TLC271AC, TLC271BC						UNIT
				V _{DD} = 5 V			V _{DD} = 10 V			
				MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO}	Input offset voltage	TLC271C TLC271AC TLC271BC	V _O = 1.4 V, V _{IC} = 0 V, R _S = 50 Ω, R _L = 10 kΩ	25°C	1.1	10	1.1	10	mV	
				Full range	12	12				
				25°C	0.9	5	0.9	5		
				Full range	6.5	6.5				
				25°C	0.34	2	0.39	2		
				Full range	3	3				
α _{VIO}	Average temperature coefficient of input offset voltage		25°C to 70°C	1.8	2	μV/°C				
I _{IO}	Input offset current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.1	60	0.1	60	pA		
			70°C	7	300	7	300			
I _{IB}	Input bias current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.6	60	0.7	60	pA		
			70°C	40	600	50	600			
V _{ICR}	Common-mode input voltage range (see Note 5)		25°C	−0.2 to 4	−0.3 to 4.2	−0.2 to 9	−0.3 to 9.2	V		
			Full range	−0.2 to 3.5	−0.2 to 8.5	V				
V _{OH}	High-level output voltage	V _{ID} = 100 mV, R _L = 10 kΩ	25°C	3.2	3.8	8	8.5	V		
			0°C	3	3.8	7.8	8.5			
			70°C	3	3.8	7.8	8.4			
V _{OL}	Low-level output voltage	V _{ID} = −100 mV, I _{OL} = 0	25°C	0	50	0	50	mV		
			0°C	0	50	0	50			
			70°C	0	50	0	50			
A _{VD}	Large-signal differential voltage amplification	R _L = 10 kΩ, See Note 6	25°C	5	23	10	36	V/mV		
			0°C	4	27	7.5	42			
			70°C	4	20	7.5	32			
CMRR	Common-mode rejection ratio	V _{IC} = V _{ICRmin}	25°C	65	80	65	85	dB		
			0°C	60	84	60	88			
			70°C	60	85	60	88			
k _{SVR}	Supply-voltage rejection ratio (ΔV _{DD} /ΔV _{IO})	V _{DD} = 5 V to 10 V V _O = 1.4 V	25°C	65	95	65	95	dB		
			0°C	60	94	60	94			
			70°C	60	96	60	96			
I _{I(SEL)}	Input current (BIAS SELECT)	V _{I(SEL)} = 0	25°C	−1.4	−1.9	μA				
I _{DD}	Supply current	V _O = V _{DD} /2, V _{IC} = V _{DD} /2, No load	25°C	675	1600	950	2000	μA		
			0°C	775	1800	1125	2200			
			70°C	575	1300	750	1700			

[†] Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

6. At V_{DD} = 5 V, V_O = 0.25 V to 2 V; at V_{DD} = 10 V, V_O = 1 V to 6 V.



HIGH-BIAS MODE

electrical characteristics at specified free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T _A [†]	TLC271I, TLC271AI, TLC271BI						UNIT
				V _{DD} = 5 V			V _{DD} = 10 V			
				MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO}	Input offset voltage	V _O = 1.4 V, V _{IC} = 0 V, R _S = 50 Ω, R _L = 10 kΩ	25°C	1.1		10	1.1		10	mV
			Full range	13			13			
			25°C	0.9		5	0.9		5	
			Full range	7			7			
			25°C	0.34		2	0.39		2	
			Full range	3.5			3.5			
α _{VIO}	Average temperature coefficient of input offset voltage		25°C to 85°C	1.8			2		μV/°C	
I _{IO}	Input offset current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.1		60	0.1		60	pA
			85°C	24		1000	26		1000	
I _{IB}	Input bias current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.6		60	0.7		60	pA
			85°C	200		2000	220		2000	
V _{ICR}	Common-mode input voltage range (see Note 5)		25°C	–0.2 to 4	–0.3 to 4.2		–0.2 to 9	–0.3 to 9.2	V	
			Full range	–0.2 to 3.5		–0.2 to 8.5		V		
V _{OH}	High-level output voltage	V _{ID} = 100 mV, R _L = 10 kΩ	25°C	3.2	3.8		8	8.5	V	
			–40°C	3	3.8		7.8	8.5		
			85°C	3	3.8		7.8	8.5		
V _{OL}	Low-level output voltage	V _{ID} = –100 mV, I _{OL} = 0	25°C	0		50	0		50	mV
			–40°C	0		50	0		50	
			85°C	0		50	0		50	
A _{VD}	Large-signal differential voltage amplification	R _L = 10 kΩ, See Note 6	25°C	5	23		10	36	V/mV	
			–40°C	3.5	32		7	46		
			85°C	3.5	19		7	31		
CMRR	Common-mode rejection ratio	V _{IC} = V _{ICRmin}	25°C	65	80		65	85	dB	
			–40°C	60	81		60	87		
			85°C	60	86		60	88		
k _{SVR}	Supply-voltage rejection ratio (ΔV _{DD} /ΔV _{IO})	V _{DD} = 5 V to 10 V V _O = 1.4 V	25°C	65	95		65	95	dB	
			–40°C	60	92		60	92		
			85°C	60	96		60	96		
I _{I(SEL)}	Input current (BIAS SELECT)	V _{I(SEL)} = 0	25°C	–1.4			–1.9		μA	
I _{DD}	Supply current	V _O = V _{DD} /2, V _{IC} = V _{DD} /2, No load	25°C	675	1600		950	2000	μA	
			–40°C	950	2200		1375	2500		
			85°C	525	1200		725	1600		

[†] Full range is –40°C to 85°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

6. At V_{DD} = 5 V, V_O = 0.25 V to 2 V; at V_{DD} = 10 V, V_O = 1 V to 6 V.

TLC271, TLC271A, TLC271B

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OPERATIONAL AMPLIFIERS

SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

HIGH-BIAS MODE

electrical characteristics at specified free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T _A [†]	TLC271M						UNIT
				V _{DD} = 5 V			V _{DD} = 10 V			
				MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO}	Input offset voltage	V _O = 1.4 V, V _{IC} = 0 V, R _S = 50 Ω, R _L = 10 kΩ	25°C	1.1	10		1.1	10		mV
			Full range		12			12		
α _{VIO}	Average temperature coefficient of input offset voltage		25°C to 125°C	2.1			2.2			μV/°C
I _{IO}	Input offset current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.1	60		0.1	60		pA
			125°C	1.4	15		1.8	15		nA
I _{IB}	Input bias current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.6	60		0.7	60		pA
			125°C	9	35		10	35		nA
V _{ICR}	Common-mode input voltage range (see Note 5)		25°C	0 to 4	–0.3 to 4.2		0 to 9	–0.3 to 9.2		V
			Full range	0 to 3.5		0 to 8.5			V	
V _{OH}	High-level output voltage	V _{ID} = 100 mV, R _L = 10 kΩ	25°C	3.2	3.8		8	8.5		V
			–55°C	3	3.8		7.8	8.5		
			125°C	3	3.8		7.8	8.4		
V _{OL}	Low-level output voltage	V _{ID} = –100 mV, I _{OL} = 0	25°C		0	50		0	50	mV
			–55°C		0	50		0	50	
			125°C		0	50		0	50	
A _{VD}	Large-signal differential voltage amplification	R _L = 10 kΩ, See Note 6	25°C	5	23		10	36		V/mV
			–55°C	3.5	35		7	50		
			125°C	3.5	16		7	27		
CMRR	Common-mode rejection ratio	V _{IC} = V _{ICRmin}	25°C	65	80		65	85		dB
			–55°C	60	81		60	87		
			125°C	60	84		60	86		
k _{SVR}	Supply-voltage rejection ratio (ΔV _{DD} /ΔV _{IO})	V _{DD} = 5 V to 10 V V _O = 1.4 V	25°C	65	95		65	95		dB
			–55°C	60	90		60	90		
			125°C	60	97		60	97		
I _{I(SEL)}	Input current (BIAS SELECT)	V _{I(SEL)} = 0	25°C	–1.4			–1.9		μA	
I _{DD}	Supply current	V _O = V _{DD} /2, V _{IC} = V _{DD} /2, No load	25°C	675	1600		950	2000		μA
			–55°C	1000	2500		1475	3000		
			125°C	475	1100		625	1400		

[†] Full range is –55°C to 125°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

6. At V_{DD} = 5 V, V_O = 0.25 V to 2 V; at V_{DD} = 10 V, V_O = 1 V to 6 V.



HIGH-BIAS MODE

operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS		T_A	TLC271C, TLC271AC, TLC271BC			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C		3.6		V/ μ s
			0°C		4		
			70°C		3		
		$V_{I(PP)} = 2.5\text{ V}$	25°C		2.9		
			0°C		3.1		
			70°C		2.5		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 20\text{ }\Omega$,	25°C		25		nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 10\text{ k}\Omega$,	$C_L = 20\text{ pF}$, See Figure 98	25°C		320		kHz
			0°C		340		
			70°C		260		
B_1 Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$,	25°C		1.7		MHz
			0°C		2		
			70°C		1.3		
ϕ_m Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 100	25°C		46°		
			0°C		47°		
			70°C		44°		

operating characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS		T_A	TLC271C, TLC271AC, TLC271BC			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C		5.3		V/ μ s
			0°C		5.9		
			70°C		4.3		
		$V_{I(PP)} = 5.5\text{ V}$	25°C		4.6		
			0°C		5.1		
			70°C		3.8		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 20\text{ }\Omega$,	25°C		25		nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 10\text{ k}\Omega$,	$C_L = 20\text{ pF}$, See Figure 98	25°C		200		kHz
			0°C		220		
			70°C		140		
B_1 Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$,	25°C		2.2		MHz
			0°C		2.5		
			70°C		1.8		
ϕ_m Phase margin	$f = B_1$, $C_L = 20\text{ pF}$,	$V_I = 10\text{ mV}$, See Figure 100	25°C		49°		
			0°C		50°		
			70°C		46°		

TLC271, TLC271A, TLC271B
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OPERATIONAL AMPLIFIERS

SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

HIGH-BIAS MODE

operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS	T_A	TLC271I, TLC271AI, TLC271BI			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C	3.6		V/ μ s
			–40°C	4.5		
			85°C	2.8		
		$V_{I(PP)} = 2.5\text{ V}$	25°C	2.9		
			–40°C	3.5		
			85°C	2.3		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 20\text{ }\Omega$, 25°C		25		nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 10\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	25°C		320		kHz
		–40°C		380		
		85°C		250		
B_1 Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$, 25°C		1.7		MHz
		–40°C		2.6		
		85°C		1.2		
ϕ_m Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$, $f = B_1$, See Figure 100	25°C		46°		
		–40°C		49°		
		85°C		43°		

operating characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS	T_A	TLC271I, TLC271AI, TLC271BI			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C	5.3		V/ μ s
			–40°C	6.8		
			85°C	4		
		$V_{I(PP)} = 5.5\text{ V}$	25°C	4.6		
			–40°C	5.8		
			85°C	3.5		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 20\text{ }\Omega$, 25°C		25		nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 10\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	25°C		200		kHz
		–40°C		260		
		85°C		130		
B_1 Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$, 25°C		2.2		MHz
		–40°C		3.1		
		85°C		1.7		
ϕ_m Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$, $f = B_1$, See Figure 100	25°C		49°		
		–40°C		52°		
		85°C		46°		



HIGH-BIAS MODE

operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS	T_A	TLC271M			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C	3.6		V/ μ s
			–55°C	4.7		
			125°C	2.3		
		$V_{I(PP)} = 2.5\text{ V}$	25°C	2.9		
			–55°C	3.7		
			125°C	2		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 20\text{ }\Omega$, 25°C		25		nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 10\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	25°C		320		kHz
		–55°C		400		
		125°C		230		
B_1 Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$, 25°C		1.7		MHz
		–55°C		2.9		
		125°C		1.1		
ϕ_m Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$, $f = B_1$, See Figure 100	25°C		46°		
		–55°C		49°		
		125°C		41°		

operating characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS	T_A	TLC271M			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 10\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C	5.3		V/ μ s
			–55°C	7.1		
			125°C	3.1		
		$V_{I(PP)} = 5.5\text{ V}$	25°C	4.6		
			–55°C	6.1		
			125°C	2.7		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 20\text{ }\Omega$, 25°C		25		nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 10\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	25°C		200		kHz
		–55°C		280		
		125°C		110		
B_1 Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$, 25°C		2.2		MHz
		–55°C		3.4		
		125°C		1.6		
ϕ_m Phase margin	$f = B_1$, $C_L = 20\text{ pF}$, $V_I = 10\text{ mV}$, See Figure 100	25°C		49°		
		–55°C		52°		
		125°C		44°		

TLC271, TLC271A, TLC271B
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)

Table of Graphs

			FIGURE
V_{IO}	Input offset voltage	Distribution	2, 3
α_{VIO}	Temperature coefficient	Distribution	4, 5
V_{OH}	High-level output voltage	vs High-level output current vs Supply voltage vs Free-air temperature	6, 7 8 9
V_{OL}	Low-level output voltage	vs Common-mode input voltage vs Differential input voltage vs Free-air temperature vs Low-level output current	10, 11 12 13 14, 15
A_{VD}	Large-signal differential voltage amplification	vs Supply voltage vs Free-air temperature vs Frequency	16 17 28, 29
I_{IB}	Input bias current	vs Free-air temperature	18
I_{IO}	Input offset current	vs Free-air temperature	18
V_{IC}	Common-mode input voltage	vs Supply voltage	19
I_{DD}	Supply current	vs Supply voltage vs Free-air temperature	20 21
SR	Slew rate	vs Supply voltage vs Free-air temperature	22 23
	Bias-select current	vs Supply voltage	24
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency	25
B_1	Unity-gain bandwidth	vs Free-air temperature vs Supply voltage	26 27
A_{VD}	Large-signal differential voltage amplification	vs Frequency	28, 29
ϕ_m	Phase margin	vs Supply voltage vs Free-air temperature vs Capacitive load	30 31 32
V_n	Equivalent input noise voltage	vs Frequency	33
	Phase shift	vs Frequency	28, 29



TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)[†]

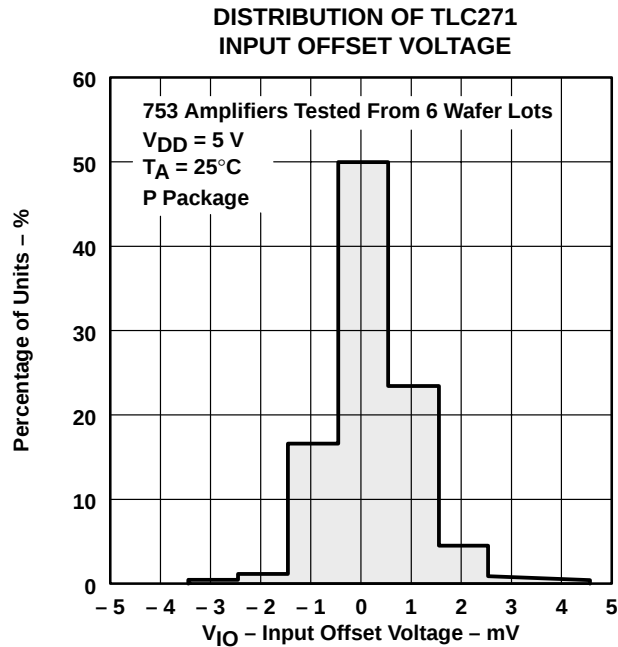


Figure 2

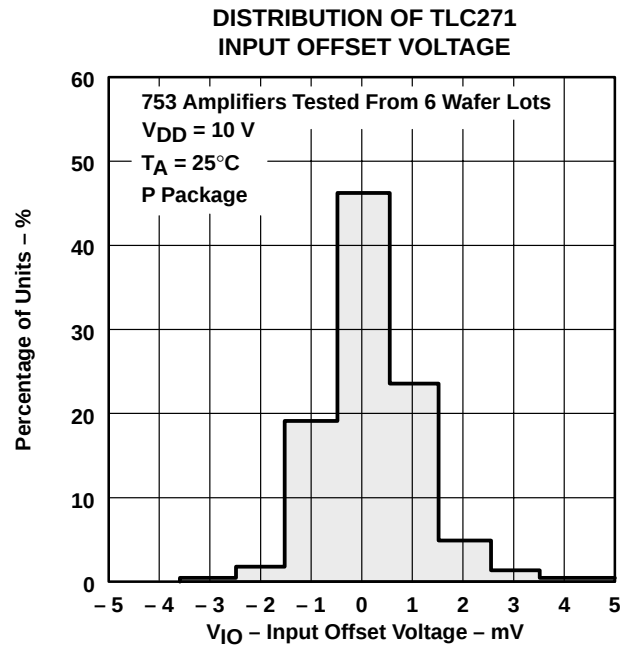


Figure 3

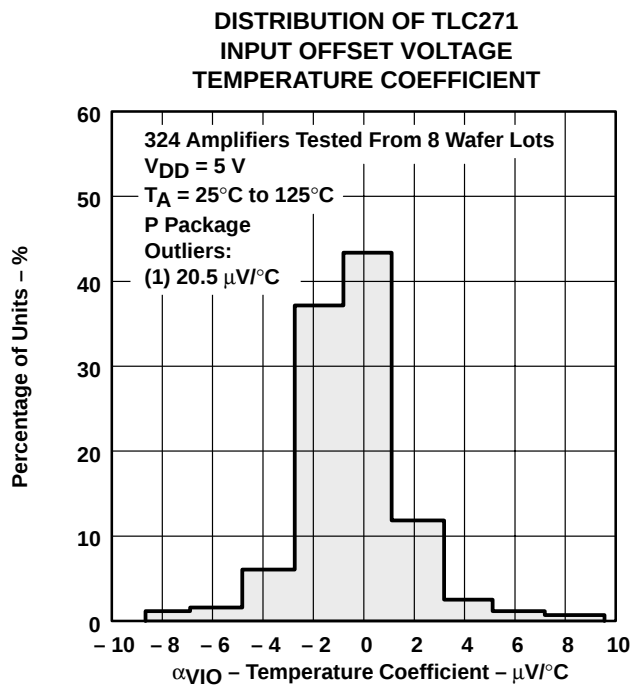


Figure 4

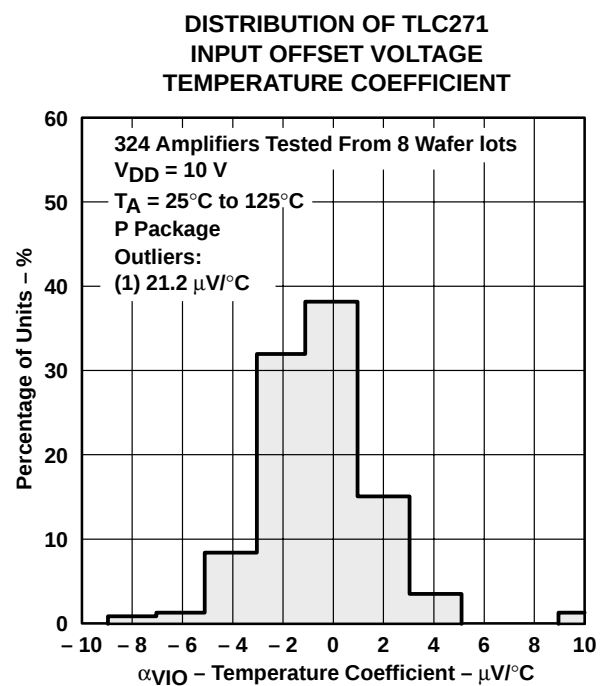


Figure 5

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC271, TLC271A, TLC271B

LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)[†]

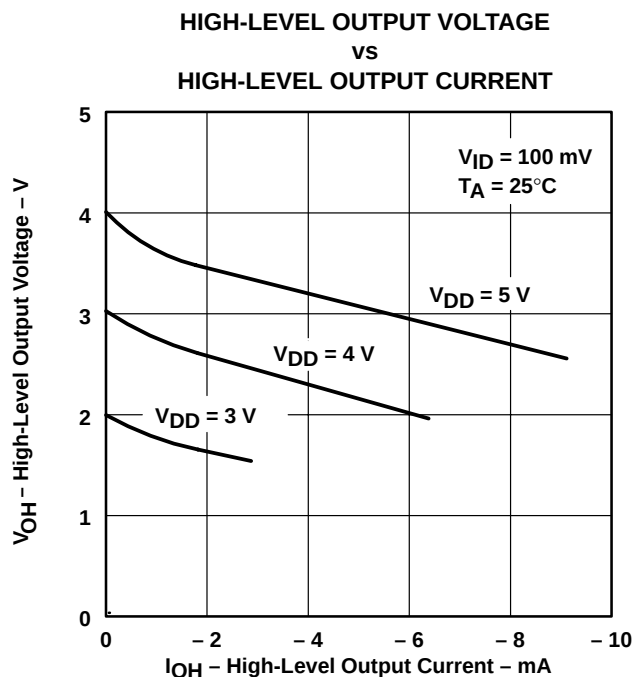


Figure 6

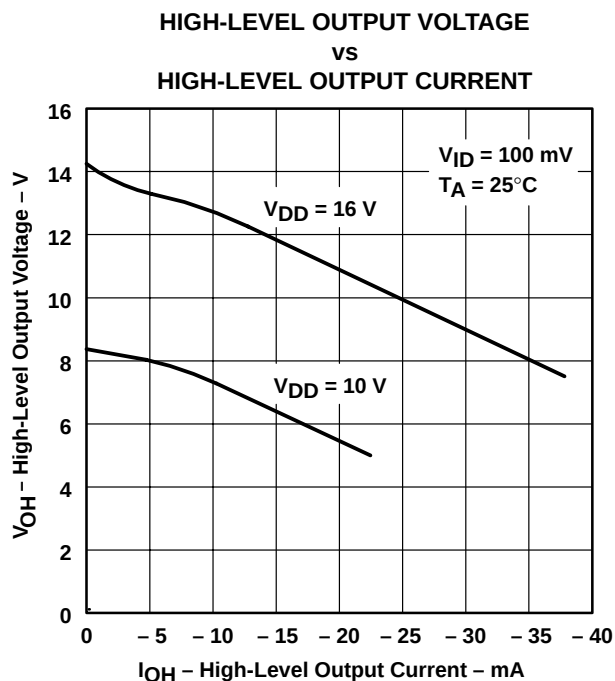


Figure 7

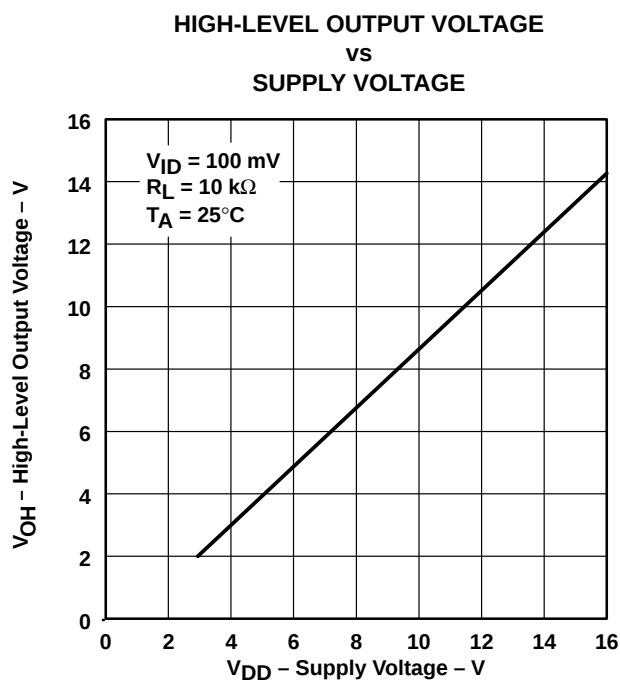


Figure 8

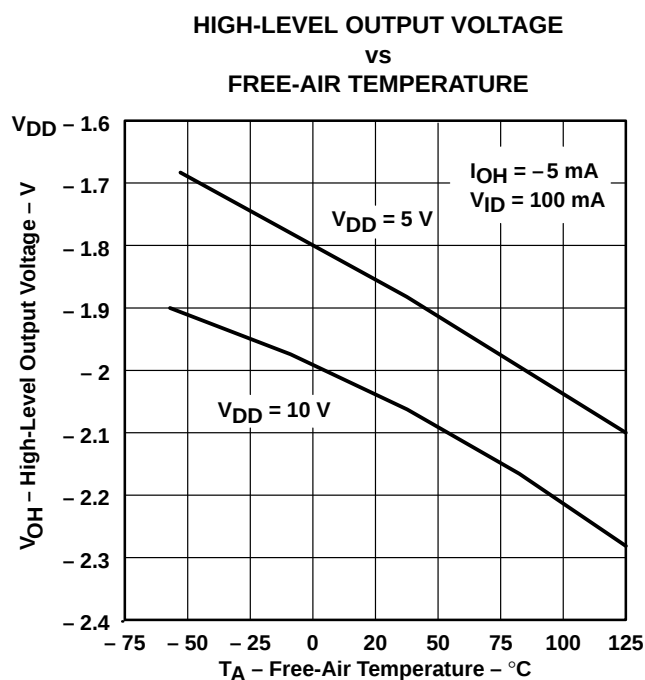


Figure 9

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)[†]

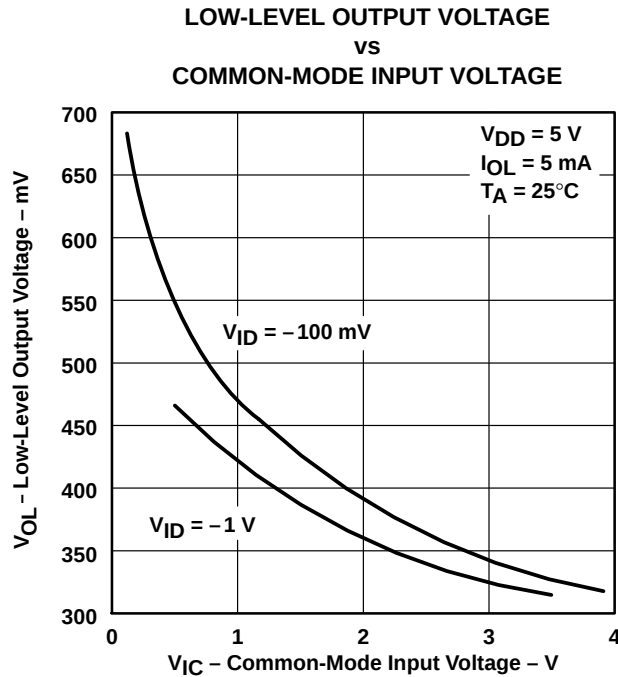


Figure 10

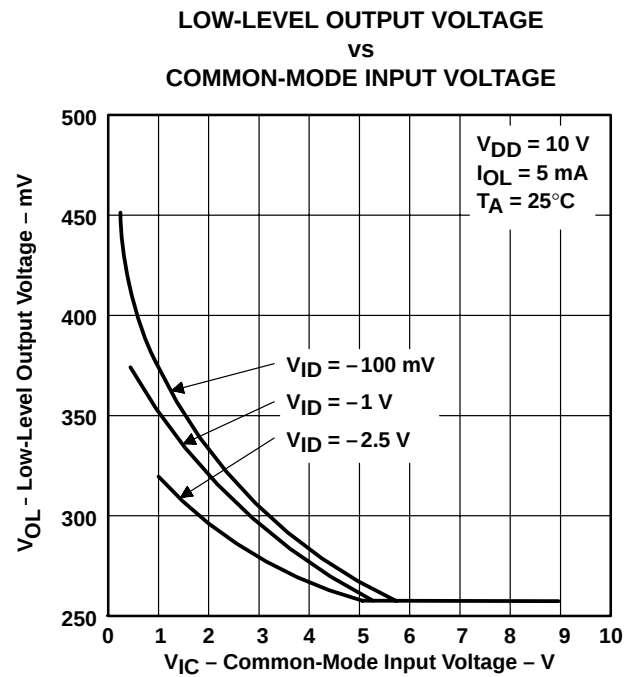


Figure 11

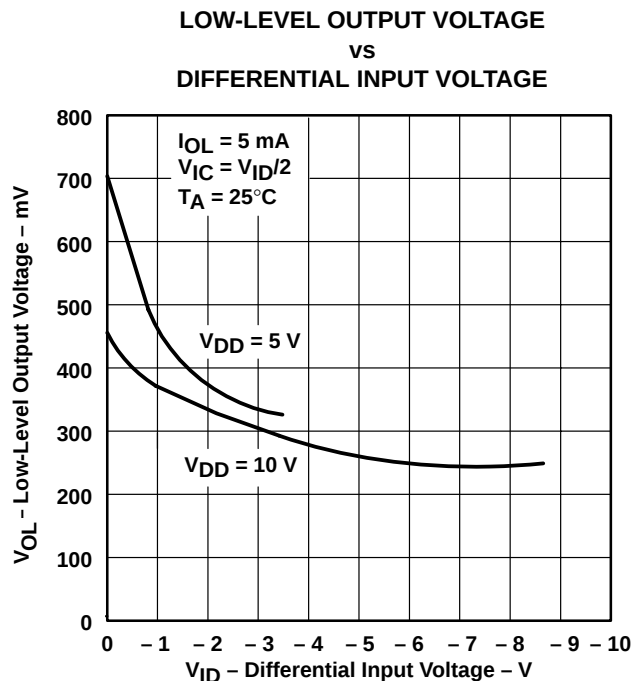


Figure 12

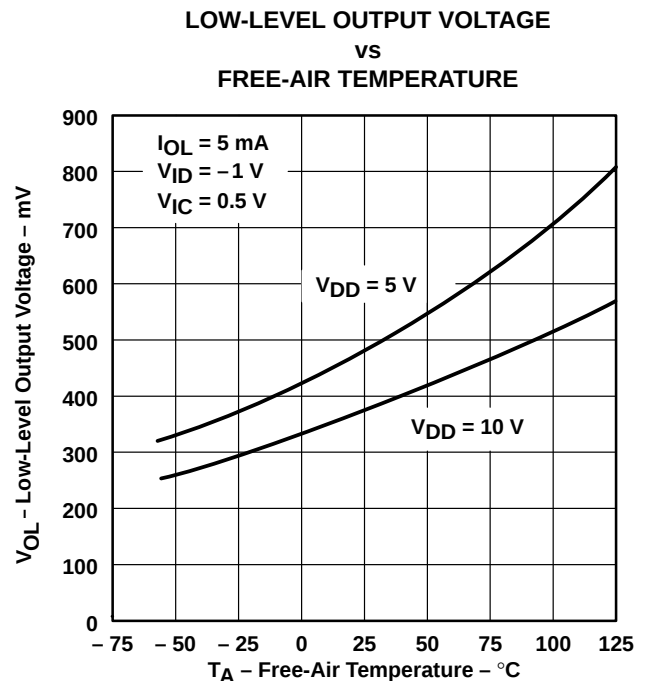


Figure 13

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC271, TLC271A, TLC271B

LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)[†]

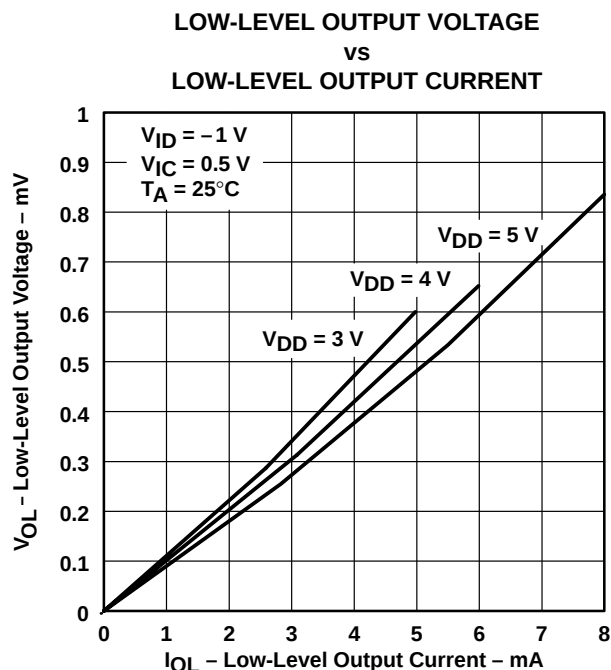


Figure 14

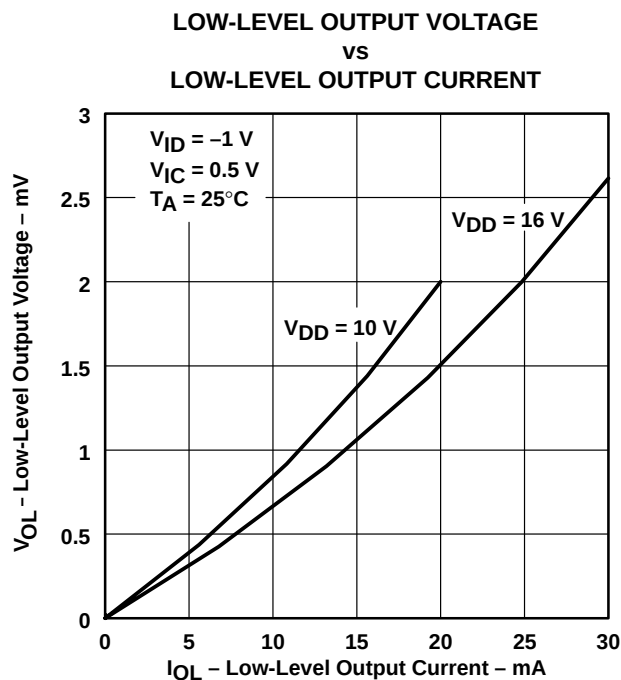


Figure 15

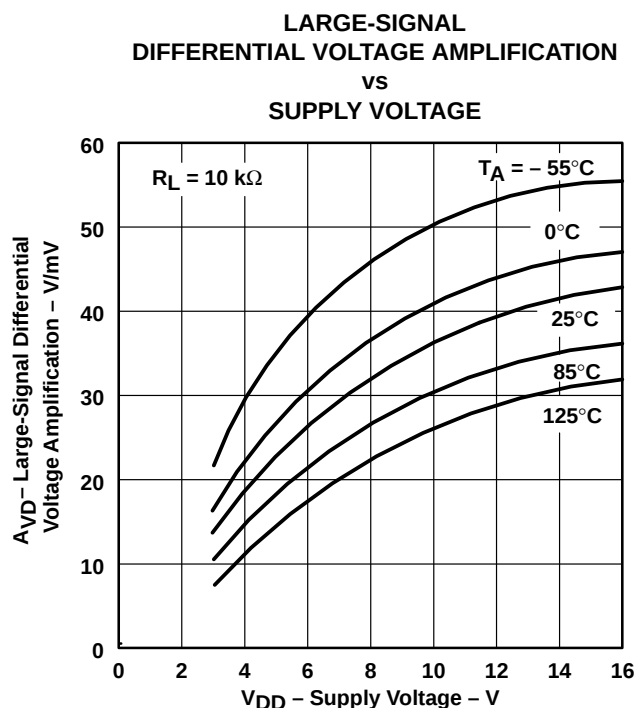


Figure 16

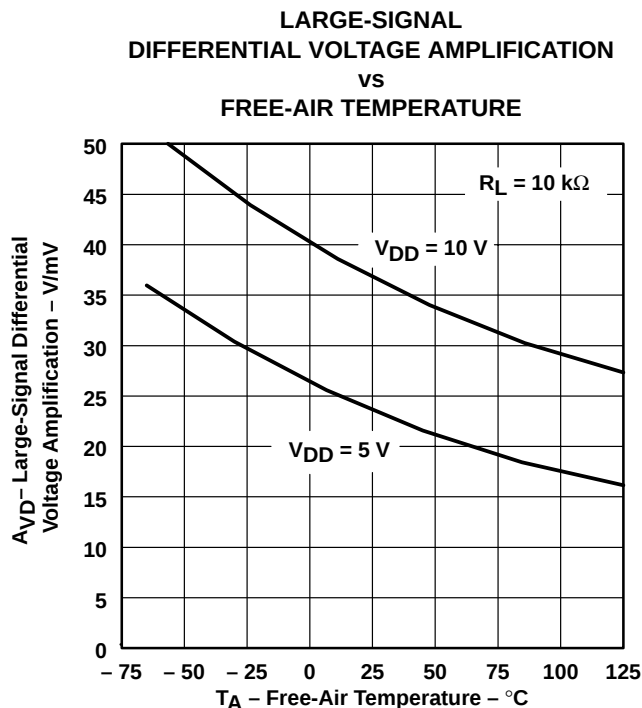
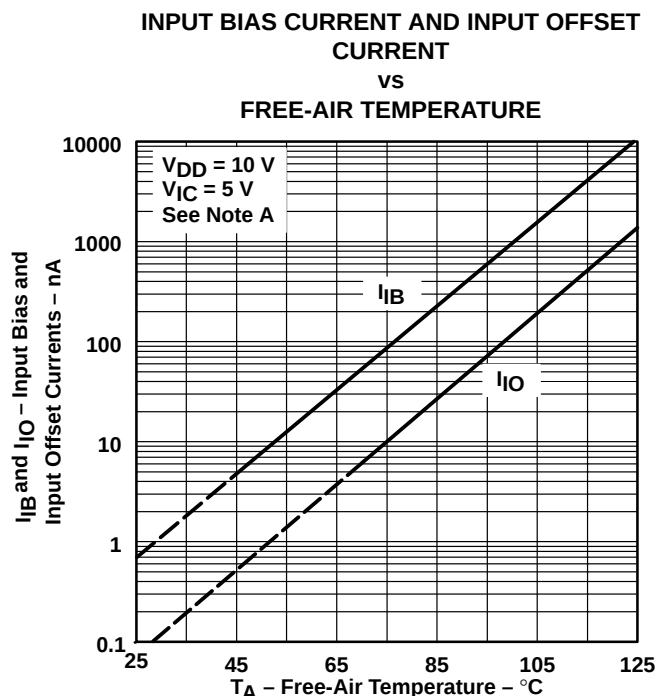


Figure 17

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)[†]



NOTE A: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

Figure 18

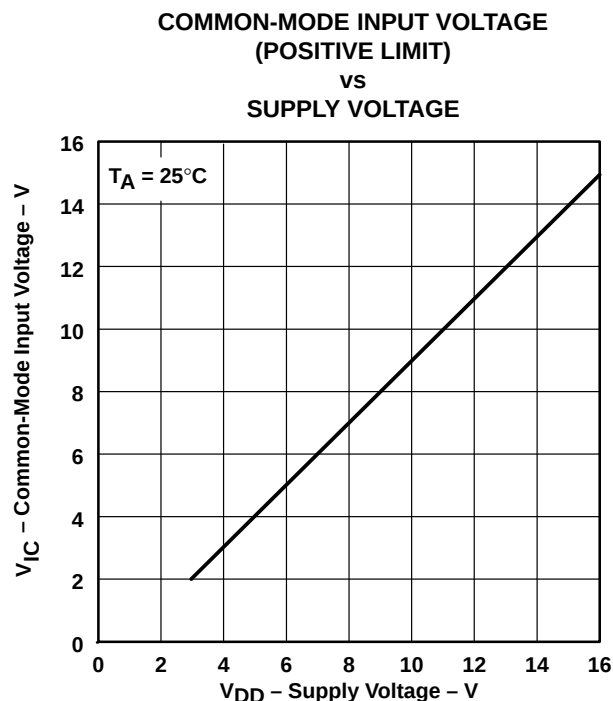


Figure 19

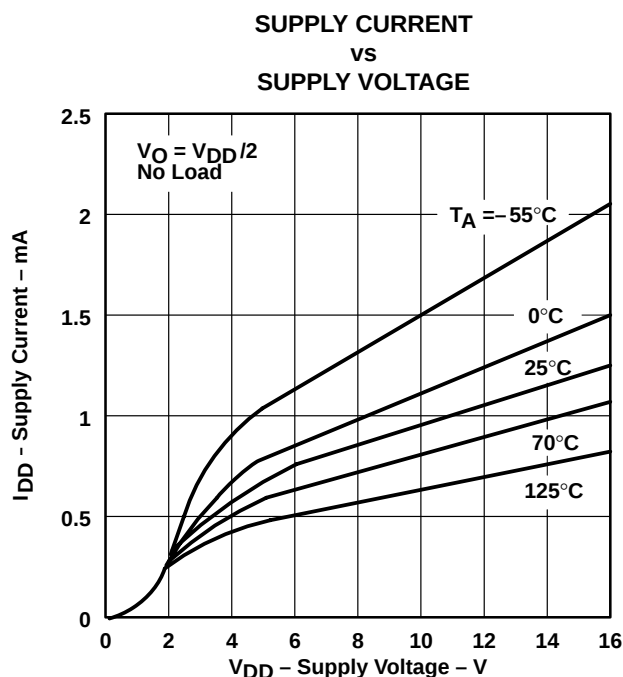


Figure 20

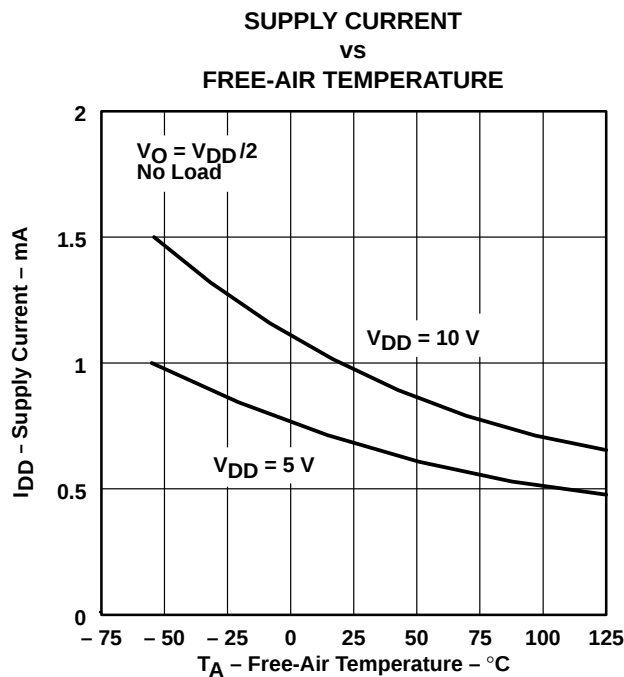


Figure 21

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC271, TLC271A, TLC271B

LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)[†]

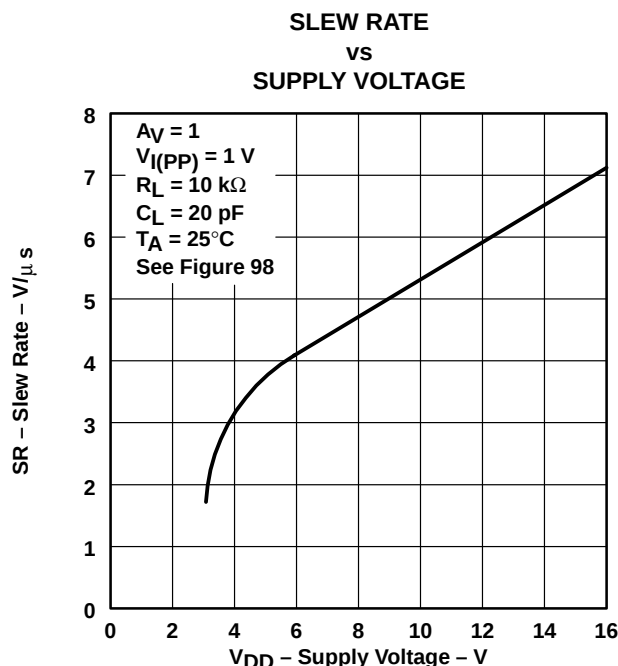


Figure 22

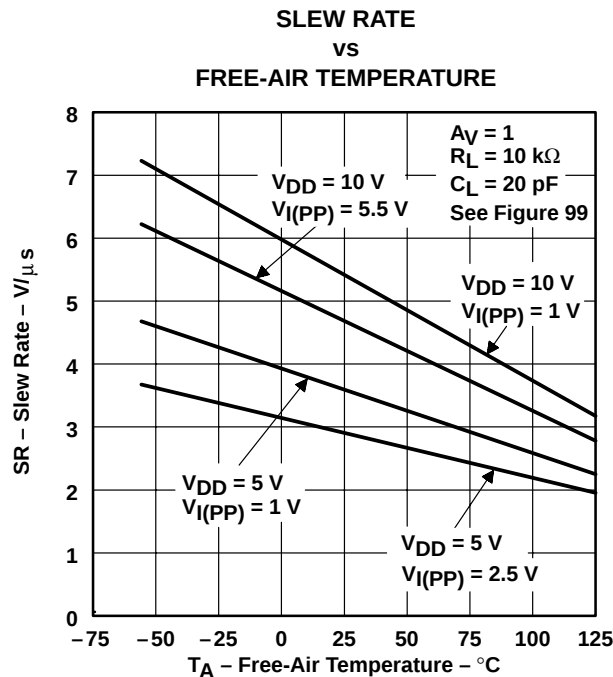


Figure 23

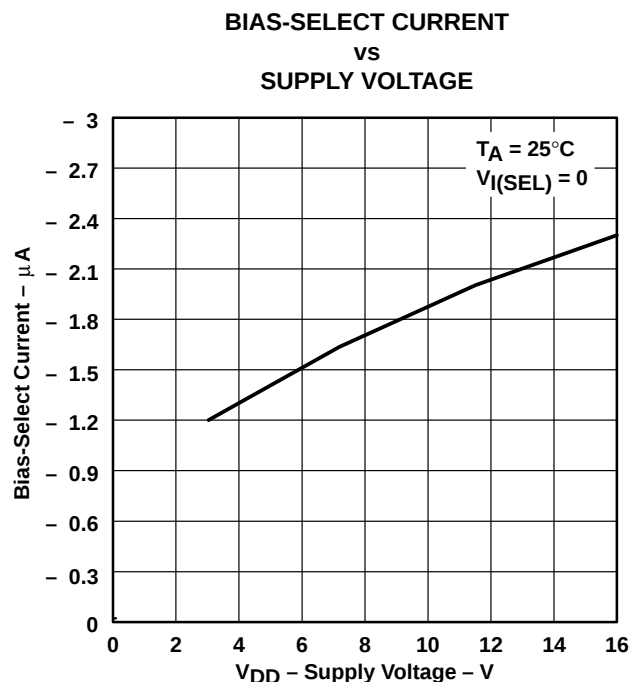


Figure 24

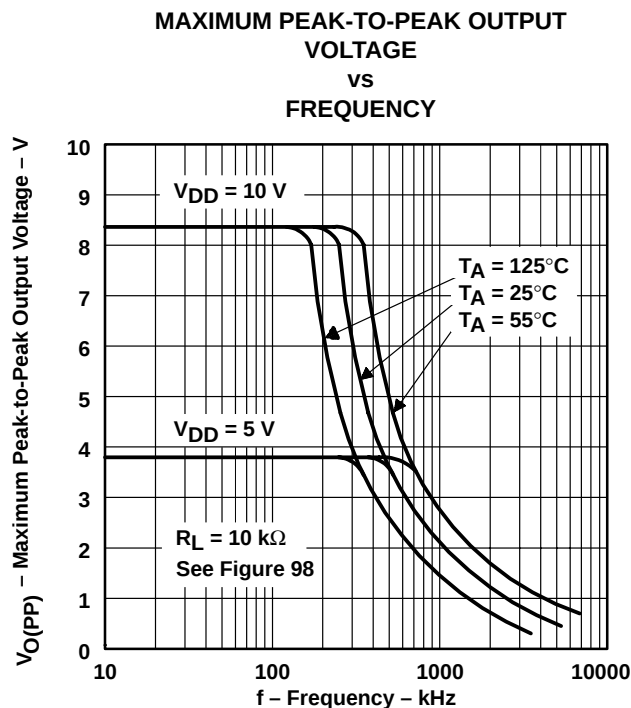


Figure 25

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)[†]

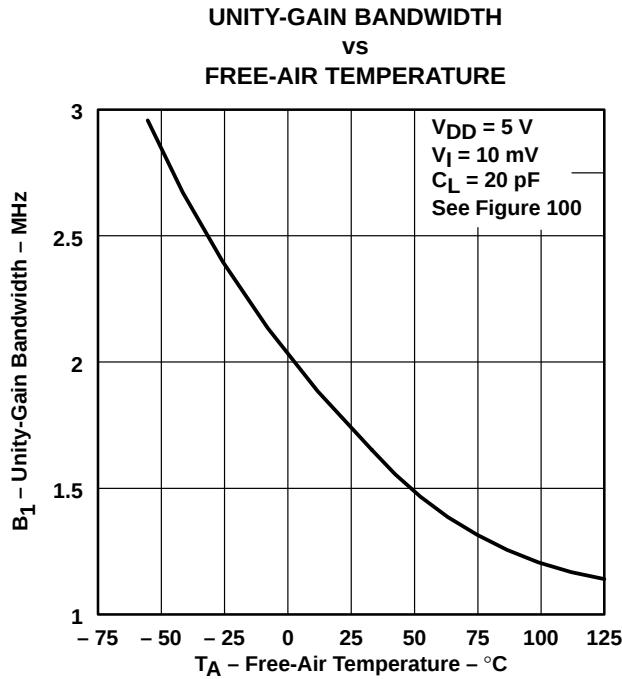


Figure 26

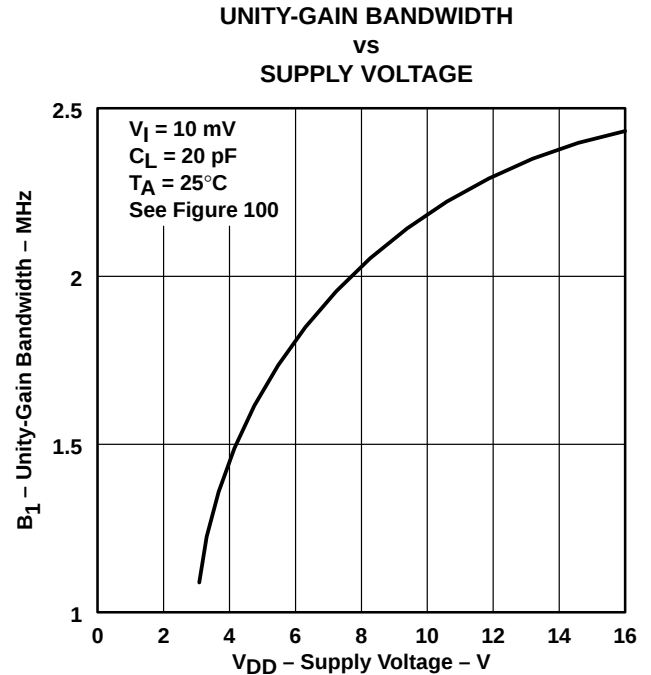


Figure 27

**LARGE-SIGNAL DIFFERENTIAL VOLTAGE
AMPLIFICATION AND PHASE SHIFT
VS
FREQUENCY**

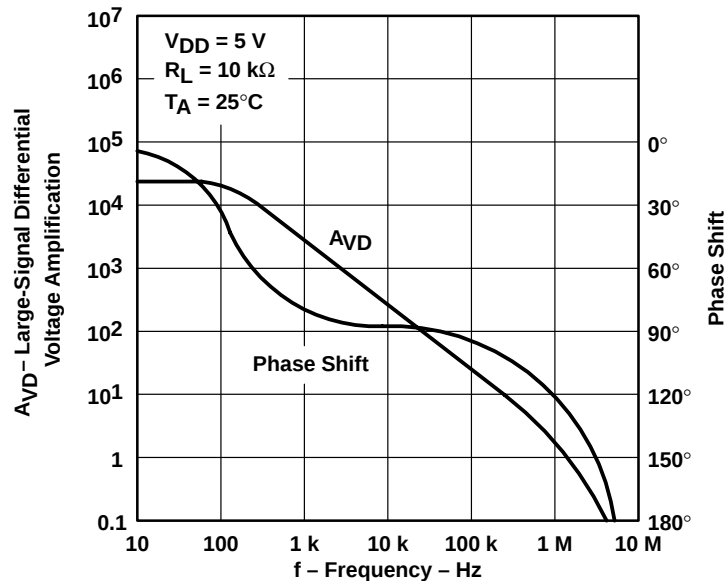


Figure 28

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC271, TLC271A, TLC271B

LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)[†]

LARGE-SCALE DIFFERENTIAL VOLTAGE AMPLIFICATION AND PHASE SHIFT

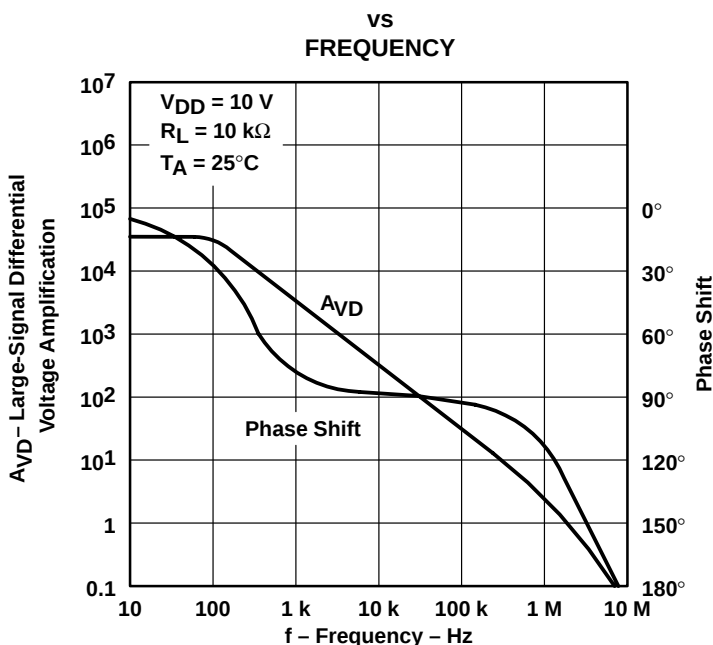


Figure 29

PHASE MARGIN vs SUPPLY VOLTAGE

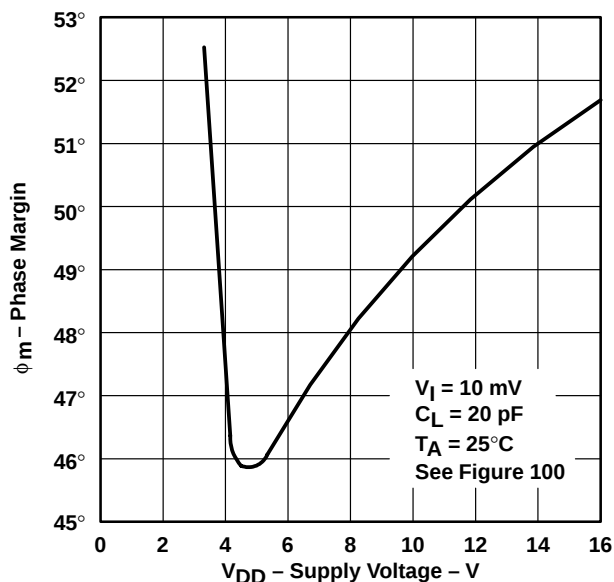


Figure 30

PHASE MARGIN vs FREE-AIR TEMPERATURE

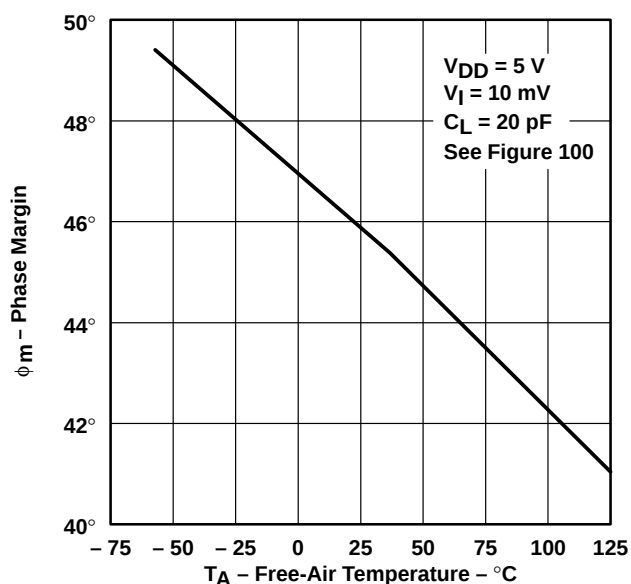


Figure 31

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS (HIGH-BIAS MODE)[†]

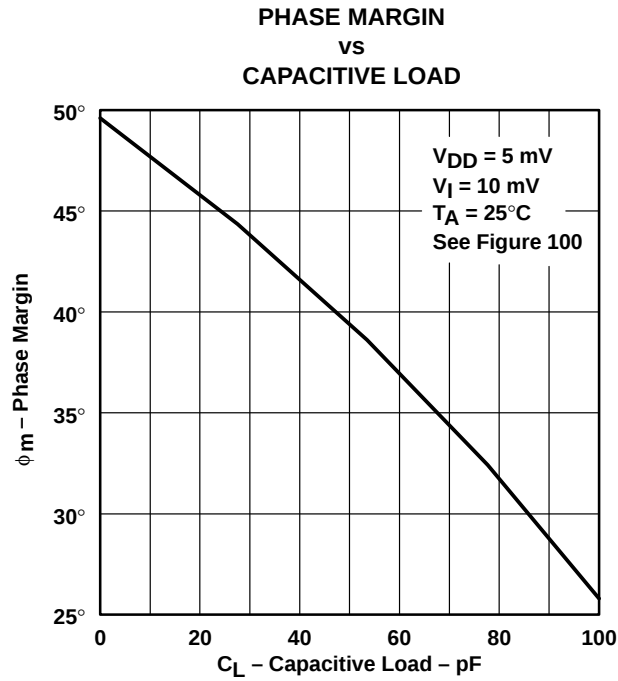


Figure 32

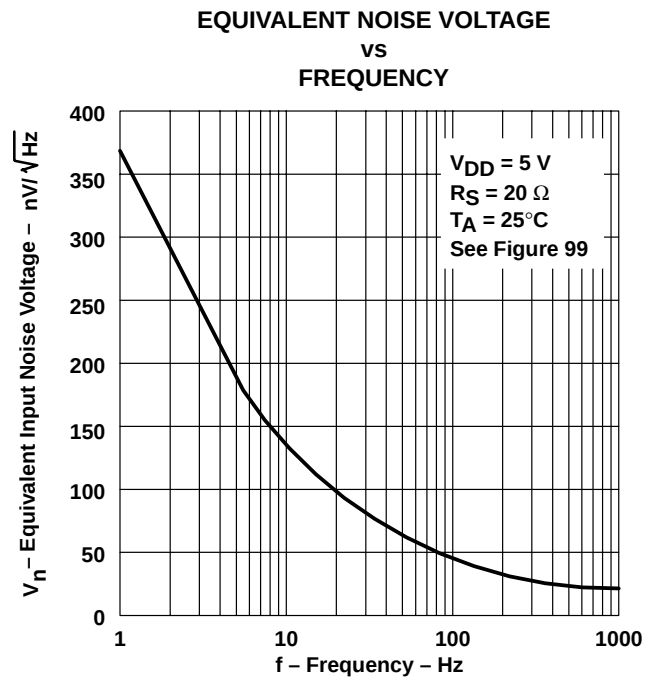


Figure 33

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC271, TLC271A, TLC271B

LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

MEDIUM-BIAS MODE

electrical characteristics at specified free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T _A [†]	TLC271C, TLC271AC, TLC271BC						UNIT
				V _{DD} = 5 V			V _{DD} = 10 V			
				MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO}	Input offset voltage	V _O = 1.4 V, V _{IC} = 0 R _S = 50 Ω, R _I = 100 kΩ	25°C	1.1	10		1.1	10	mV	
			Full range		12		12			
			25°C	0.9	5		0.9	5		
			Full range		6.5		6.5			
			25°C	0.25	2		0.26	2		
			Full range		3		3			
α _{VIO}	Average temperature coefficient of input offset voltage		25°C to 70°C	1.7			2.1		μV/°C	
I _{IO}	Input offset current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.1	60		0.1	60	pA	
			70°C	7	300		7	300		
I _{IB}	Input bias current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.6	60		0.7	60	pA	
			70°C	40	600		50	600		
V _{ICR}	Common-mode input voltage range (see Note 5)		25°C	−0.2 to 4	−0.3 to 4.2		−0.2 to 9	−0.3 to 9.2	V	
			Full range	−0.2 to 3.5			−0.2 to 8.5		V	
V _{OH}	High-level output voltage	V _{ID} = 100 mV, R _L = 100 kΩ	25°C	3.2	3.9		8	8.7	V	
			0°C	3	3.9		7.8	8.7		
			70°C	3	4		7.8	8.7		
V _{OL}	Low-level output voltage	V _{ID} = −100 mV, I _{OL} = 0	25°C	0	50		0	50	mV	
			0°C	0	50		0	50		
			70°C	0	50		0	50		
A _{VD}	Large-signal differential voltage amplification	R _L = 100 kΩ, See Note 6	25°C	25	170		25	275	V/mV	
			0°C	15	200		15	320		
			70°C	15	140		15	230		
CMRR	Common-mode rejection ratio	V _{IC} = V _{ICR} min	25°C	65	91		65	94	dB	
			0°C	60	91		60	94		
			70°C	60	92		60	94		
k _{SVR}	Supply-voltage rejection ratio (ΔV _{DD} /ΔV _{IO})	V _{DD} = 5 V to 10 V V _O = 1.4 V	25°C	70	93		70	93	dB	
			0°C	60	92		60	92		
			70°C	60	94		60	94		
I _{I(SEL)}	Input current (BIAS SELECT)	V _{I(SEL)} = V _{DD} /2	25°C	−130			−160		nA	
I _{DD}	Supply current	V _O = V _{DD} /2, V _{IC} = V _{DD} /2, No load	25°C	105	280		143	300	μA	
			0°C	125	320		173	400		
			70°C	85	220		110	280		

[†] Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

6. At V_{DD} = 5 V, V_O = 0.25 V to 2 V; at V_{DD} = 10 V, V_O = 1 V to 6 V.



MEDIUM-BIAS MODE

electrical characteristics at specified free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T _A [†]	TLC271I, TLC271AI, TLC271BI						UNIT
				V _{DD} = 5 V			V _{DD} = 10 V			
				MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO}	Input offset voltage	V _O = 1.4 V, V _{IC} = 0 V, R _S = 50 Ω, R _L = 100 kΩ	25°C	1.1		10	1.1		10	mV
			Full range	13			13			
			25°C	0.9		5	0.9		5	
			Full range	7			7			
			25°C	0.25		2	0.26		2	
			Full range	3.5			3.5			
α _{VIO}	Average temperature coefficient of input offset voltage		25°C to 85°C	1.7			2.1		μV/°C	
I _{IO}	Input offset current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.1		60	0.1		60	pA
			85°C	24		1000	26		1000	
I _{IB}	Input bias current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.6		60	0.7		60	pA
			85°C	200		2000	220		2000	
V _{ICR}	Common-mode input voltage range (see Note 5)		25°C	−0.2 to 4	−0.3 to 4.2		−0.2 to 9	−0.3 to 9.2	V	
			Full range	−0.2 to 3.5			−0.2 to 8.5		V	
V _{OH}	High-level output voltage	V _{ID} = 100 mV, R _L = 100 kΩ	25°C	3.2	3.9		8	8.7	V	
			−40°C	3	3.9		7.8	8.7		
			85°C	3	4		7.8	8.7		
V _{OL}	Low-level output voltage	V _{ID} = −100 mV, I _{OL} = 0	25°C	0		50	0		50	mV
			−40°C	0		50	0		50	
			85°C	0		50	0		50	
A _{VD}	Large-signal differential voltage amplification	R _L = 100 kΩ, See Note 6	25°C	25	170		25	275	V/mV	
			−40°C	15	270		15	390		
			85°C	15	130		15	220		
CMRR	Common-mode rejection ratio	V _{IC} = V _{ICRmin}	25°C	65	91		65	94	dB	
			−40°C	60	90		60	93		
			85°C	60	90		60	94		
k _{SVR}	Supply-voltage rejection ratio (ΔV _{DD} /ΔV _{IO})	V _{DD} = 5 V to 10 V V _O = 1.4 V	25°C	70	93		70	93	dB	
			−40°C	60	91		60	91		
			85°C	60	94		60	94		
I _{I(SEL)}	Input current (BIAS SELECT)	V _{I(SEL)} = V _{DD} /2	25°C	−130			−160		nA	
I _{DD}	Supply current	V _O = V _{DD} /2, V _{IC} = V _{DD} /2, No load	25°C	105		280	143		300	μA
			−40°C	158		400	225		450	
			85°C	80		200	103		260	

[†] Full range is –40°C to 85°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

6. At V_{DD} = 5 V, V_O = 0.25 V to 2 V; at V_{DD} = 10 V, V_O = 1 V to 6 V.

TLC271, TLC271A, TLC271B

LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

MEDIUM-BIAS MODE

electrical characteristics at specified free-air temperature (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T _A [†]	TLC271M						UNIT
			V _{DD} = 5 V			V _{DD} = 10 V			
			MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO} Input offset voltage	V _O = 1.4 V, V _{IC} = 0 V, R _S = 50 Ω, R _L = 100 kΩ	25°C	1.1		10	1.1		10	mV
		Full range			12			12	
αV _{IO} Average temperature coefficient of input offset voltage		25°C to 125°C	1.7			2.1			μV/°C
I _{IO} Input offset current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.1		60	0.1		60	pA
		125°C	1.4		15	1.8		15	nA
I _{IB} Input bias current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.6		60	0.7		60	pA
		125°C	9		35	10		35	nA
V _{ICR} Common-mode input voltage range (see Note 5)		25°C	0 to 4	–0.3 to 4.2		0 to 9	–0.3 to 9.2		V
		Full range	0 to 3.5			0 to 8.5			V
V _{OH} High-level output voltage	V _{ID} = 100 mV, R _L = 100 kΩ	25°C	3.2	3.9		8	8.7		V
		–55°C	3	3.9		7.8	8.6		
		125°C	3	4		7.8	8.6		
V _{OL} Low-level output voltage	V _{ID} = –100 mV, I _{OL} = 0	25°C	0		50	0		50	mV
		–55°C	0		50	0		50	
		125°C	0		50	0		50	
A _{VD} Large-signal differential voltage amplification	R _L = 10 kΩ See Note 6	25°C	25	170		25	275		V/mV
		–55°C	15	290		15	420		
		125°C	15	120		15	190		
CMRR Common-mode rejection ratio	V _{IC} = V _{ICRmin}	25°C	65	91		65	94		dB
		–55°C	60	89		60	93		
		125°C	60	91		60	93		
k _{SVR} Supply-voltage rejection ratio (ΔV _{DD} /ΔV _{IO})	V _{DD} = 5 V to 10 V V _O = 1.4 V	25°C	70	93		70	93		dB
		–55°C	60	91		60	91		
		125°C	60	94		60	94		
I _{I(SEL)} Input current (BIAS SELECT)	V _{I(SEL)} = V _{DD} /2	25°C	–130			–160			nA
I _{DD} Supply current	V _O = V _{DD} /2, V _{IC} = V _{DD} /2, No load	25°C	105		280	143		300	μA
		–55°C	170		440	245		500	
		125°C	70		180	90		240	

[†] Full range is –55°C to 125°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

6. At V_{DD} = 5 V, V_O = 0.25 V to 2 V; at V_{DD} = 10 V, V_O = 1 V to 6 V.



MEDIUM-BIAS MODE

operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS	T_A	TLC271C, TLC271AC, TLC271BC			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C	0.43		V/ μ s
			0°C	0.46		
			70°C	0.36		
		$V_{I(PP)} = 2.5\text{ V}$	25°C	0.40		
			0°C	0.43		
			70°C	0.34		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 20\text{ }\Omega$, 25°C		32		nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$, See Figure 98	$C_L = 20\text{ pF}$, See Figure 98	25°C	55		kHz
			0°C	60		
			70°C	50		
B_1 Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$, See Figure 100	25°C	525		kHz
			0°C	600		
			70°C	400		
ϕ_m Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$, See Figure 100	$f = B_1$, See Figure 100	25°C	40°		
			0°C	41°		
			70°C	39°		

operating characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS	T_A	TLC271C, TLC271AC, TLC271BC			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C	0.62		V/ μ s
			0°C	0.67		
			70°C	0.51		
		$V_{I(PP)} = 5.5\text{ V}$	25°C	0.56		
			0°C	0.61		
			70°C	0.46		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 20\text{ }\Omega$, 25°C		32		nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$, See Figure 98	$C_L = 20\text{ pF}$, See Figure 98	25°C	35		kHz
			0°C	40		
			70°C	30		
B_1 Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$, See Figure 100	25°C	635		kHz
			0°C	710		
			70°C	510		
ϕ_m Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$, See Figure 100	$f = B_1$, See Figure 100	25°C	43°		
			0°C	44°		
			70°C	42°		

TLC271, TLC271A, TLC271B
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

MEDIUM-BIAS MODE

operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS	T_A	TLC271I, TLC271AI, TLC271BI			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C	0.43		V/ μ s
			–40°C	0.51		
			85°C	0.35		
		$V_{I(PP)} = 2.5\text{ V}$	25°C	0.40		
			–40°C	0.48		
			85°C	0.32		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 20\text{ }\Omega$, 25°C		32		nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	25°C		55		kHz
		–40°C		75		
		85°C		45		
B_1 Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$, 25°C		525		MHz
		–40°C		770		
		85°C		370		
ϕ_m Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$, $f = B_1$, See Figure 100	25°C		40°		
		–40°C		43°		
		85°C		38°		

operating characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS	T_A	TLC271I, TLC271AI, TLC271BI			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C	0.62		V/ μ s
			–40°C	0.77		
			85°C	0.47		
		$V_{I(PP)} = 5.5\text{ V}$	25°C	0.56		
			–40°C	0.70		
			85°C	0.44		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 20\text{ }\Omega$, 25°C		32		nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output-swing bandwidth	$V_O = V_{OH}$, ³ $R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	25°C		35		kHz
		–40°C		45		
		85°C		25		
B_1 Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$, 25°C		635		kHz
		–40°C		880		
		85°C		480		
ϕ_m Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$, $f = B_1$, See Figure 100	25°C		43°		
		–40°C		46°		
		85°C		41°		



MEDIUM-BIAS MODE

operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS	T_A	TLC271M			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C	0.43		V/ μ s
			–55°C	0.54		
			125°C	0.29		
		$V_{I(PP)} = 2.5\text{ V}$	25°C	0.40		
			–55°C	0.50		
			125°C	0.28		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 20\text{ }\Omega$, 25°C		32		nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	25°C		55		kHz
		–55°C		80		
		125°C		40		
B_1 Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$, 25°C		525		kHz
		–55°C		850		
		125°C		330		
ϕ_m Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$, $f = B_1$, See Figure 100	25°C		40°		
		–55°C		43°		
		125°C		36°		

operating characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS	T_A	TLC271M			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C	0.62		V/ μ s
			–55°C	0.81		
			125°C	0.38		
		$V_{I(PP)} = 5.5\text{ V}$	25°C	0.56		
			–55°C	0.73		
			125°C	0.35		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 20\text{ }\Omega$, 25°C		32		nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 100\text{ k}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	25°C		35		kHz
		–55°C		50		
		125°C		20		
B_1 Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$, 25°C		635		kHz
		–55°C		960		
		125°C		440		
ϕ_m Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$, $f = B_1$, See Figure 100	25°C		43°		
		–55°C		47°		
		125°C		39°		

TLC271, TLC271A, TLC271B
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)

Table of Graphs

			FIGURE
V_{IO}	Input offset voltage	Distribution	34, 35
α_{VIO}	Temperature coefficient	Distribution	36, 37
V_{OH}	High-level output voltage	vs High-level output current	38, 39
		vs Supply voltage	40
		vs Free-air temperature	41
V_{OL}	Low-level output voltage	vs Common-mode input voltage	42, 43
		vs Differential input voltage	44
		vs Free-air temperature	45
		vs Low-level output current	46, 47
A_{VD}	Large-signal differential voltage amplification	vs Supply voltage	48
		vs Free-air temperature	49
		vs Frequency	60, 61
I_{IB}	Input bias current	vs Free-air temperature	50
I_{IO}	Input offset current	vs Free-air temperature	50
V_I	Maximum Input voltage	vs Supply voltage	51
I_{DD}	Supply current	vs Supply voltage	52
		vs Free-air temperature	53
SR	Slew rate	vs Supply voltage	54
		vs Free-air temperature	55
	Bias-select current	vs Supply voltage	56
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency	57
B_1	Unity-gain bandwidth	vs Free-air temperature	58
		vs Supply voltage	59
ϕ_m	Phase margin	vs Supply voltage	62
		vs Free-air temperature	63
		vs Capacitive load	64
V_n	Equivalent input noise voltage	vs Frequency	65
	Phase shift	vs Frequency	60, 61

TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)[†]

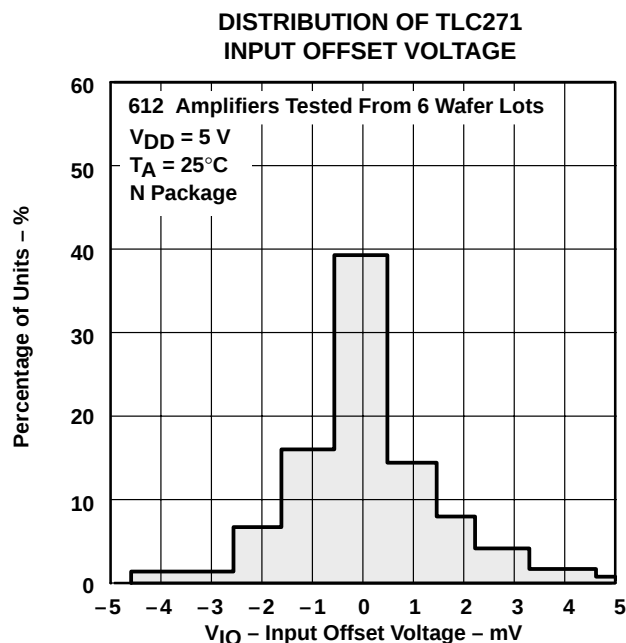


Figure 34

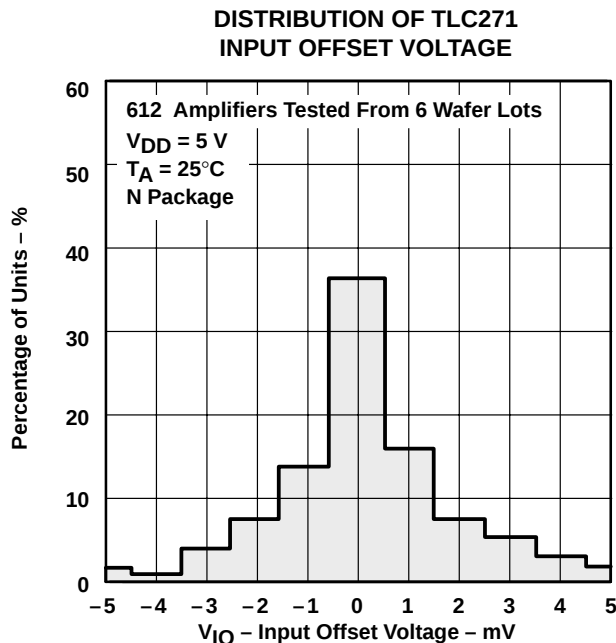


Figure 35

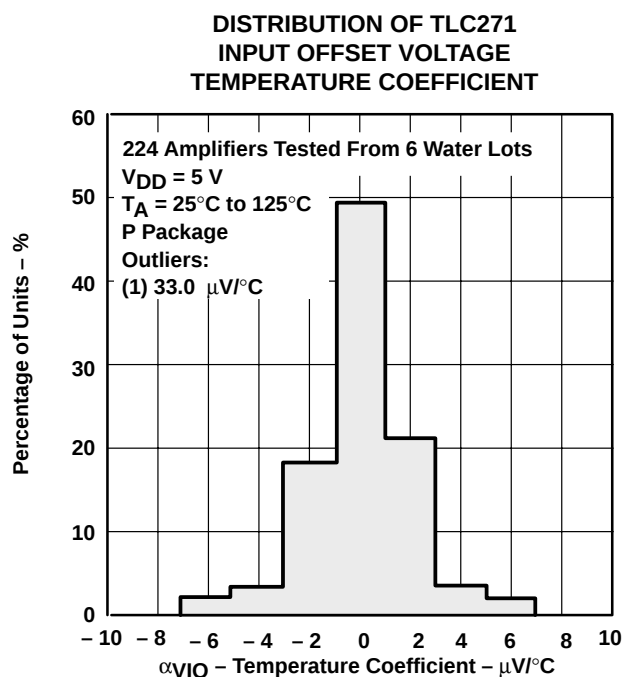


Figure 36

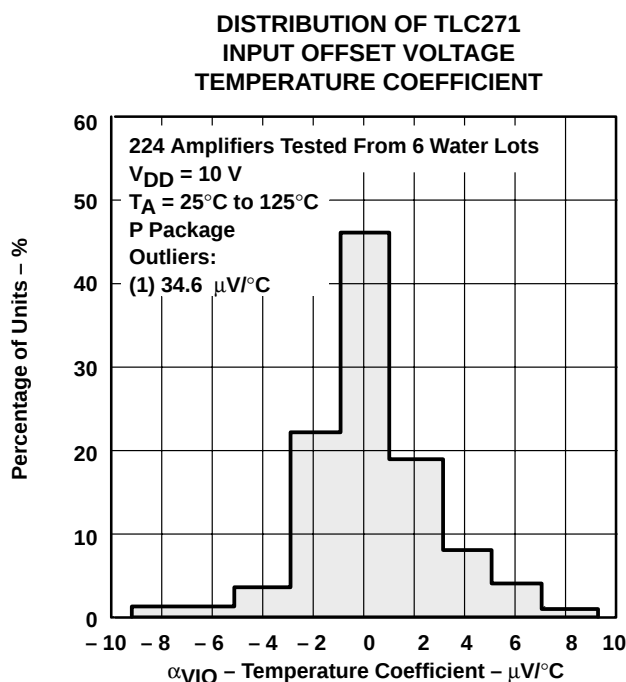


Figure 37

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC271, TLC271A, TLC271B

LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)[†]

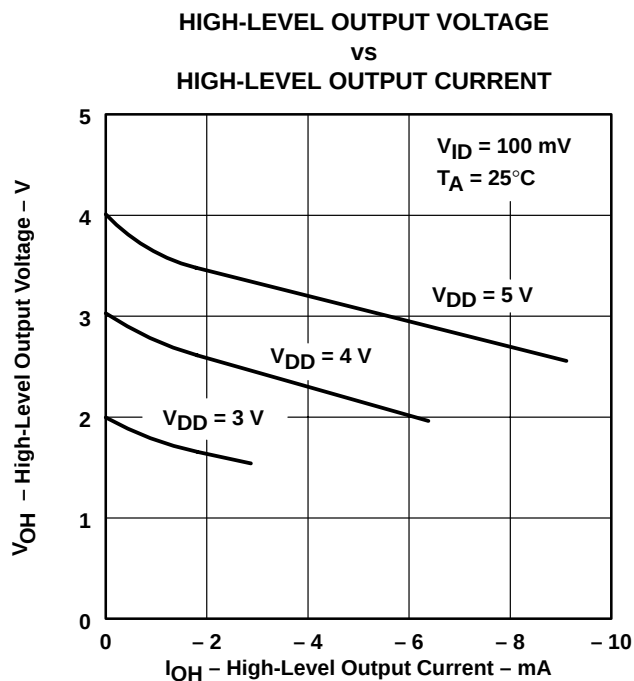


Figure 38

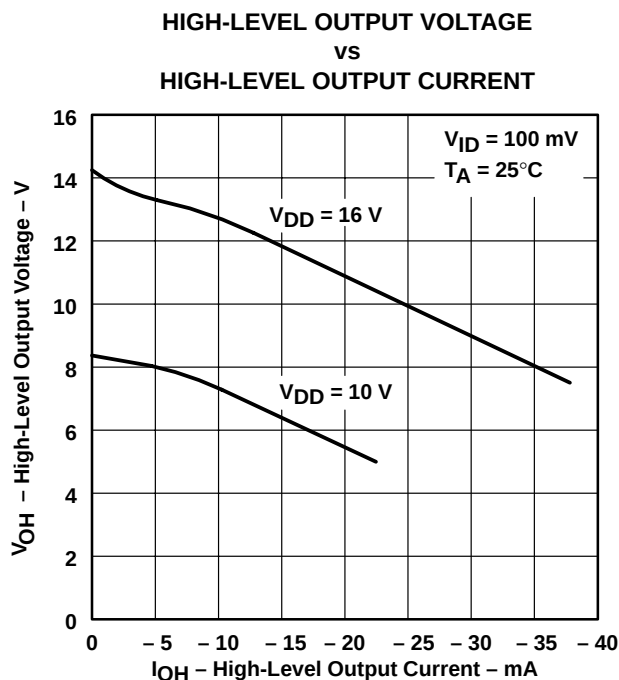


Figure 39

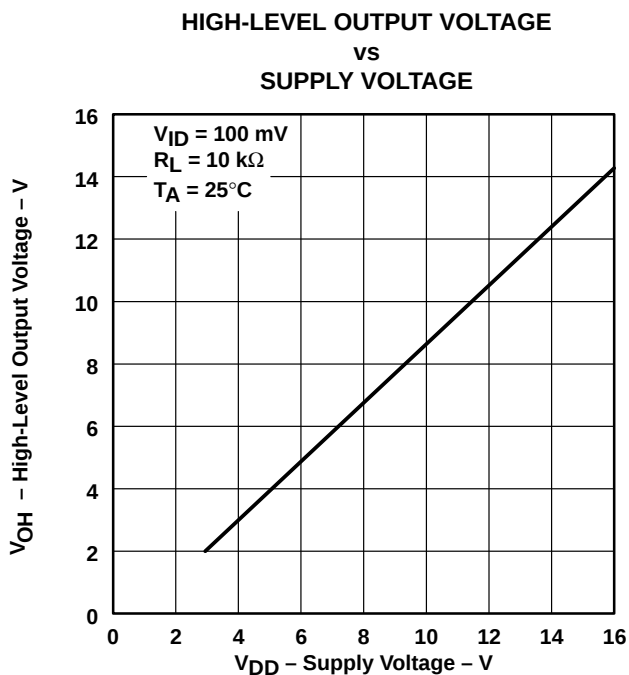


Figure 40

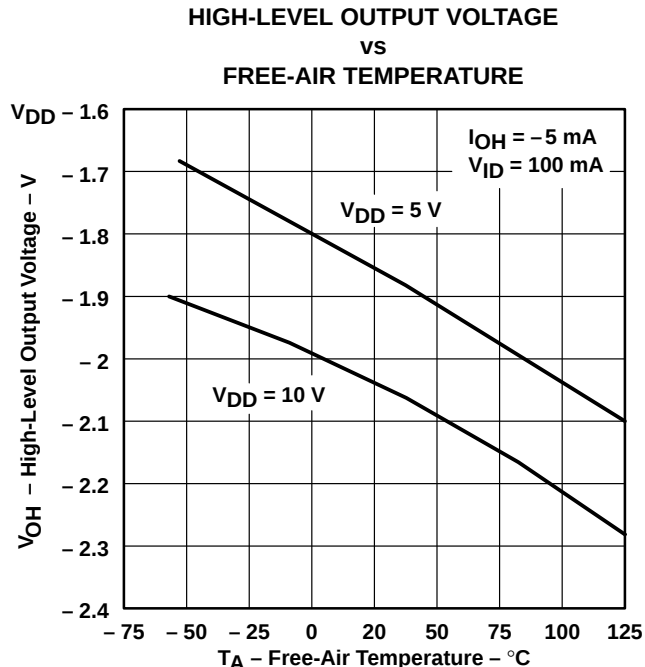


Figure 41

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)[†]

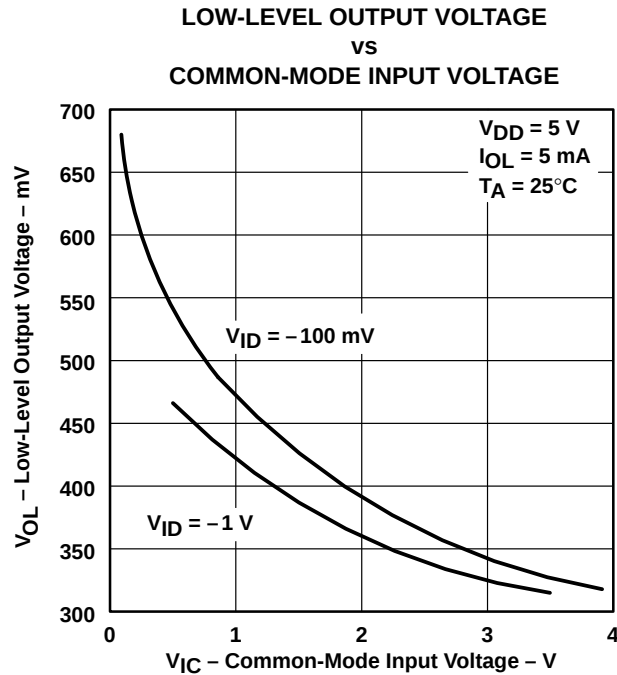


Figure 42

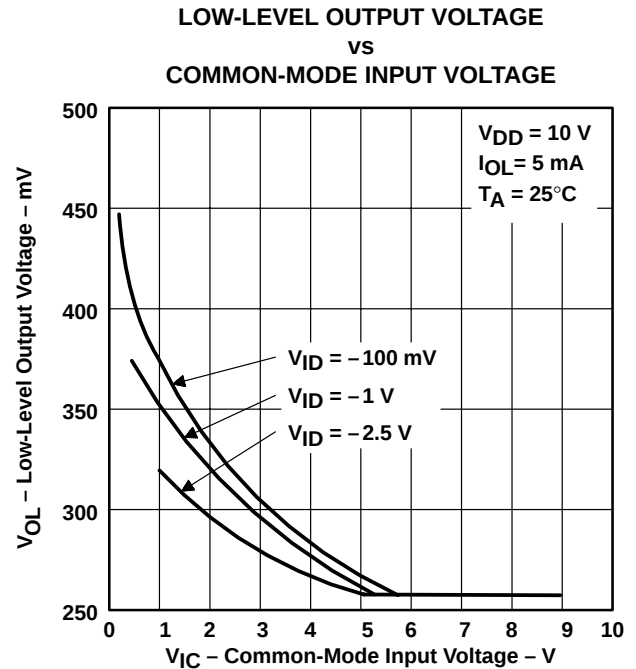


Figure 43

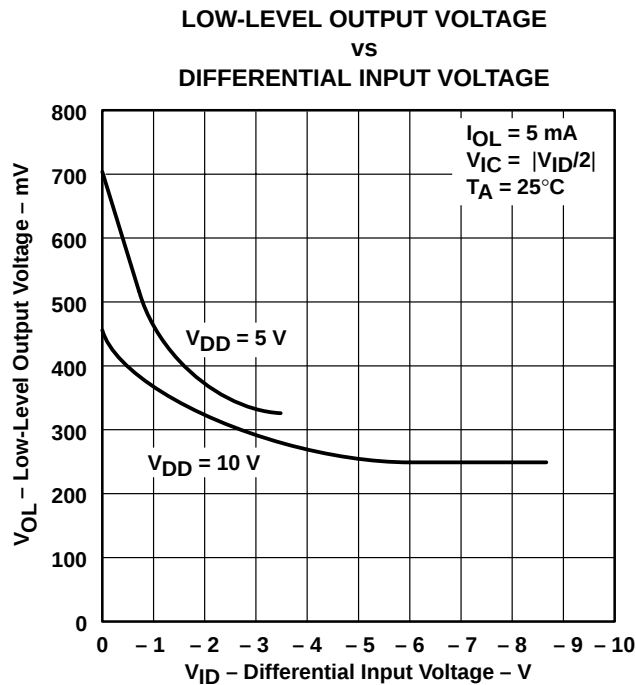


Figure 44

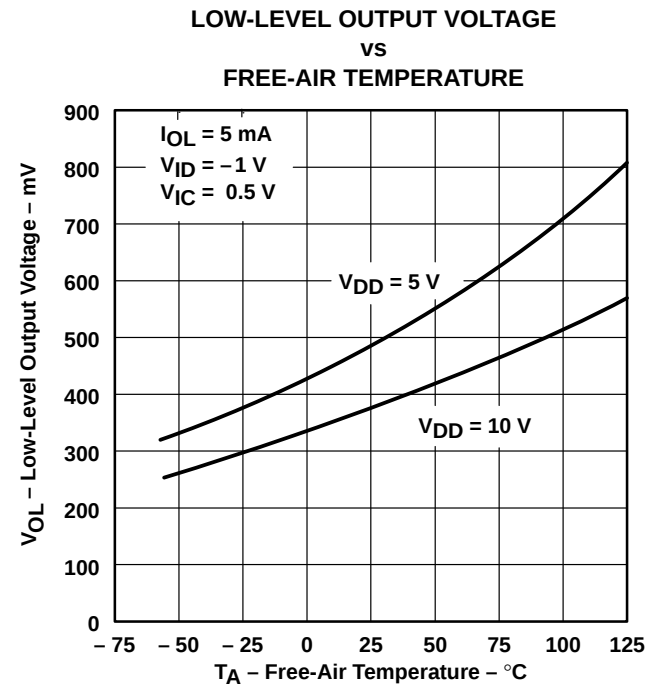


Figure 45

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC271, TLC271A, TLC271B

LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)[†]

LOW-LEVEL OUTPUT VOLTAGE
vs
LOW-LEVEL OUTPUT CURRENT

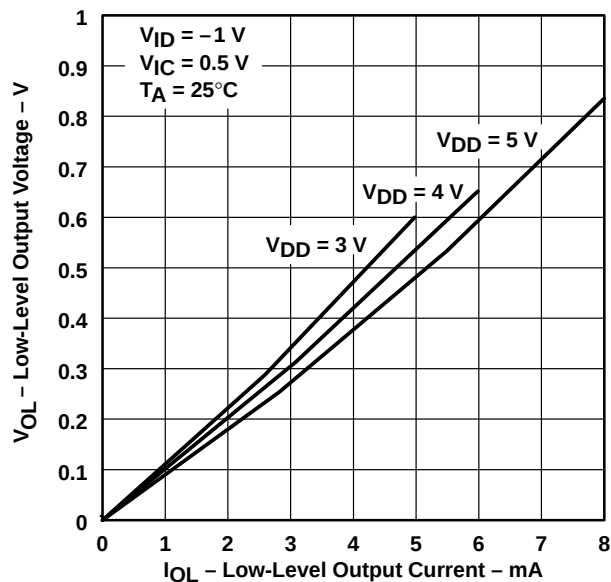


Figure 46

LOW-LEVEL OUTPUT VOLTAGE
vs
LOW-LEVEL OUTPUT CURRENT

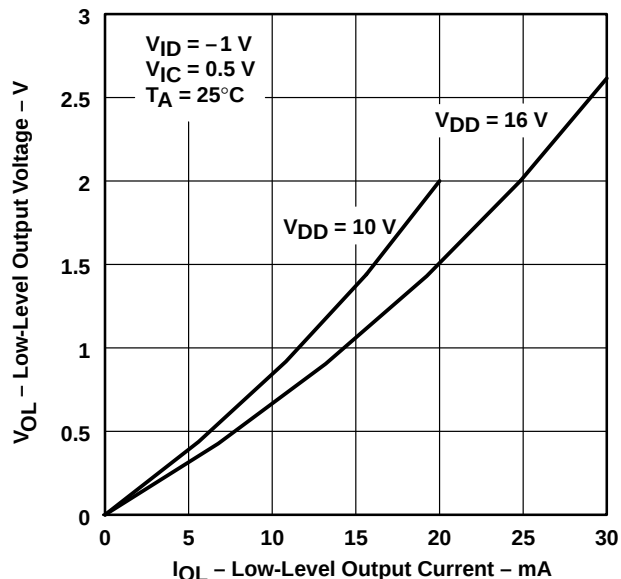


Figure 47

LARGE-SIGNAL
DIFFERENTIAL VOLTAGE AMPLIFICATION
vs
SUPPLY VOLTAGE

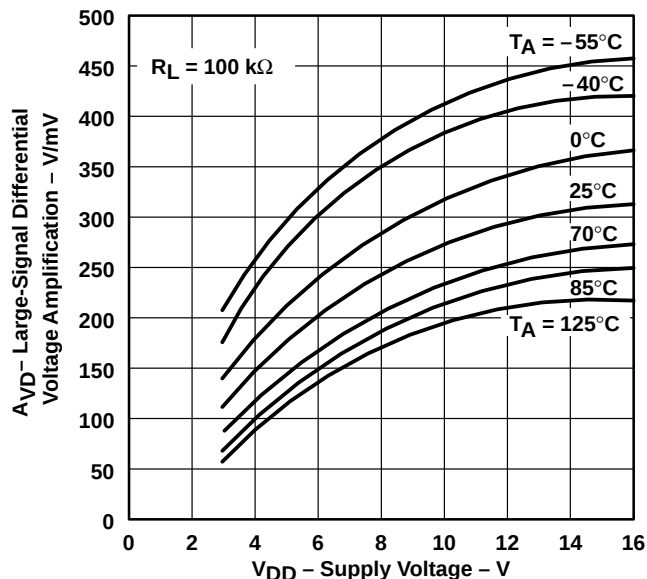


Figure 48

LARGE-SIGNAL
DIFFERENTIAL VOLTAGE AMPLIFICATION
vs
FREE-AIR TEMPERATURE

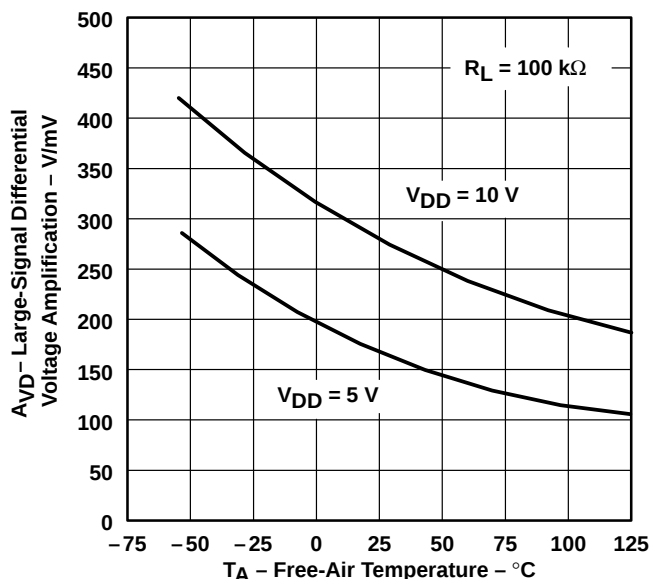
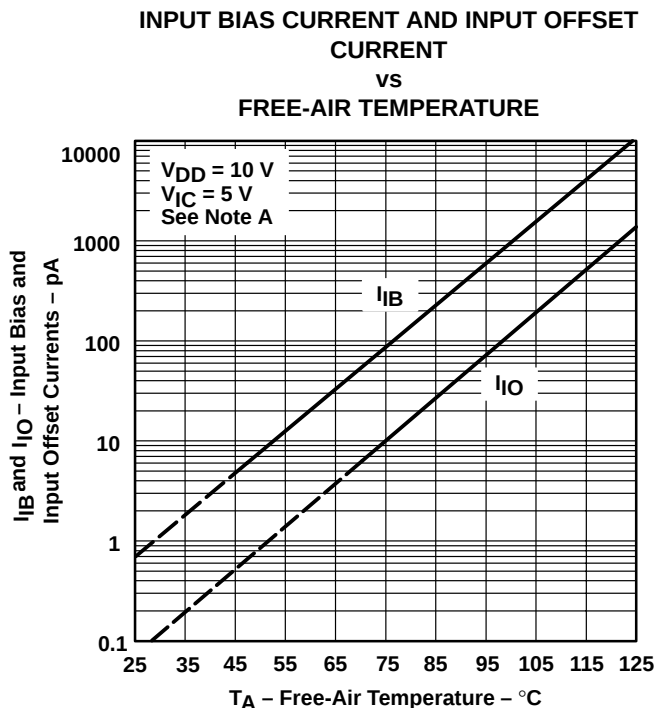


Figure 49

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)[†]



NOTE A: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

Figure 50

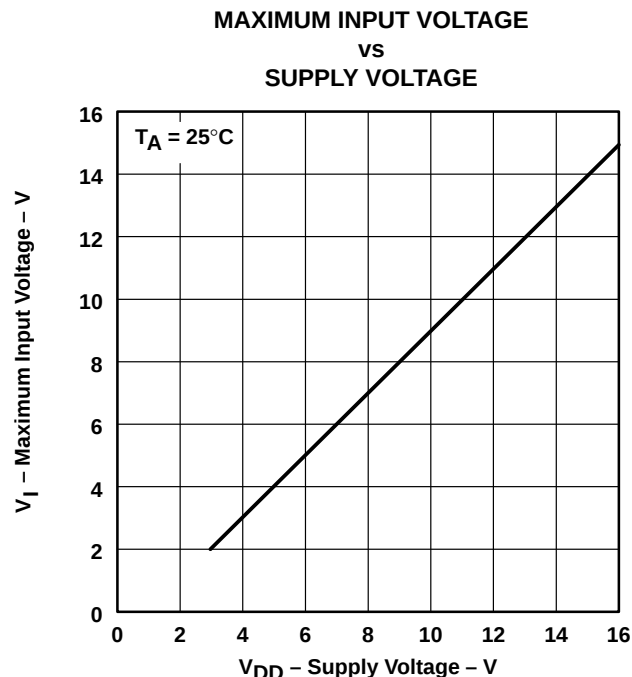


Figure 51

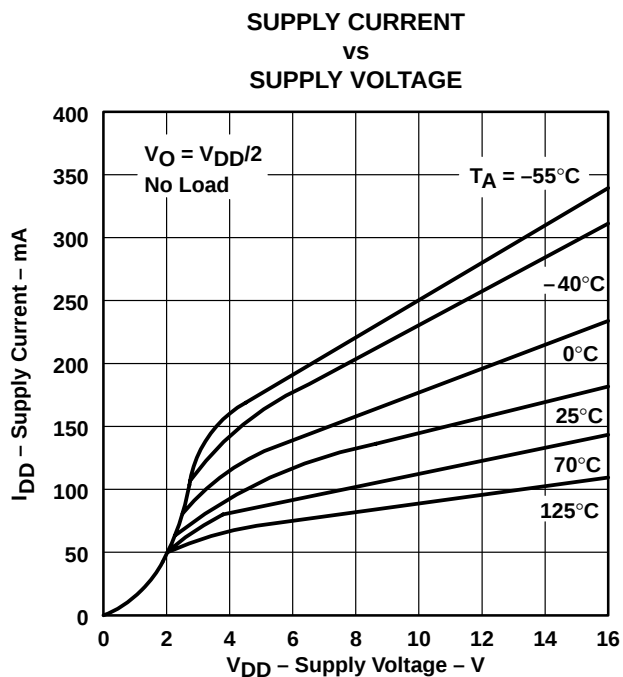


Figure 52

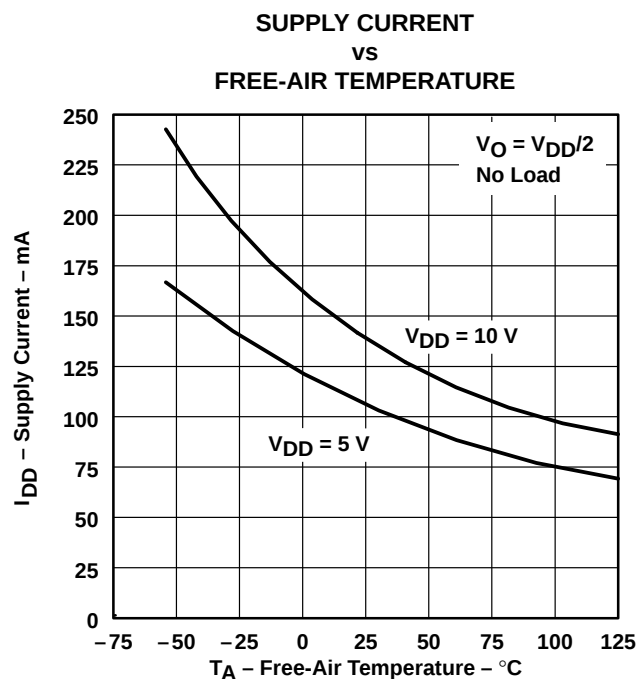


Figure 53

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC271, TLC271A, TLC271B

LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)[†]

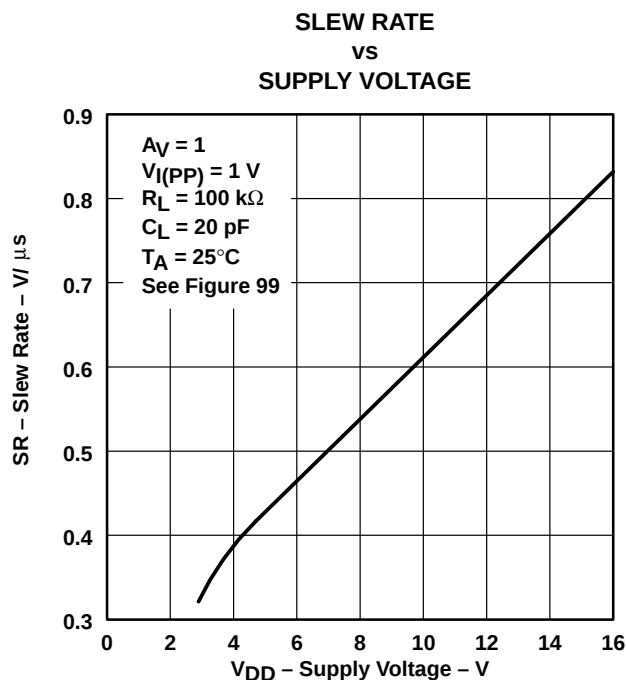


Figure 54

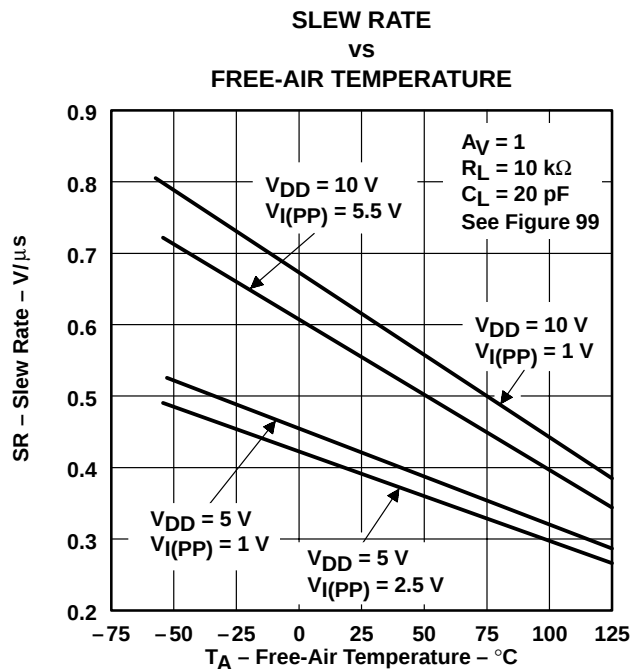


Figure 55

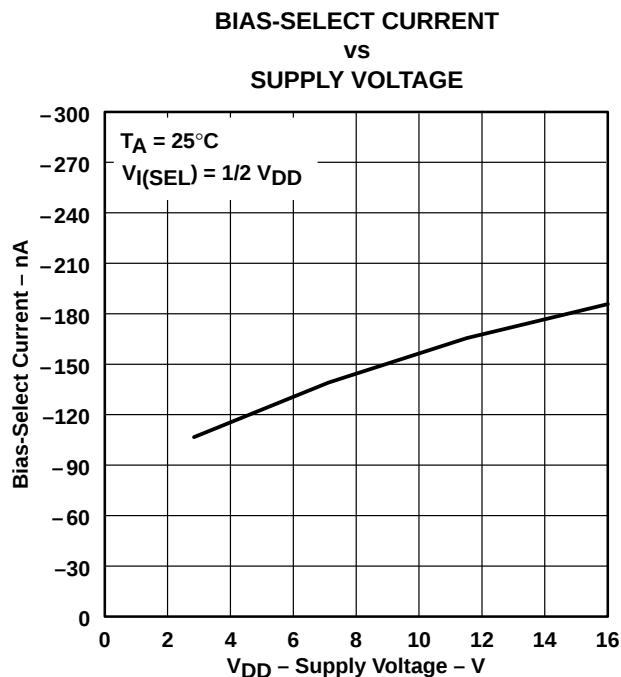


Figure 56

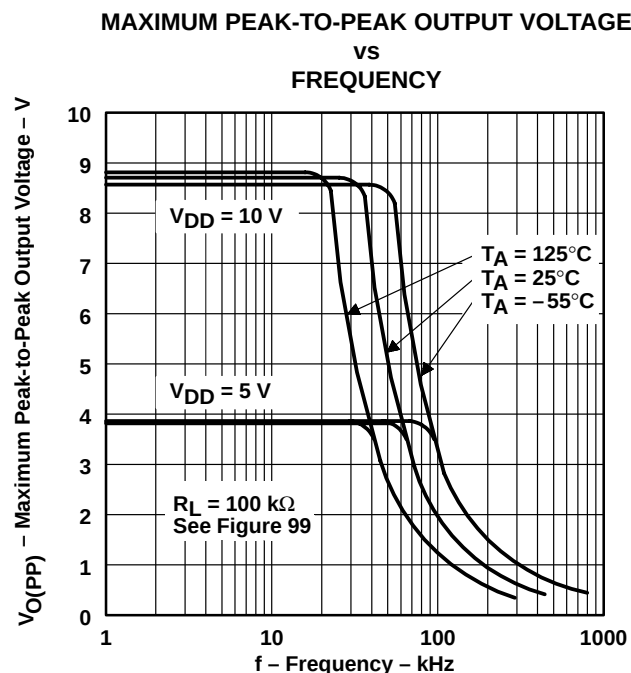


Figure 57

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)[†]

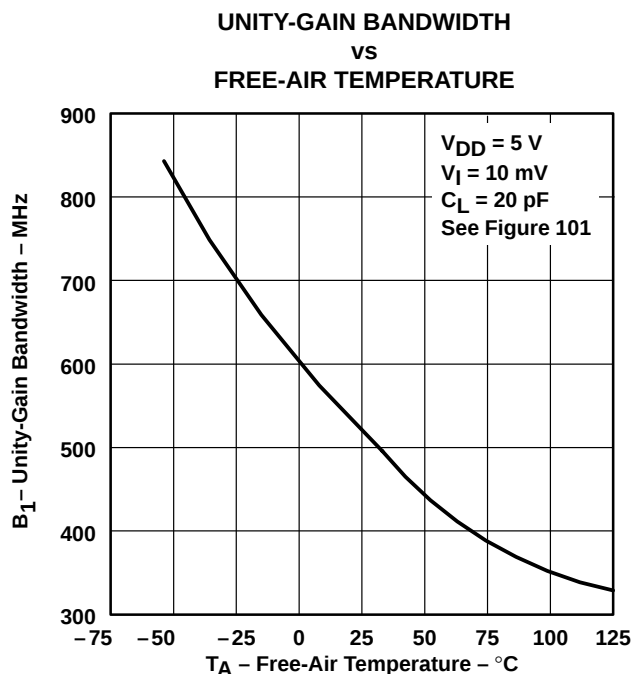


Figure 58

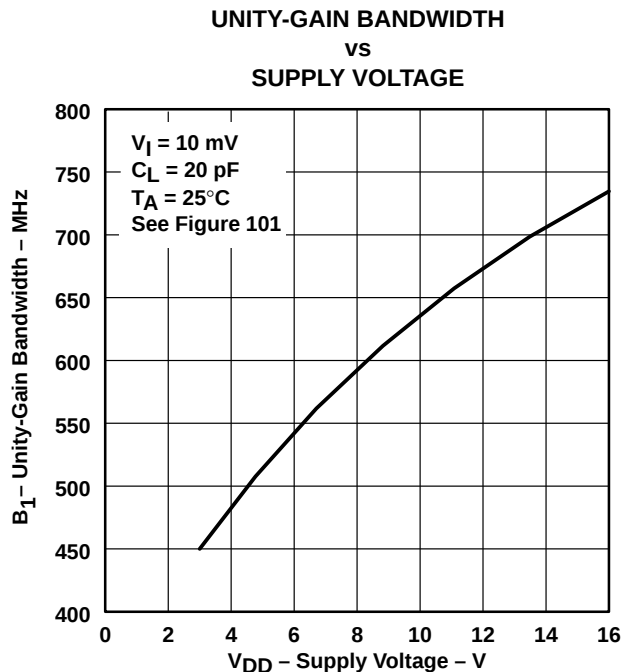


Figure 59

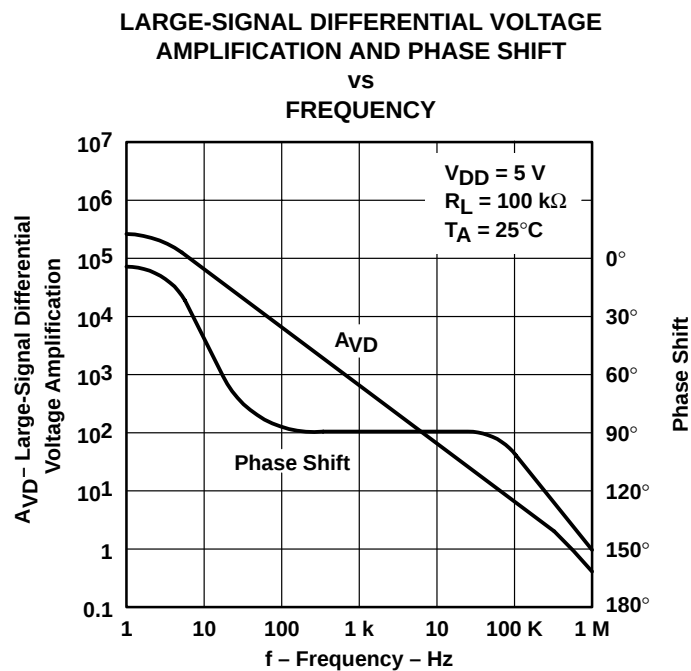


Figure 60

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC271, TLC271A, TLC271B

LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)[†]

LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION AND PHASE SHIFT VS FREQUENCY

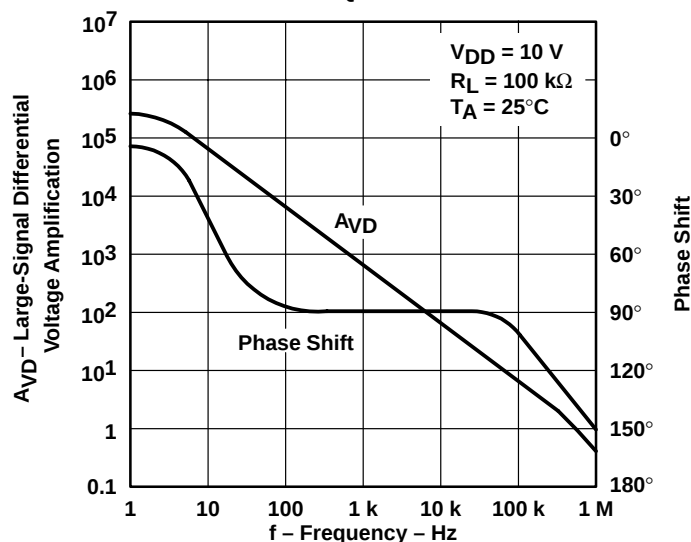


Figure 61

PHASE MARGIN VS SUPPLY VOLTAGE

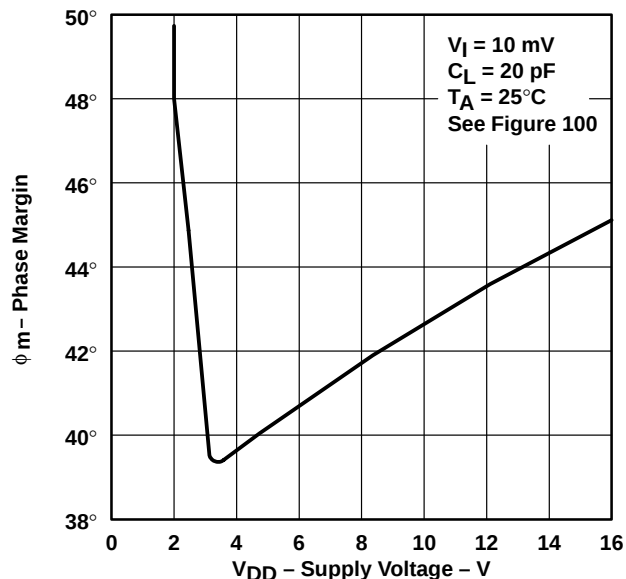


Figure 62

PHASE MARGIN VS FREE-AIR TEMPERATURE

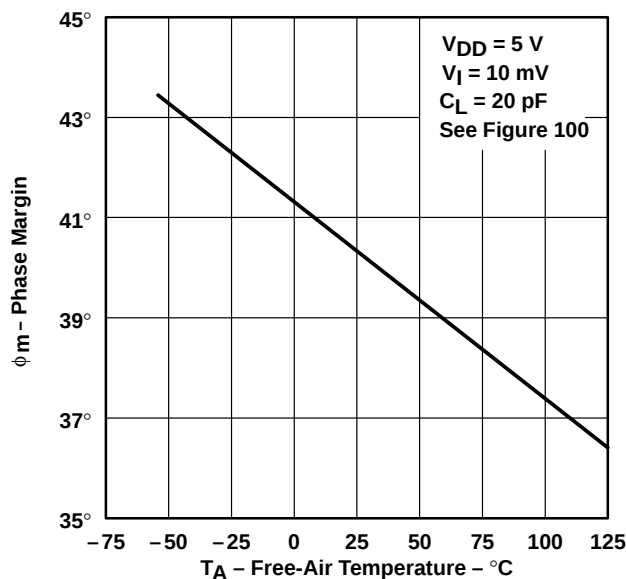


Figure 63

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS (MEDIUM-BIAS MODE)[†]

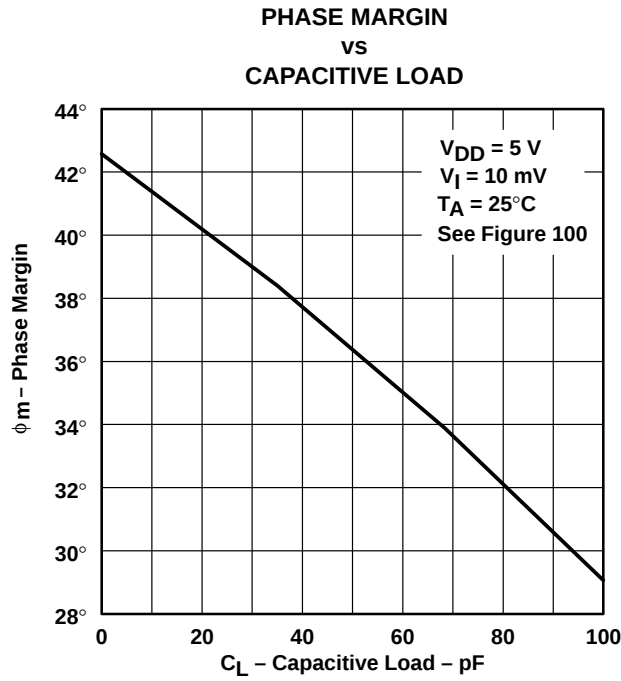


Figure 64

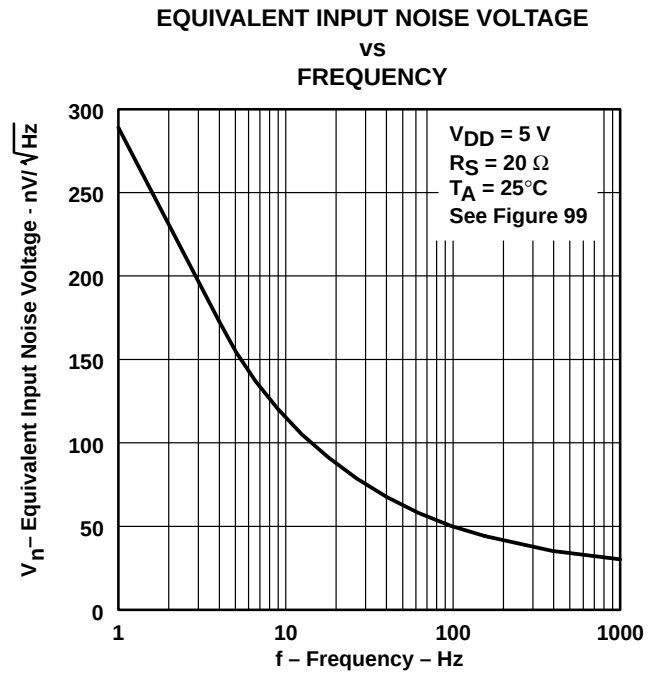


Figure 65

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC271, TLC271A, TLC271B

LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

LOW-BIAS MODE

electrical characteristics at specified free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T _A [†]	TLC271C, TLC271AC, TLC271BC						UNIT
				V _{DD} = 5 V			V _{DD} = 10 V			
				MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO}	Input offset voltage	V _O = 1.4 V, V _{IC} = 0 V, R _S = 50 Ω, R _I = 1 MΩ	25°C	1.1	10	1.1	10	mV		
			Full range	12	12					
			25°C	0.9	5	0.9	5			
			Full range	6.5	6.5					
			25°C	0.24	2	0.26	2			
			Full range	3	3					
αV _{IO}	Average temperature coefficient of input offset voltage		25°C to 70°C	1.1	1	μV/°C				
I _{IO}	Input offset current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.1	60	0.1	60	pA		
			70°C	7	300	8	300			
I _{IB}	Input bias current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.6	60	0.7	60	pA		
			70°C	40	600	50	600			
V _{ICR}	Common-mode input voltage range (see Note 5)		25°C	−0.2 to 4	−0.3 to 4.2	−0.2 to 9	−0.3 to 9.2	V		
			Full range	−0.2 to 3.5	−0.2 to 8.5	V				
V _{OH}	High-level output voltage	V _{ID} = 100 mV, R _L = 1 MΩ	25°C	3.2	4.1	8	8.9	V		
			0°C	3	4.1	7.8	8.9			
			70°C	3	4.2	7.8	8.9			
V _{OL}	Low-level output voltage	V _{ID} = −100 mV, I _{OL} = 0	25°C	0	50	0	50	mV		
			0°C	0	50	0	50			
			70°C	0	50	0	50			
A _{VD}	Large-signal differential voltage amplification	R _L = 1 MΩ, See Note 6	25°C	50	520	50	870	V/mV		
			0°C	50	700	50	1030			
			70°C	50	380	50	660			
CMRR	Common-mode rejection ratio	V _{IC} = V _{ICRmin}	25°C	65	94	65	97	dB		
			0°C	60	95	60	97			
			70°C	60	95	60	97			
k _{SVR}	Supply-voltage rejection ratio (ΔV _{DD} /ΔV _{IO})	V _{DD} = 5 V to 10 V V _O = 1.4 V	25°C	70	97	70	97	dB		
			0°C	60	97	60	97			
			70°C	60	98	60	98			
I _{I(SEL)}	Input current (BIAS SELECT)	V _{I(SEL)} = V _{DD}	25°C	65	95	nA				
I _{DD}	Supply current	V _O = V _{DD} /2, V _{IC} = V _{DD} /2, No load	25°C	10	17	14	23	μA		
			0°C	12	21	18	33			
			70°C	8	14	11	20			

[†] Full range is 0°C to 70°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

6. At V_{DD} = 5 V, V_O = 0.25 V to 2 V; at V_{DD} = 10 V, V_O = 1 V to 6 V.



LOW-BIAS MODE

electrical characteristics at specified free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T _A [†]	TLC271I, TLC271AI, TLC271BI						UNIT			
				V _{DD} = 5 V			V _{DD} = 10 V						
				MIN	TYP	MAX	MIN	TYP	MAX				
V _{IO}	Input offset voltage	TLC271I TLC271AI TLC271BI	V _O = 1.4 V, V _{IC} = 0 V, R _S = 50 Ω, R _L = 1 MΩ	25°C	1.1		10		1.1		10		mV
				Full range			13		13				
				25°C	0.9		5		0.9		5		
				Full range			7		7				
				25°C	0.24		2		0.26		2		
				Full range			3.5		3.5				
αV _{IO}	Average temperature coefficient of input offset voltage		25°C to 85°C	1.1				1				μV/°C	
I _{IO}	Input offset current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.1		60		0.1		60		pA	
			85°C	24		1000		26		1000			
I _{IB}	Input bias current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.6		60		0.7		60		pA	
			85°C	200		2000		220		2000			
V _{ICR}	Common-mode input voltage range (see Note 5)		25°C	−0.2 to 4	−0.3 to 4.2			−0.2 to 9	−0.3 to 9.2			V	
			Full range	−0.2 to 3.5			−0.2 to 8.5				V		
V _{OH}	High-level output voltage	V _{ID} = 100 mV, R _L = 1 MΩ	25°C	3	4.1			8	8.9			V	
			−40°C	3	4.1			7.8	8.9				
			85°C	3	4.2			7.8	8.9				
V _{OL}	Low-level output voltage	V _{ID} = −100 mV, I _{OL} = 0	25°C	0		50		0		50		mV	
			−40°C	0		50		0		50			
			85°C	0		50		0		50			
A _{VD}	Large-signal differential voltage amplification	R _L = 1 MΩ See Note 6	25°C	50	520			50	870			V/mV	
			−40°C	50	900			50	1550				
			85°C	50	330			50	585				
CMRR	Common-mode rejection ratio	V _{IC} = V _{ICRmin}	25°C	65	94			65	97			dB	
			−40°C	60	95			60	97				
			85°C	60	95			60	98				
k _{SVR}	Supply-voltage rejection ratio (ΔV _{DD} /ΔV _{IO})	V _{DD} = 5 V to 10 V V _O = 1.4 V	25°C	70	97			70	97			dB	
			−40°C	60	97			60	97				
			85°C	60	98			60	98				
I _{I(SEL)}	Input current (BIAS SELECT)	V _{I(SEL)} = V _{DD}	25°C	65				95				nA	
I _{DD}	Supply current	V _O = V _{DD} /2, V _{IC} = V _{DD} /2, No load	25°C	10		17		14		23		μA	
			−40°C	16		27		25		43			
			85°C	17		13		10		18			

[†] Full range is –40 to 85°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

6. At V_{DD} = 5 V, V_O = 0.25 V to 2 V; at V_{DD} = 10 V, V_O = 1 V to 6 V.

TLC271, TLC271A, TLC271B

LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

LOW-BIAS MODE

electrical characteristics at specified free-air temperature (unless otherwise noted)

PARAMETER		TEST CONDITIONS	T _A [†]	TLC271M						UNIT
				V _{DD} = 5 V			V _{DD} = 10 V			
				MIN	TYP	MAX	MIN	TYP	MAX	
V _{IO}	Input offset voltage	V _O = 1.4 V, V _{IC} = 0 V, R _S = 50 Ω, R _L = 1 MΩ	25°C	1.1		10	1.1		10	mV
		Full range			12			12		
α _{VIO}	Average temperature coefficient of input offset voltage		25°C to 125°C	1.4			1.4			μV/°C
I _{IO}	Input offset current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.1		60	0.1		60	pA
			125°C	1.4		15	1.8		15	nA
I _{IB}	Input bias current (see Note 4)	V _O = V _{DD} /2, V _{IC} = V _{DD} /2	25°C	0.6		60	0.7		60	pA
			125°C	9		35	10		35	nA
V _{ICR}	Common-mode input voltage range (see Note 5)		25°C	0 to 4	–0.3 to 4.2		0 to 9	–0.3 to 9.2		V
			Full range	0 to 3.5		0 to 8.5			V	
V _{OH}	High-level output voltage	V _{ID} = 100 mV, R _L = 1 MΩ	25°C	3.2	4.1		8	8.9		V
			–55°C	3	4.1		7.8	8.8		
			125°C	3	4.2		7.8	9		
V _{OL}	Low-level output voltage	V _{ID} = –100 mV, I _{OL} = 0	25°C	0		50	0		50	mV
			–55°C	0		50	0		50	
			125°C	0		50	0		50	
A _{VD}	Large-signal differential voltage amplification	R _L = 1 MΩ, See Note 6	25°C	50	520		50	870		V/mV
			–55°C	25	1000		25	1775		
			125°C	25	200		25	380		
CMRR	Common-mode rejection ratio	V _{IC} = V _{ICRmin}	25°C	65	94		65	97		dB
			–55°C	60	95		60	97		
			125°C	60	85		60	91		
k _{SVR}	Supply-voltage rejection ratio (ΔV _{DD} /ΔV _{IO})	V _{DD} = 5 V to 10 V V _O = 1.4 V	25°C	70	97		70	97		dB
			–55°C	60	97		60	97		
			125°C	60	98		60	98		
I _{I(SEL)}	Input current (BIAS SELECT)	V _{I(SEL)} = V _{DD}	25°C	65			95			nA
I _{DD}	Supply current	V _O = V _{DD} /2, V _{IC} = V _{DD} /2, No load	25°C	10		17	14		23	μA
			–55°C	17		30	28		48	
			125°C	7		12	9		15	

† Full range is –55°C to 125°C.

NOTES: 4. The typical values of input bias current and input offset current below 5 pA were determined mathematically.

5. This range also applies to each input individually.

6. At V_{DD} = 5 V, V_O = 0.25 V to 2 V; at V_{DD} = 10 V, V_O = 1 V to 6 V.



LOW-BIAS MODE

operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS		T_A	TLC271C, TLC271AC, TLC271BC			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C		0.03		V/ μ s
			0°C		0.04		
			70°C		0.03		
		$V_{I(PP)} = 2.5\text{ V}$	25°C		0.03		
			0°C		0.03		
			70°C		0.02		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 20\text{ }\Omega$,	25°C		68		nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 1\text{ M}\Omega$,	$C_L = 20\text{ pF}$, See Figure 98	25°C		5		kHz
			0°C		6		
			70°C		4.5		
B_1 Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$,	25°C		85		kHz
			0°C		100		
			70°C		65		
ϕ_m Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 100	25°C		34°		
			0°C		36°		
			70°C		30°		

operating characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS		T_A	TLC271C, TLC271AC, TLC271BC			UNIT
				MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C		0.05		V/ μ s
			0°C		0.05		
			70°C		0.04		
		$V_{I(PP)} = 5.5\text{ V}$	25°C		0.04		
			0°C		0.05		
			70°C		0.04		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 20\text{ }\Omega$,	25°C		68		nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 1\text{ M}\Omega$,	$C_L = 20\text{ pF}$, See Figure 98	25°C		1		kHz
			0°C		1.3		
			70°C		0.9		
B_1 Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$,	25°C		110		kHz
			0°C		125		
			70°C		90		
ϕ_m Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$,	$f = B_1$, See Figure 100	25°C		38°		
			0°C		40°		
			70°C		34°		

TLC271, TLC271A, TLC271B
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

LOW-BIAS MODE

operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS	T_A	TLC271I, TLC271AI, TLC271BI			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C	0.03		V/ μs
			–40°C	0.04		
			85°C	0.03		
		$V_{I(PP)} = 2.5\text{ V}$	25°C	0.03		
			–40°C	0.04		
			85°C	0.02		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 20\text{ }\Omega$, 25°C		68		nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	25°C		5		kHz
		–40°C		7		
		85°C		4		
B_1 Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$, 25°C		85		MHz
		–40°C		130		
		85°C		55		
ϕ_m Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$, $f = B_1$, See Figure 100	25°C		34°		
		–40°C		38°		
		85°C		28°		

operating characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS	T_A	TLC271C, TLC271AC, TLC271BC			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C	0.05		V/ μs
			–40°C	0.06		
			85°C	0.03		
		$V_{I(PP)} = 5.5\text{ V}$	25°C	0.04		
			–40°C	0.05		
			85°C	0.03		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 20\text{ }\Omega$, 25°C		68		nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	25°C		1		kHz
		–40°C		1.4		
		85°C		0.8		
B_1 Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$, 25°C		110		MHz
		–40°C		155		
		85°C		80		
ϕ_m Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$, $f = B_1$, See Figure 100	25°C		38°		
		–40°C		42°		
		85°C		32°		



LOW-BIAS MODE

operating characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS	T_A	TLC271M			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C	0.03		V/ μ s
			–55°C	0.04		
			125°C	0.02		
		$V_{I(PP)} = 2.5\text{ V}$	25°C	0.03		
			–55°C	0.04		
			125°C	0.02		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 20\text{ }\Omega$, 25°C		68		nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	25°C		5		kHz
		–55°C		8		
		125°C		3		
B_1 Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$, 25°C		85		kHz
		–55°C		140		
		125°C		45		
ϕ_m Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$, $f = B_1$, See Figure 100	25°C		34°		
		–55°C		39°		
		125°C		25°		

operating characteristics at specified free-air temperature, $V_{DD} = 10\text{ V}$

PARAMETER	TEST CONDITIONS	T_A	TLC271M			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	$V_{I(PP)} = 1\text{ V}$	25°C	0.05		V/ μ s
			–55°C	0.06		
			125°C	0.03		
		$V_{I(PP)} = 5.5\text{ V}$	25°C	0.04		
			–55°C	0.06		
			125°C	0.03		
V_n Equivalent input noise voltage	$f = 1\text{ kHz}$, See Figure 99	$R_S = 20\text{ }\Omega$, 25°C		68		nV/ $\sqrt{\text{Hz}}$
B_{OM} Maximum output-swing bandwidth	$V_O = V_{OH}$, $R_L = 1\text{ M}\Omega$, $C_L = 20\text{ pF}$, See Figure 98	25°C		1		kHz
		–55°C		1.5		
		125°C		0.7		
B_1 Unity-gain bandwidth	$V_I = 10\text{ mV}$, See Figure 100	$C_L = 20\text{ pF}$, 25°C		110		kHz
		–55°C		165		
		125°C		70		
ϕ_m Phase margin	$V_I = 10\text{ mV}$, $C_L = 20\text{ pF}$, $f = B_1$, See Figure 100	25°C		38°		
		–55°C		43°		
		125°C		29°		

TLC271, TLC271A, TLC271B
LinCMOS™ PROGRAMMABLE LOW-POWER
OPERATIONAL AMPLIFIERS

SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

TYPICAL CHARACTERISTICS (LOW-BIAS MODE)

Table of Graphs

			FIGURE
V_{IO}	Input offset voltage	Distribution	66, 67
α_{VIO}	Temperature coefficient	Distribution	68, 69
V_{OH}	High-level output voltage	vs High-level output current vs Supply voltage vs Free-air temperature	70, 71 72 73
V_{OL}	Low-level output voltage	vs Common-mode input voltage vs Differential input voltage vs Free-air temperature vs Low-level output current	74, 75 76 77 78, 79
A_{VD}	Large-signal differential voltage amplification	vs Supply voltage vs Free-air temperature vs Frequency	80 81 92, 93
I_{IB}	Input bias current	vs Free-air temperature	82
I_{IO}	Input offset current	vs Free-air temperature	82
V_I	Maximum input voltage	vs Supply voltage	83
I_{DD}	Supply current	vs Supply voltage vs Free-air temperature	84 85
SR	Slew rate	vs Supply voltage vs Free-air temperature	86 87
	Bias-select current	vs Supply voltage	88
$V_{O(PP)}$	Maximum peak-to-peak output voltage	vs Frequency	89
B_1	Unity-gain bandwidth	vs Free-air temperature vs Supply voltage	90 91
ϕ_m	Phase margin	vs Supply voltage vs Free-air temperature vs Capacitive load	94 95 96
V_n	Equivalent input noise voltage	vs Frequency	97
	Phase shift	vs Frequency	92, 93



TYPICAL CHARACTERISTICS (LOW-BIAS MODE)[†]

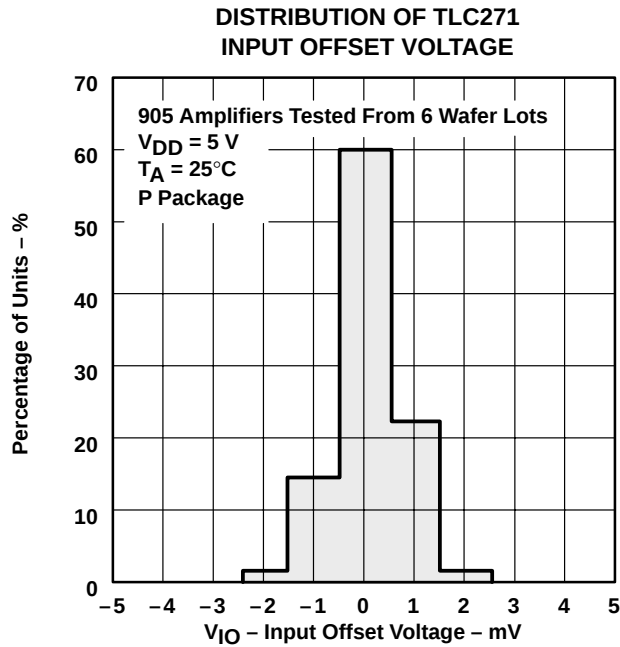


Figure 66

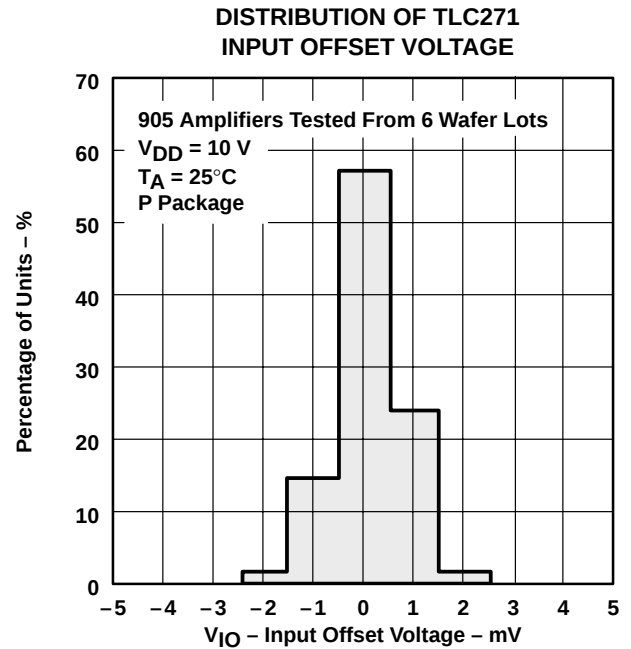


Figure 67

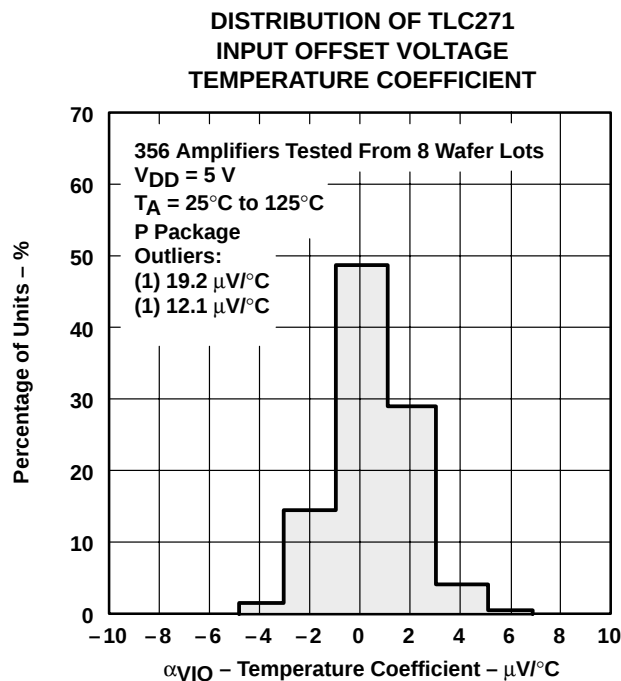


Figure 68

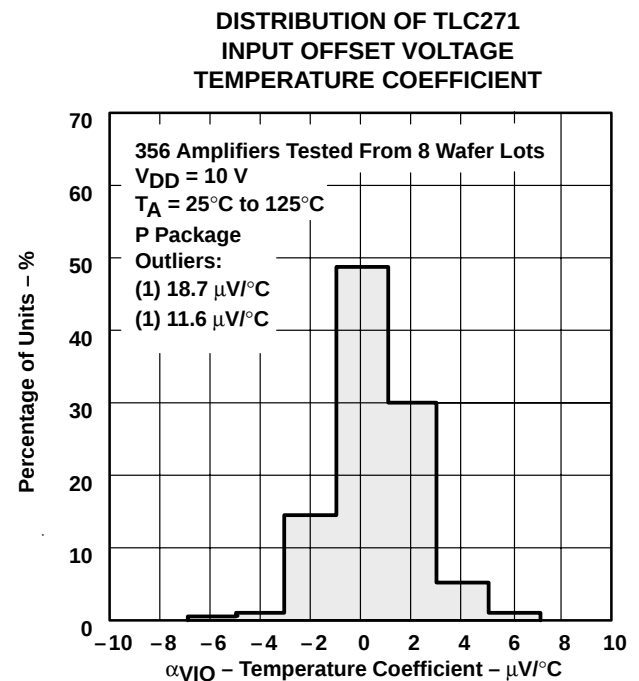


Figure 69

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC271, TLC271A, TLC271B

LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

TYPICAL CHARACTERISTICS (LOW-BIAS MODE)[†]

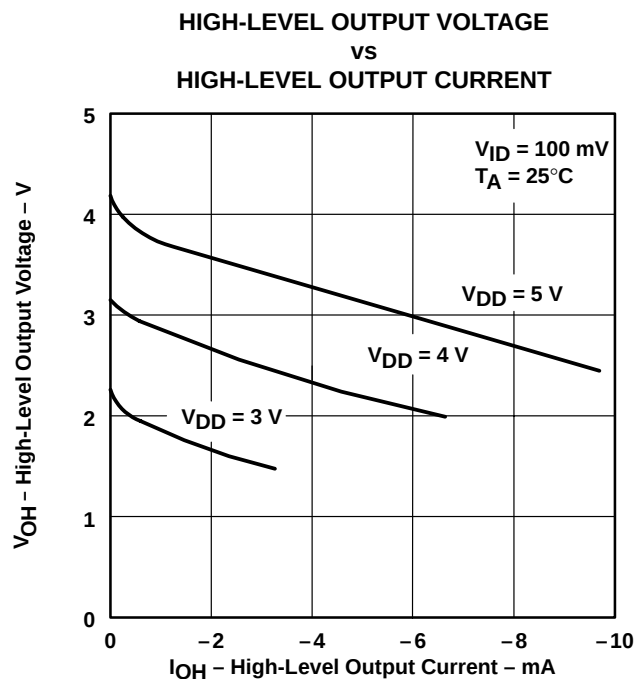


Figure 70

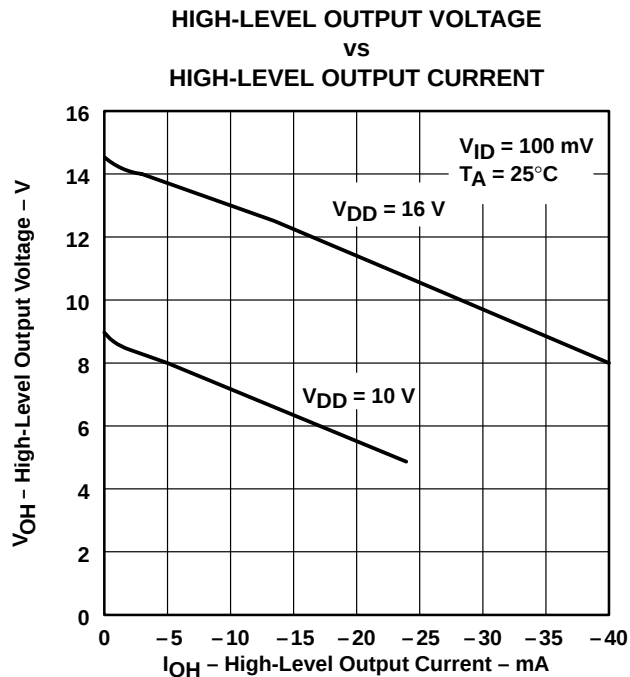


Figure 71

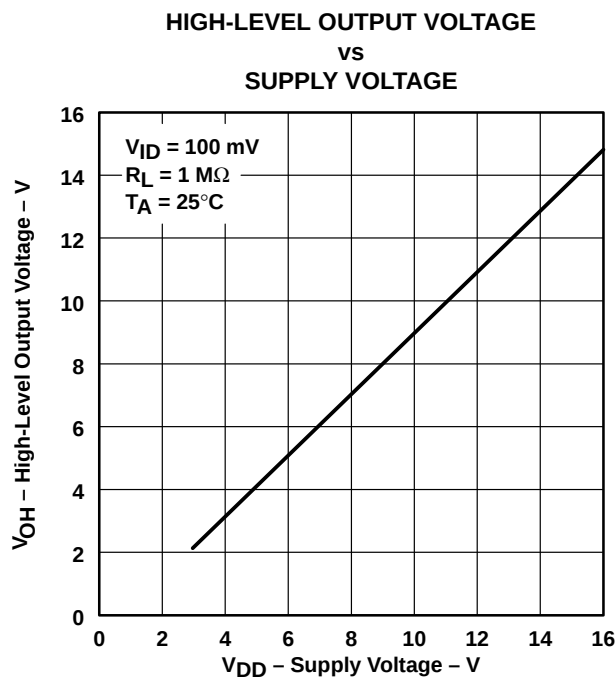


Figure 72

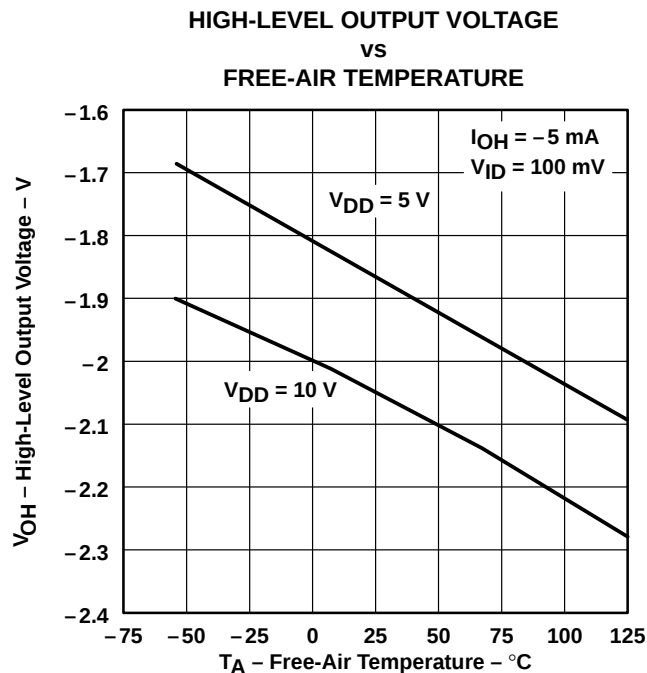


Figure 73

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS (LOW-BIAS MODE)[†]

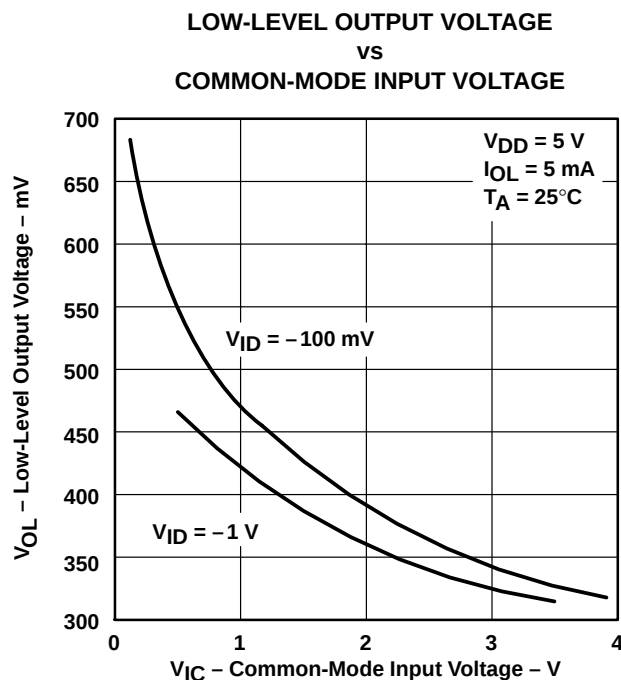


Figure 74

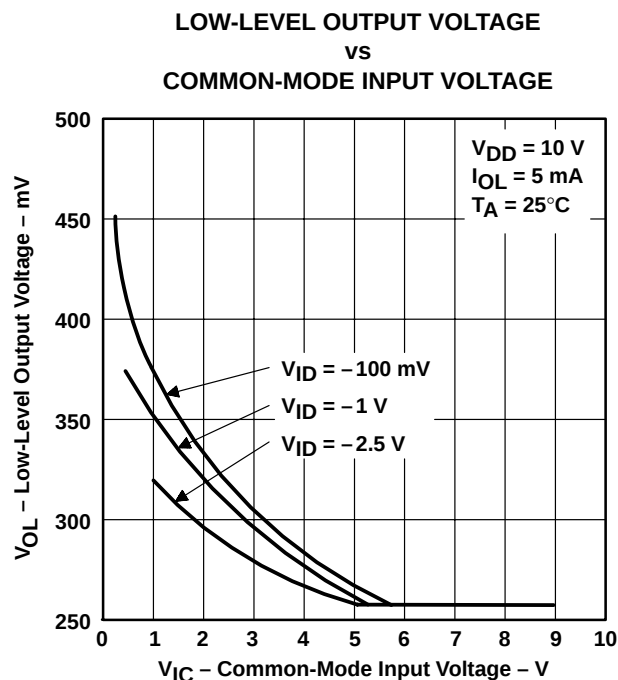


Figure 75

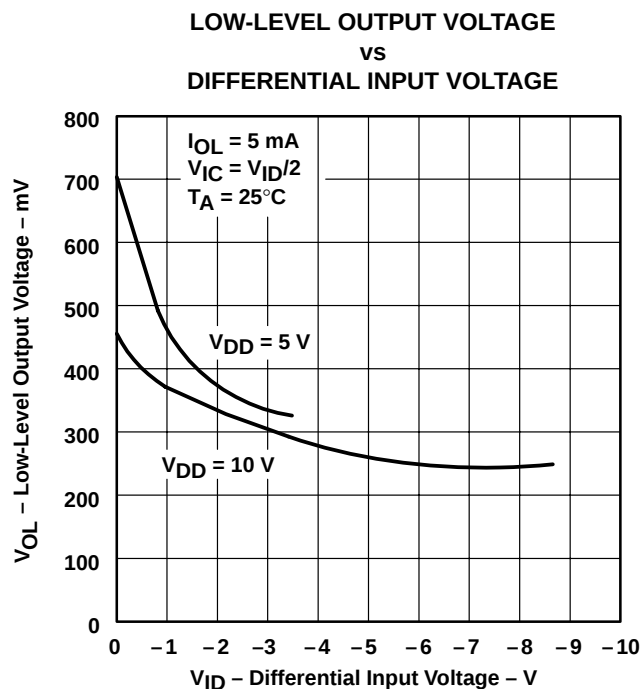


Figure 76

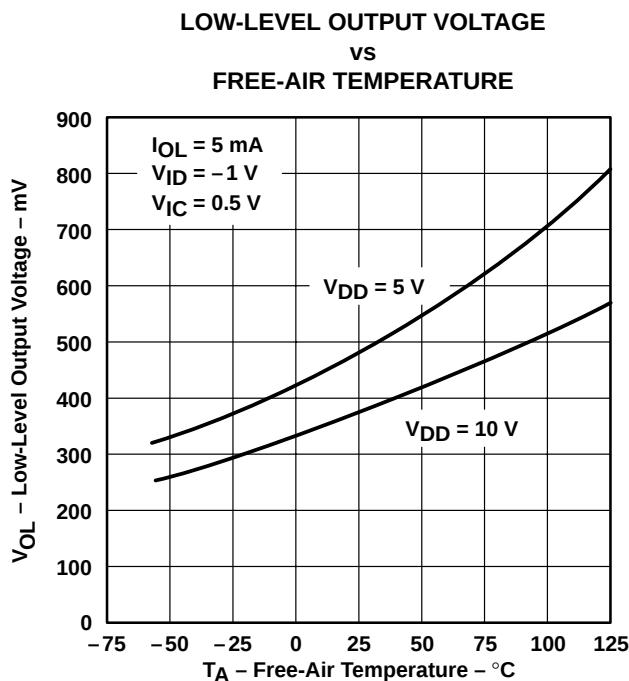


Figure 77

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC271, TLC271A, TLC271B

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SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

TYPICAL CHARACTERISTICS (LOW-BIAS MODE)[†]

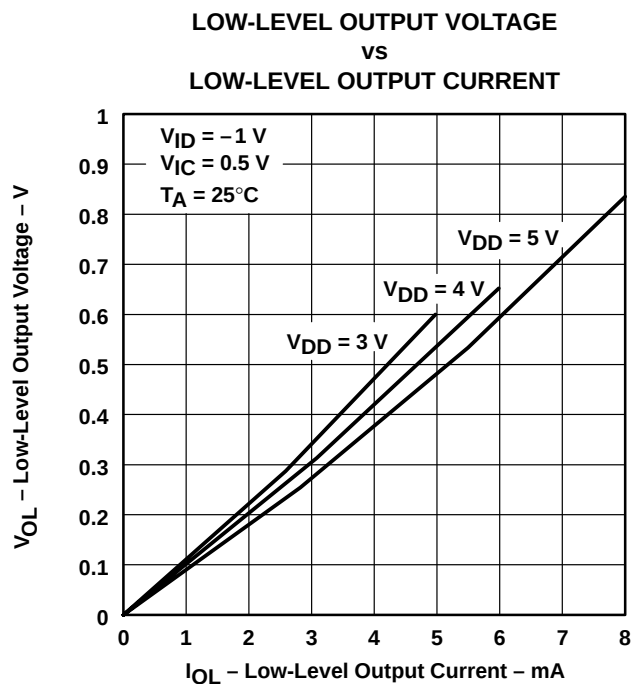


Figure 78

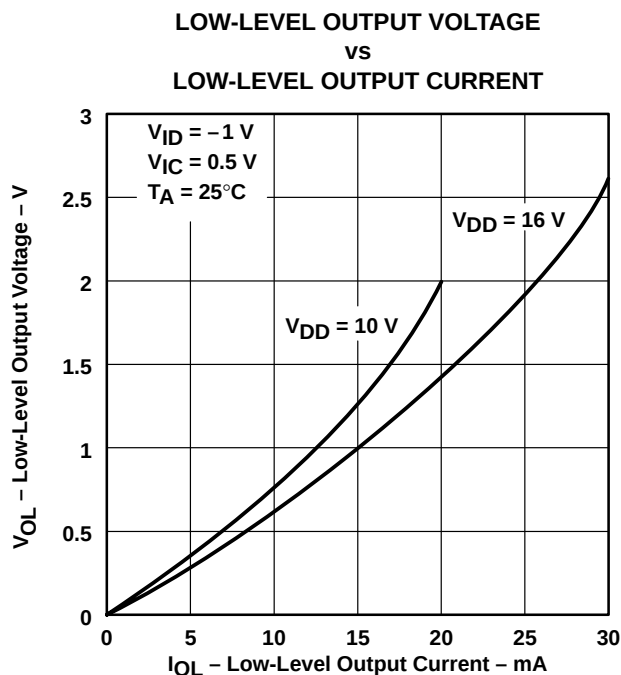


Figure 79

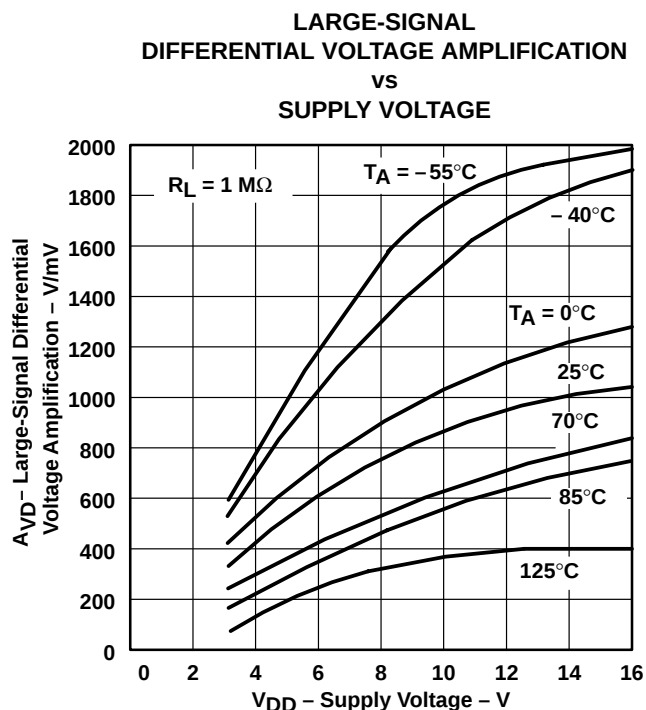


Figure 80

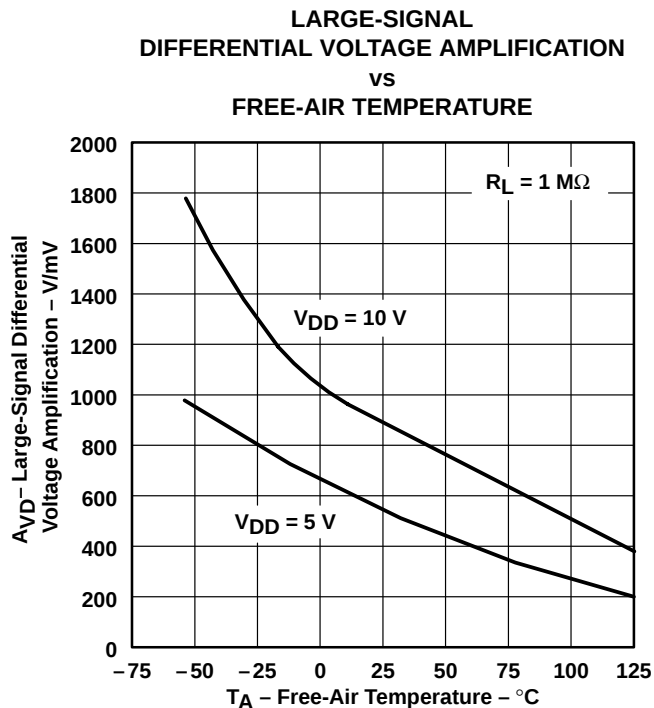
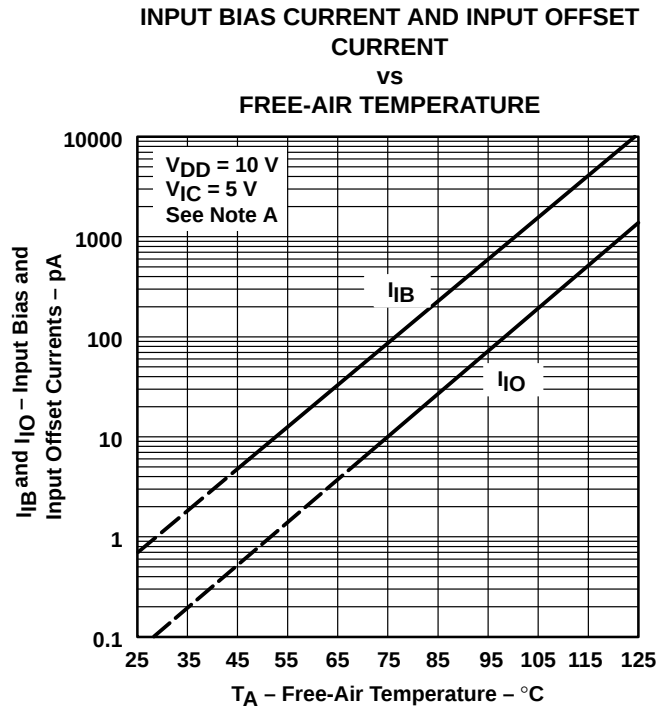


Figure 81

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS (LOW-BIAS MODE)[†]



NOTE A: The typical values of input bias current and input offset current below 5 pA were determined mathematically.

Figure 82

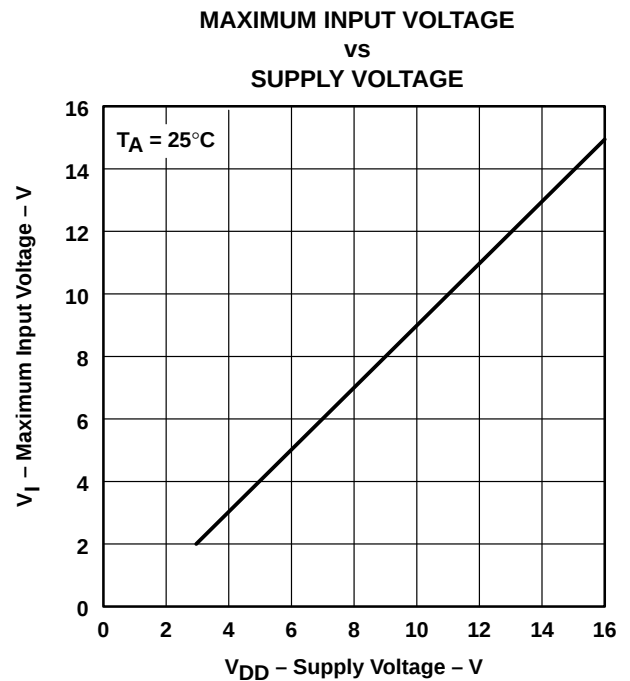


Figure 83

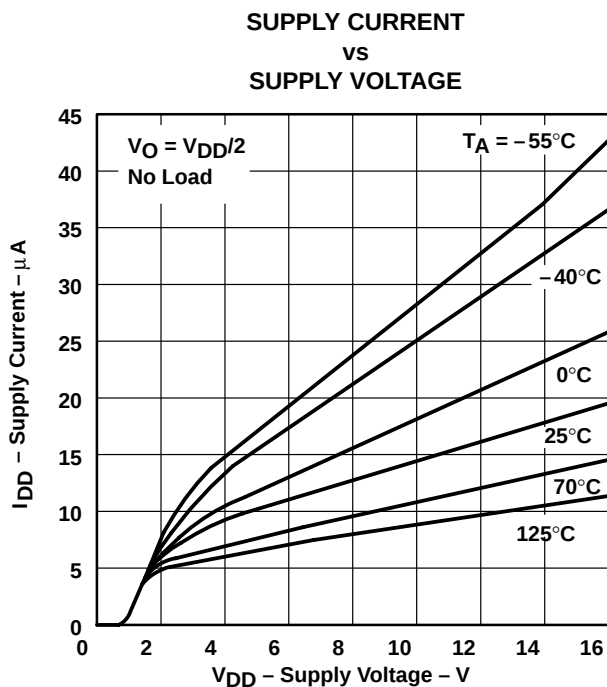


Figure 84

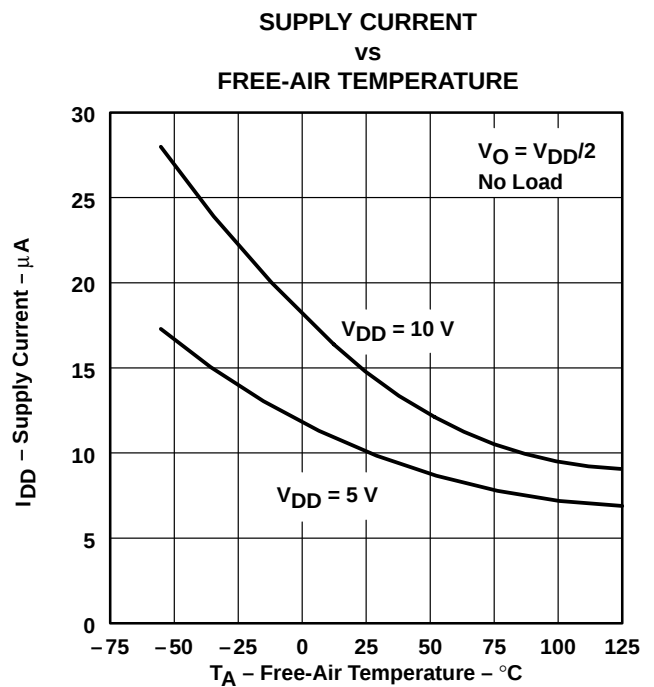


Figure 85

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC271, TLC271A, TLC271B

LinCMOS™ PROGRAMMABLE LOW-POWER OPERATIONAL AMPLIFIERS

SLOS090D – NOVEMBER 1987 – REVISED MARCH 2001

TYPICAL CHARACTERISTICS (LOW-BIAS MODE)[†]

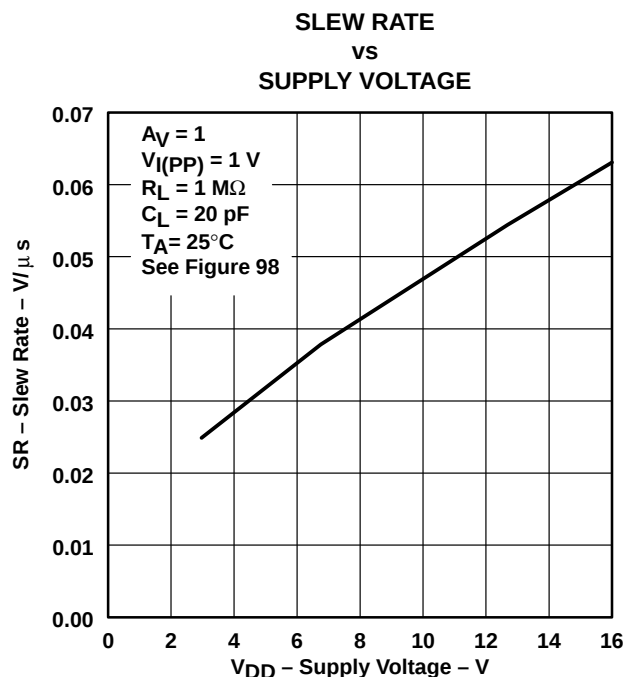


Figure 86

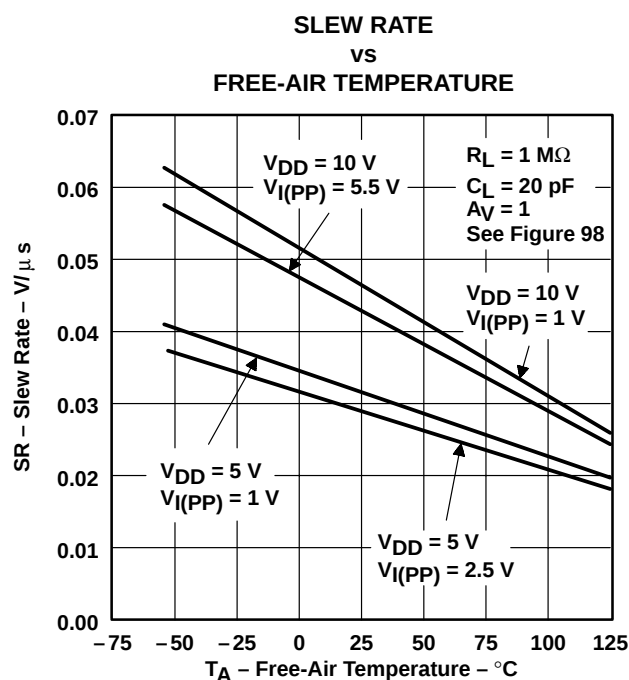


Figure 87

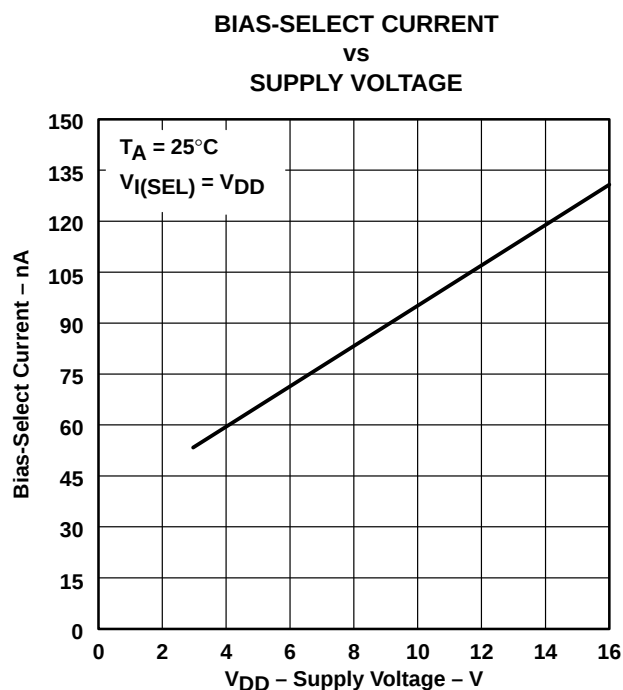


Figure 88

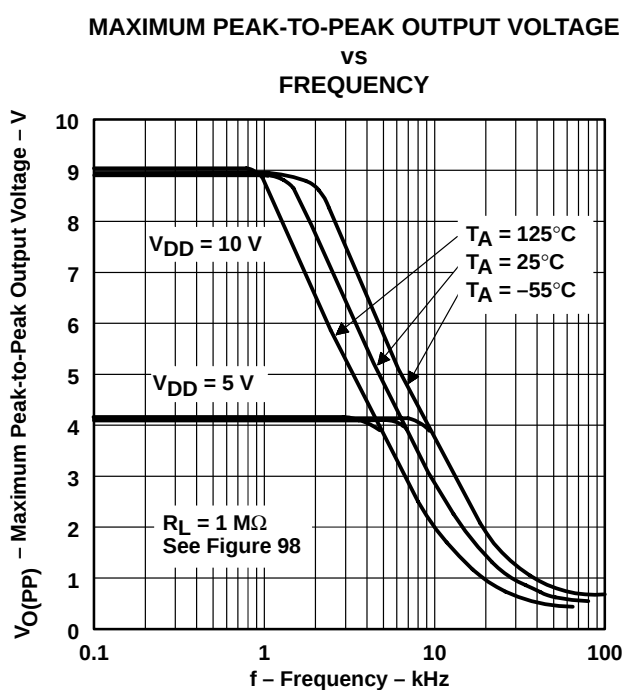


Figure 89

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS (LOW-BIAS MODE)[†]

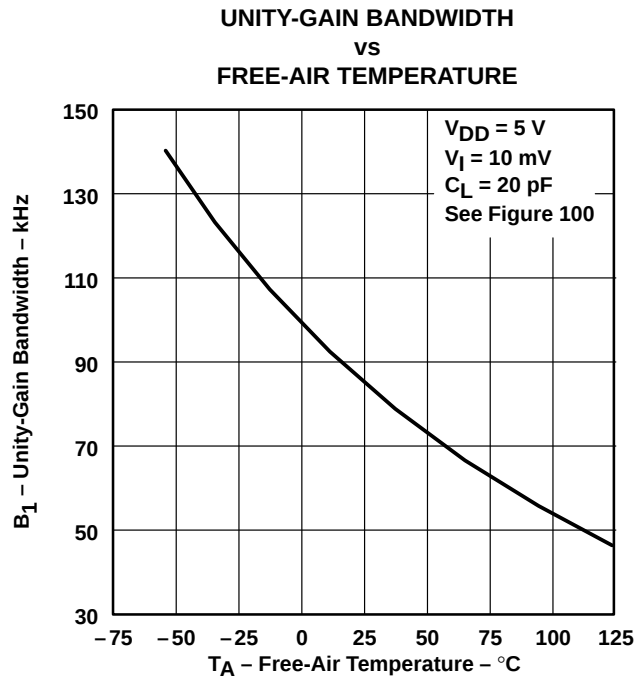


Figure 90

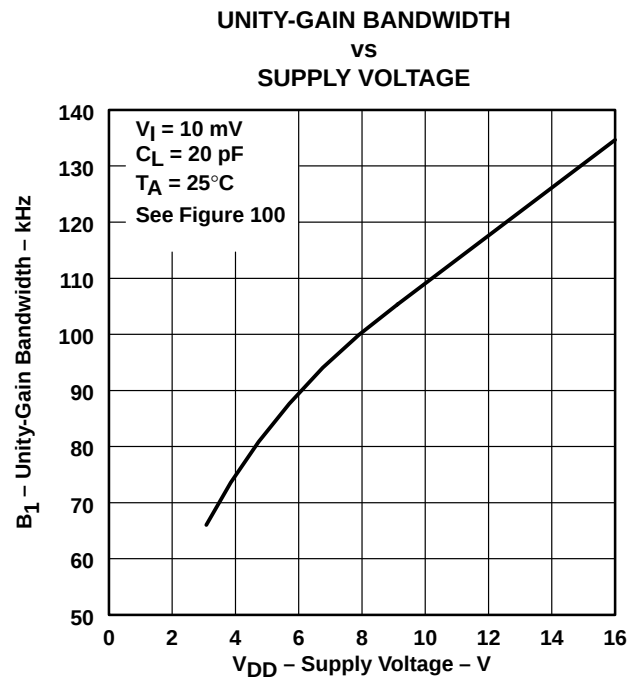


Figure 91

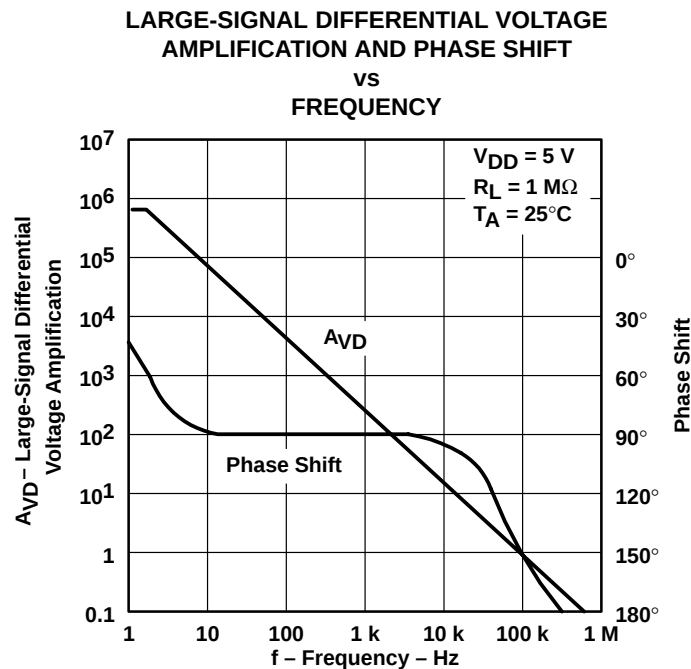


Figure 92

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TLC271, TLC271A, TLC271B

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TYPICAL CHARACTERISTICS (LOW-BIAS MODE)[†]

LARGE-SIGNAL DIFFERENTIAL VOLTAGE AMPLIFICATION AND PHASE SHIFT

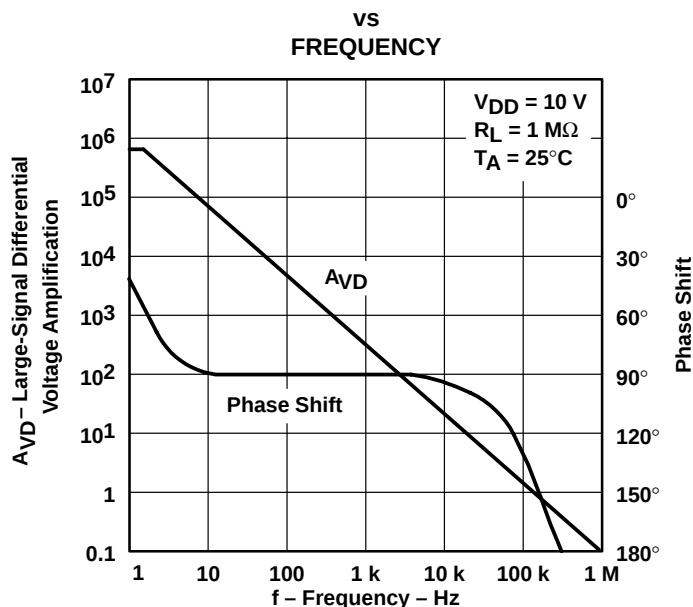


Figure 93

PHASE MARGIN VS SUPPLY VOLTAGE

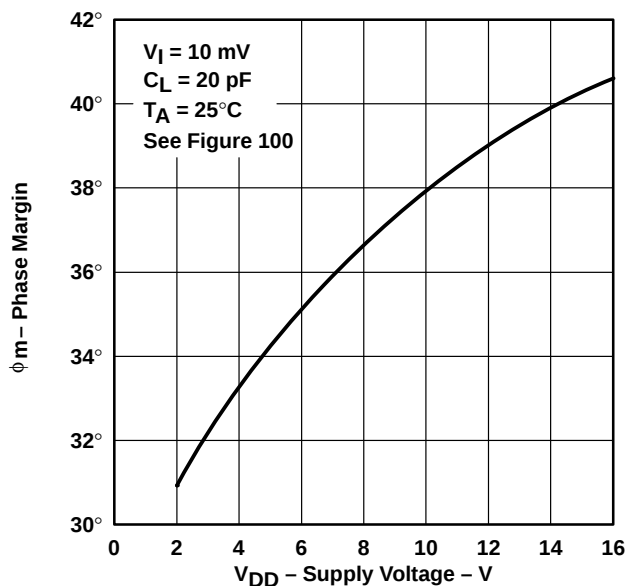


Figure 94

PHASE MARGIN VS FREE-AIR TEMPERATURE

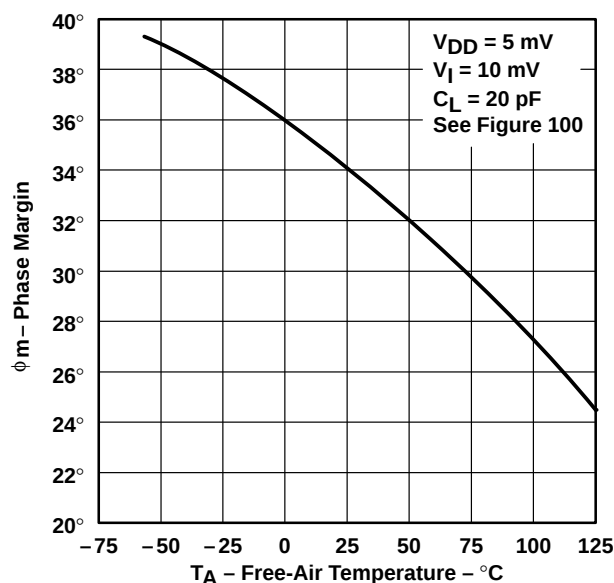


Figure 95

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

TYPICAL CHARACTERISTICS (LOW-BIAS MODE)[†]

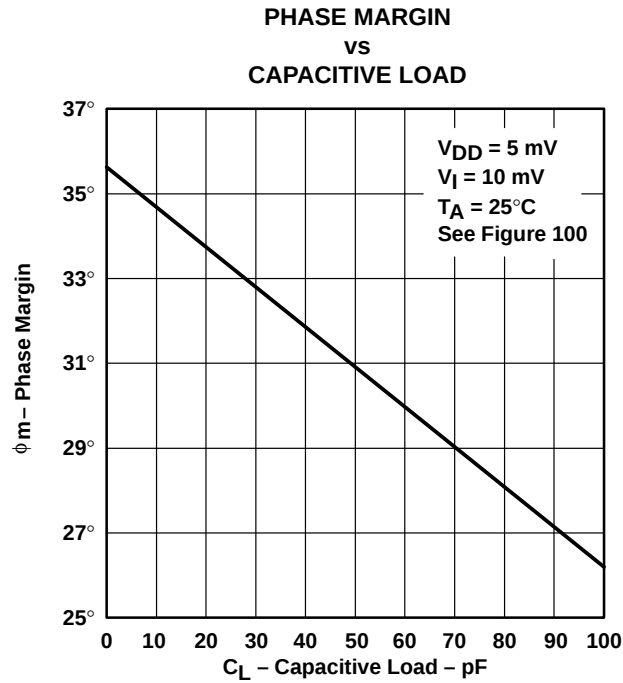


Figure 96

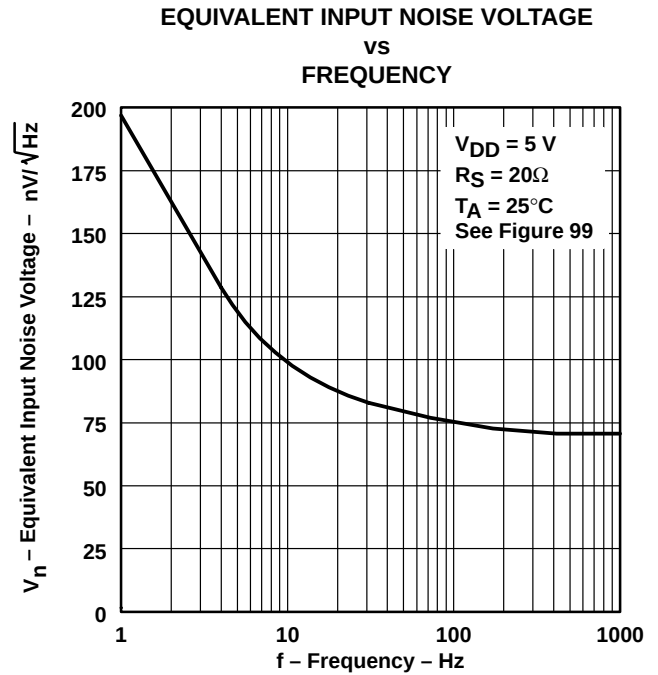


Figure 97

[†] Data at high and low temperatures are applicable only within the rated operating free-air temperature ranges of the various devices.

PARAMETER MEASUREMENT INFORMATION

single-supply versus split-supply test circuits

Because the TLC271 is optimized for single-supply operation, circuit configurations used for the various tests often present some inconvenience since the input signal, in many cases, must be offset from ground. This inconvenience can be avoided by testing the device with split supplies and the output load tied to the negative rail. A comparison of single-supply versus split-supply test circuits is shown below. The use of either circuit gives the same result.

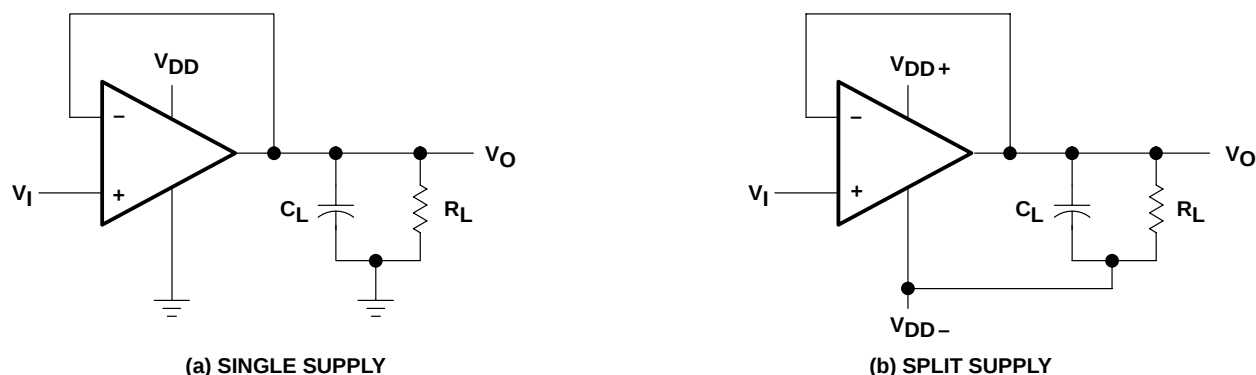


Figure 98. Unity-Gain Amplifier

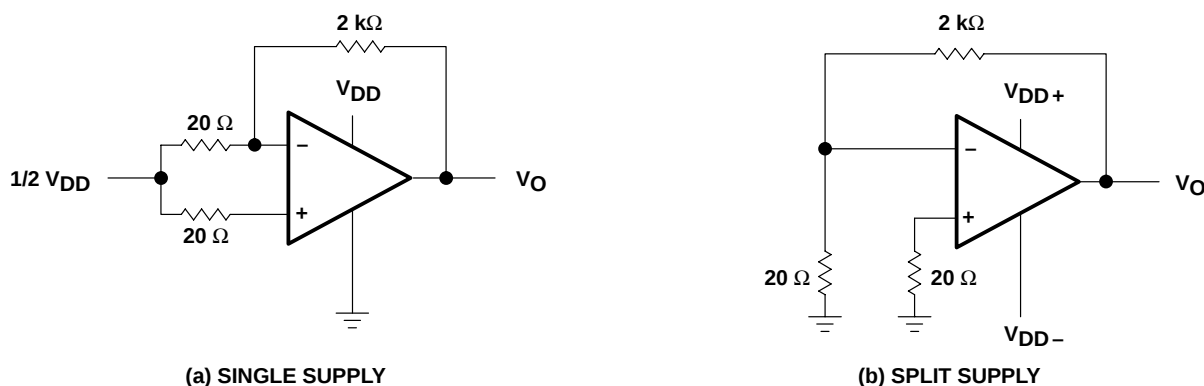


Figure 99. Noise-Test Circuit

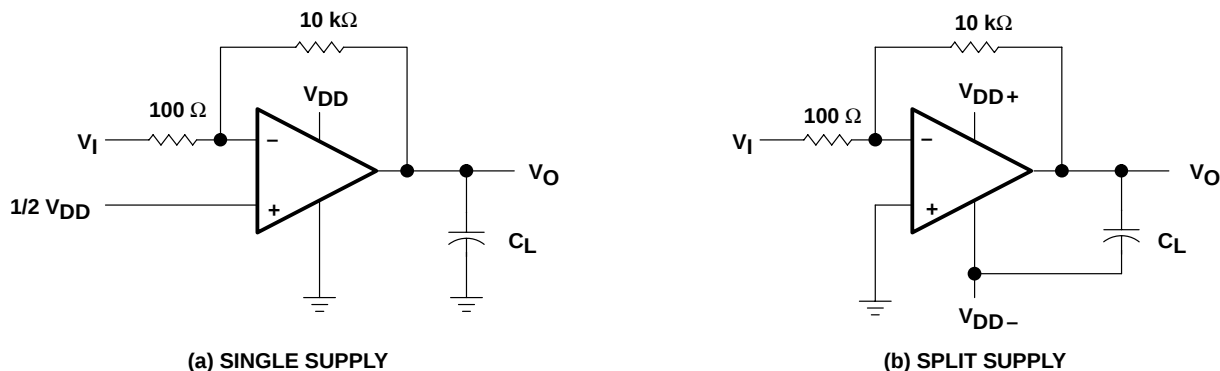


Figure 100. Gain-of-100 Inverting Amplifier

PARAMETER MEASUREMENT INFORMATION

input bias current

Because of the high input impedance of the TLC271 operational amplifiers, attempts to measure the input bias current can result in erroneous readings. The bias current at normal room ambient temperature is typically less than 1 pA, a value that is easily exceeded by leakages on the test socket. Two suggestions are offered to avoid erroneous measurements:

1. Isolate the device from other potential leakage sources. Use a grounded shield around and between the device inputs (see Figure 101). Leakages that would otherwise flow to the inputs are shunted away.
2. Compensate for the leakage of the test socket by actually performing an input bias current test (using a picoammeter) with no device in the test socket. The actual input bias current can then be calculated by subtracting the open-socket leakage readings from the readings obtained with a device in the test socket.

One word of caution: many automatic testers as well as some bench-top operational amplifier testers use the servo-loop technique with a resistor in series with the device input to measure the input bias current (the voltage drop across the series resistor is measured and the bias current is calculated). This method requires that a device be inserted into the test socket to obtain a correct reading; therefore, an open-socket reading is not feasible using this method.

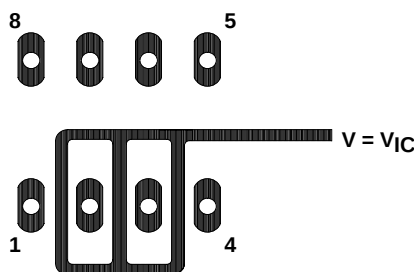


Figure 101. Isolation Metal Around Device inputs (JG and P packages)

low-level output voltage

To obtain low-supply-voltage operation, some compromise is necessary in the input stage. This compromise results in the device low-level output being dependent on both the common-mode input voltage level as well as the differential input voltage level. When attempting to correlate low-level output readings with those quoted in the electrical specifications, these two conditions should be observed. If conditions other than these are to be used, please refer to the Typical Characteristics section of this data sheet.

input offset voltage temperature coefficient

Erroneous readings often result from attempts to measure temperature coefficient of input offset voltage. This parameter is actually a calculation using input offset voltage measurements obtained at two different temperatures. When one (or both) of the temperatures is below freezing, moisture can collect on both the device and the test socket. This moisture results in leakage and contact resistance which can cause erroneous input offset voltage readings. The isolation techniques previously mentioned have no effect on the leakage since the moisture also covers the isolation metal itself, thereby rendering it useless. It is suggested that these measurements be performed at temperatures above freezing to minimize error.

PARAMETER MEASUREMENT INFORMATION

full-power response

Full-power response, the frequency above which the amplifier slew rate limits the output voltage swing, is often specified two ways: full-linear response and full-peak response. The full-linear response is generally measured by monitoring the distortion level of the output while increasing the frequency of a sinusoidal input signal until the maximum frequency is found above which the output contains significant distortion. The full-peak response is defined as the maximum output frequency, without regard to distortion, above which full peak-to-peak output swing cannot be maintained.

Because there is no industry-wide accepted value for significant distortion, the full-peak response is specified in this data sheet and is measured using the circuit of Figure 98. The initial setup involves the use of a sinusoidal input to determine the maximum peak-to-peak output of the device (the amplitude of the sinusoidal wave is increased until clipping occurs). The sinusoidal wave is then replaced with a square wave of the same amplitude. The frequency is then increased until the maximum peak-to-peak output can no longer be maintained (Figure 102). A square wave is used to allow a more accurate determination of the point at which the maximum peak-to-peak output is reached.

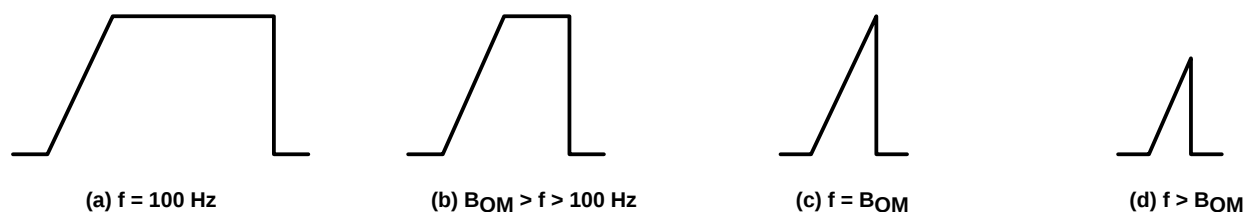


Figure 102. Full-Power-Response Output Signal

test time

Inadequate test time is a frequent problem, especially when testing CMOS devices in a high-volume, short-test-time environment. Internal capacitances are inherently higher in CMOS than in bipolar and BiFET devices, and require longer test times than their bipolar and BiFET counterparts. The problem becomes more pronounced with reduced supply levels and lower temperatures.

APPLICATION INFORMATION

single-supply operation

While the TLC271 performs well using dual power supplies (also called balanced or split supplies), the design is optimized for single-supply operation. This includes an input common mode voltage range that encompasses ground as well as an output voltage range that pulls down to ground. The supply voltage range extends down to 3 V (C-suffix types), thus allowing operation with supply levels commonly available for TTL and HCMOS; however, for maximum dynamic range, 16-V single-supply operation is recommended.

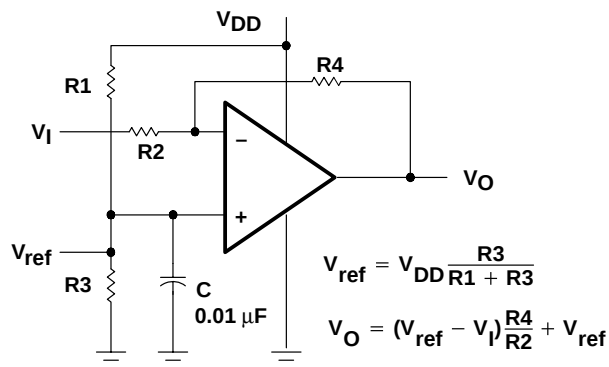


Figure 103. Inverting Amplifier With Voltage Reference

APPLICATION INFORMATION

single-supply operation (continued)

Many single-supply applications require that a voltage be applied to one input to establish a reference level that is above ground. A resistive voltage divider is usually sufficient to establish this reference level (see Figure 103). The low input bias current consumption of the TLC271 permits the use of very large resistive values to implement the voltage divider, thus minimizing power consumption.

The TLC271 works well in conjunction with digital logic; however, when powering both linear devices and digital logic from the same power supply, the following precautions are recommended:

1. Power the linear devices from separate bypassed supply lines (see Figure 104); otherwise, the linear device supply rails can fluctuate due to voltage drops caused by high switching currents in the digital logic.
2. Use proper bypass techniques to reduce the probability of noise-induced errors. Single capacitive decoupling is often adequate; however, RC decoupling may be necessary in high-frequency applications.

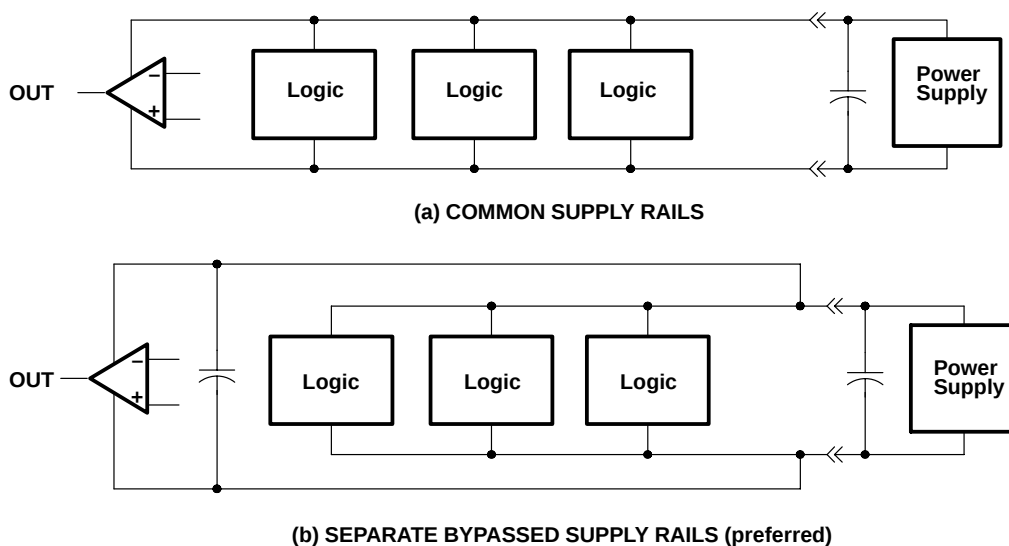


Figure 104. Common Versus Separate Supply Rails

APPLICATION INFORMATION

input offset voltage nulling

The TLC271 offers external input offset null control. Nulling of the input offset voltage may be achieved by adjusting a 25-k Ω potentiometer connected between the offset null terminals with the wiper connected as shown in Figure 105. The amount of nulling range varies with the bias selection. In the high-bias mode, the nulling range allows the maximum offset voltage specified to be trimmed to zero. In low-bias and medium-bias modes, total nulling may not be possible.

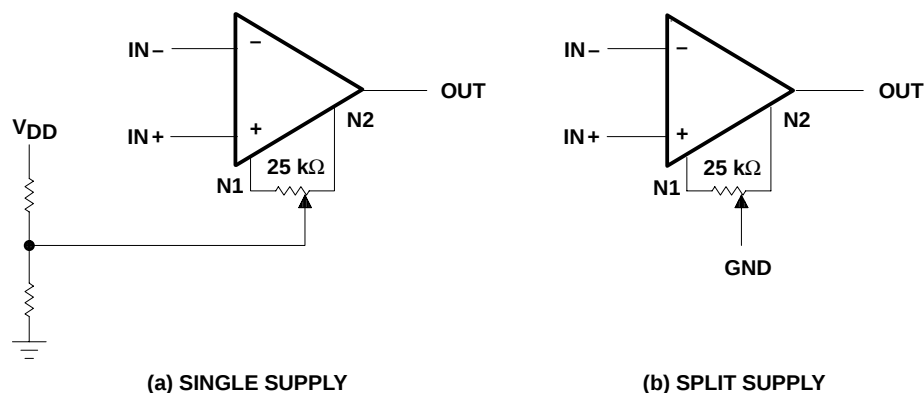
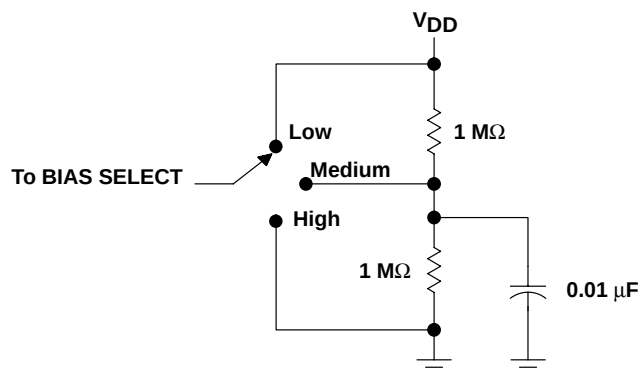


Figure 105. Input Offset Voltage Null Circuit

bias selection

Bias selection is achieved by connecting the bias select pin to one of the three voltage levels (see Figure 106). For medium-bias applications, R is recommended that the bias select pin be connected to the mid-point between the supply rails. This is a simple procedure in split-supply applications, since this point is ground. In single-supply applications, the medium-bias mode necessitates using a voltage divider as indicated. The use of large-value resistors in the voltage divider reduces the current drain of the divider from the supply line. However, large-value resistors used in conjunction with a large-value capacitor requires significant time to charge up to the supply midpoint after the supply is switched on. A voltage other than the midpoint may be used if it is within the voltages specified in the table of Figure 106.



BIAS MODE	BIAS-SELECT VOLTAGE (single supply)
Low	V_{DD}
Medium	1 V to $V_{DD} - 1$ V
High	GND

Figure 106. Bias Selection for Single-Supply Applications

APPLICATION INFORMATION

input characteristics

The TLC271 is specified with a minimum and a maximum input voltage that, if exceeded at either input, could cause the device to malfunction. Exceeding this specified range is a common problem, especially in single-supply operation. Note that the lower range limit includes the negative rail, while the upper range limit is specified at $V_{DD} - 1$ V at $T_A = 25^\circ\text{C}$ and at $V_{DD} - 1.5$ V at all other temperatures.

The use of the polysilicon-gate process and the careful input circuit design gives the TLC271 very good input offset voltage drift characteristics relative to conventional metal-gate processes. Offset voltage drift in CMOS devices is highly influenced by threshold voltage shifts caused by polarization of the phosphorus dopant implanted in the oxide. Placing the phosphorus dopant in a conductor (such as a polysilicon gate) alleviates the polarization problem, thus reducing threshold voltage shifts by more than an order of magnitude. The offset voltage drift with time has been calculated to be typically $0.1 \mu\text{V/month}$, including the first month of operation.

Because of the extremely high input impedance and resulting low bias current requirements, the TLC271 is well suited for low-level signal processing; however, leakage currents on printed circuit boards and sockets can easily exceed bias current requirements and cause a degradation in device performance. It is good practice to include guard rings around inputs (similar to those of Figure 101 in the Parameter Measurement Information section). These guards should be driven from a low-impedance source at the same voltage level as the common-mode input (see Figure 107).

The inputs of any unused amplifiers should be tied to ground to avoid possible oscillation.

noise performance

The noise specifications in operational amplifier circuits are greatly dependent on the current in the first-stage differential amplifier. The low input bias current requirements of the TLC271 results in a very low noise current, which is insignificant in most applications. This feature makes the devices especially favorable over bipolar devices when using values of circuit impedance greater than $50 \text{ k}\Omega$, since bipolar devices exhibit greater noise currents.

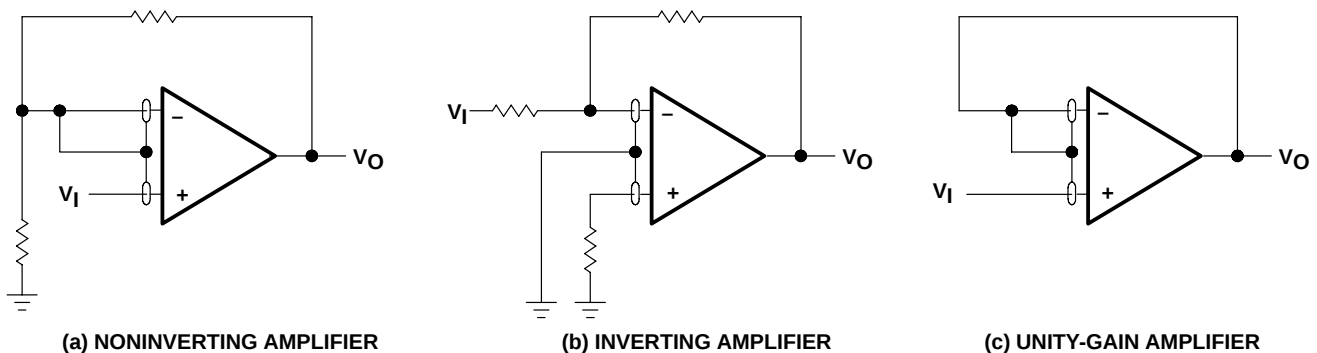


Figure 107. Guard-Ring Schemes

APPLICATION INFORMATION

feedback

Operational amplifier circuits almost always employ feedback, and since feedback is the first prerequisite for oscillation, a little caution is appropriate. Most oscillation problems result from driving capacitive loads and ignoring stray input capacitance. A small-value capacitor connected in parallel with the feedback resistor is an effective remedy (see Figure 108). The value of this capacitor is optimized empirically.

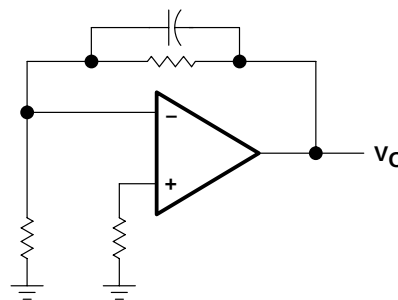


Figure 108. Compensation for Input Capacitance

electrostatic discharge protection

The TLC271 incorporates an internal electrostatic-discharge (ESD) protection circuit that prevents functional failures at voltages up to 2000 V as tested under MIL-STD-883C, Method 3015.2. Care should be exercised, however, when handling these devices as exposure to ESD may result in the degradation of the device parametric performance. The protection circuit also causes the input bias currents to be temperature dependent and have the characteristics of a reverse-biased diode.

latch-up

Because CMOS devices are susceptible to latch-up due to their inherent parasitic thyristors, the TLC271 inputs and output were designed to withstand -100-mA surge currents without sustaining latchup; however, techniques should be used to reduce the chance of latch-up whenever possible. Internal protection diodes should not by design be forward biased. Applied input and output voltage should not exceed the supply voltage by more than 300 mV. Care should be exercised when using capacitive coupling on pulse generators. Supply transients should be shunted by the use of decoupling capacitors ($0.1\text{ }\mu\text{F}$ typical) located across the supply rails as close to the device as possible.

The current path established if latch-up occurs is usually between the positive supply rail and ground and can be triggered by surges on the supply lines and/or voltages on either the output or inputs that exceed the supply voltage. Once latch-up occurs, the current flow is limited only by the impedance of the power supply and the forward resistance of the parasitic thyristor and usually results in the destruction of the device. The chance of latch-up occurring increases with increasing temperature and supply voltages.

output characteristics

The output stage of the TLC271 is designed to sink and source relatively high amounts of current (see Typical Characteristics). If the output is subjected to a short-circuit condition, this high current capability can cause device damage under certain conditions. Output current capability increases with supply voltage.

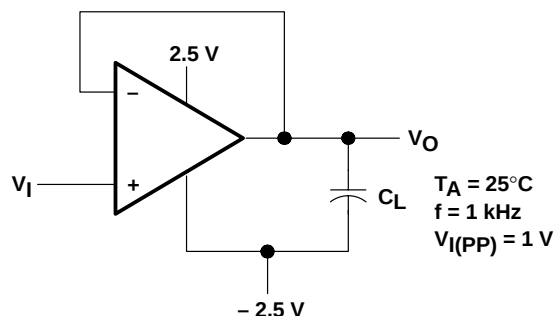


Figure 109. Test Circuit for Output Characteristics

APPLICATION INFORMATION

output characteristics (continued)

All operating characteristics of the TLC271 were measured using a 20-pF load. The devices drive higher capacitive loads; however, as output load capacitance increases, the resulting response pole occurs at lower frequencies, thereby causing ringing, peaking, or even oscillation (see Figures 110, 111, and 112). In many cases, adding some compensation in the form of a series resistor in the feedback loop alleviates the problem.

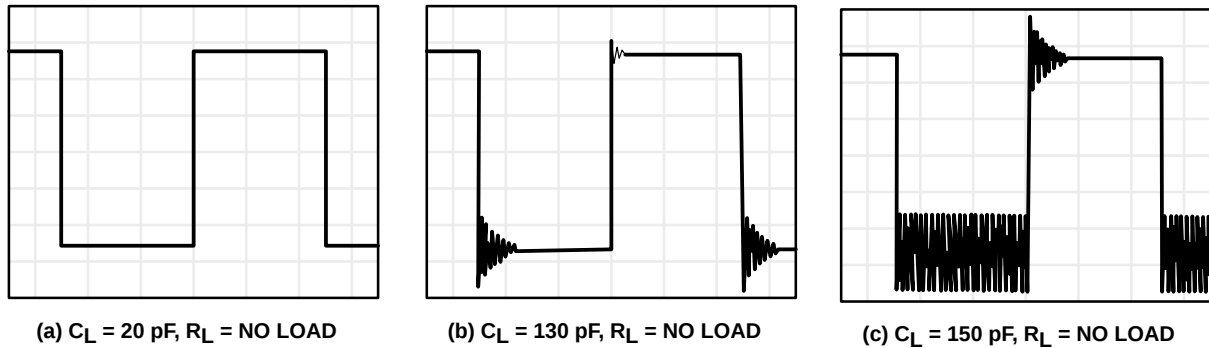


Figure 110. Effect of Capacitive Loads in High-Bias Mode

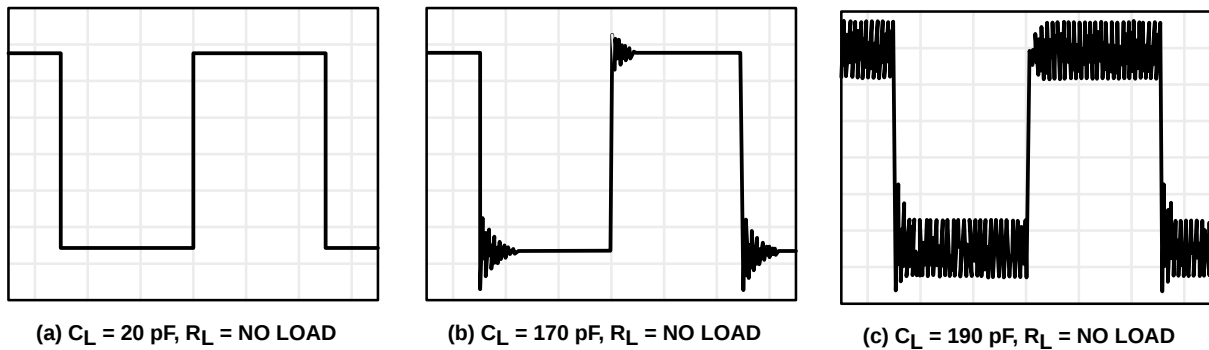


Figure 111. Effect of Capacitive Loads in Medium-Bias Mode



Figure 112. Effect of Capacitive Loads in Low-Bias Mode

APPLICATION INFORMATION

output characteristics (continued)

Although the TLC271 possesses excellent high-level output voltage and current capability, methods are available for boosting this capability, if needed. The simplest method involves the use of a pullup resistor (R_P) connected from the output to the positive supply rail (see Figure 113). There are two disadvantages to the use of this circuit. First, the NMOS pulldown transistor, N4 (see equivalent schematic) must sink a comparatively large amount of current. In this circuit, N4 behaves like a linear resistor with an on-resistance between approximately 60 Ω and 180 Ω , depending on how hard the operational amplifier input is driven. With very low values of R_P , a voltage offset from 0 V at the output occurs. Secondly, pullup resistor R_P acts as a drain load to N4 and the gain of the operational amplifier is reduced at output voltage levels where N5 is not supplying the output current.

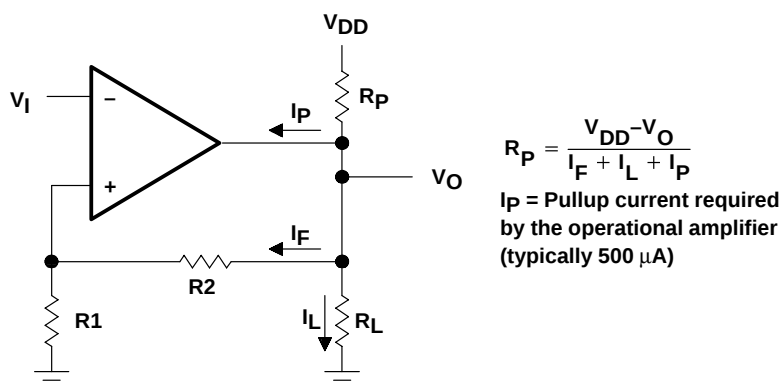
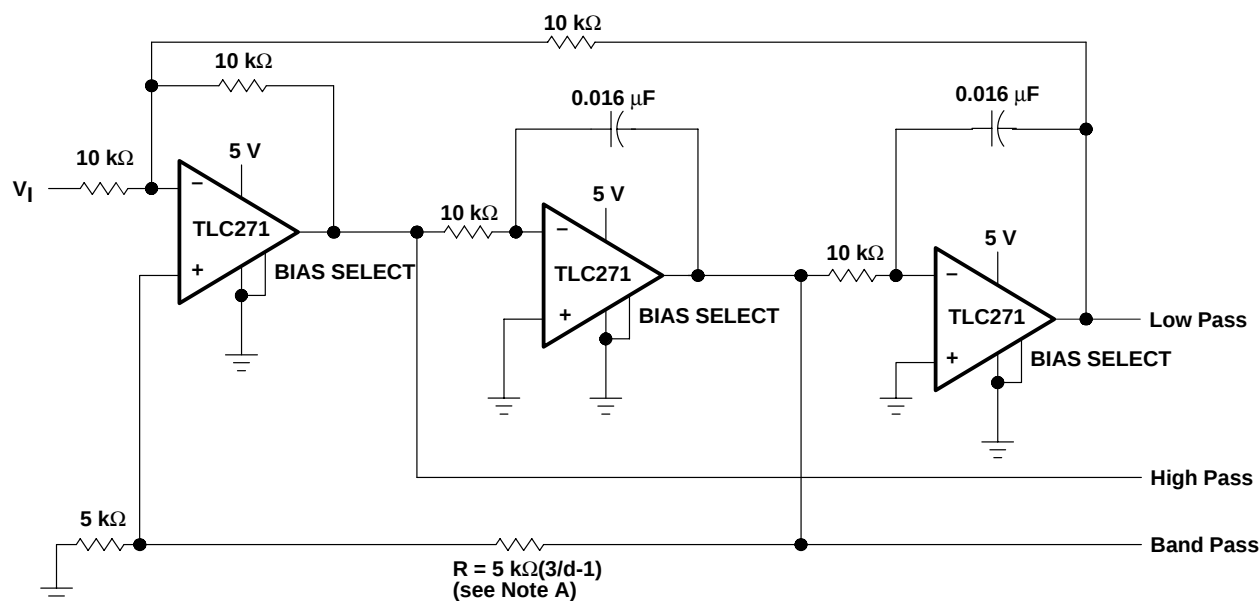


Figure 113. Resistive Pullup to Increase V_{OH}

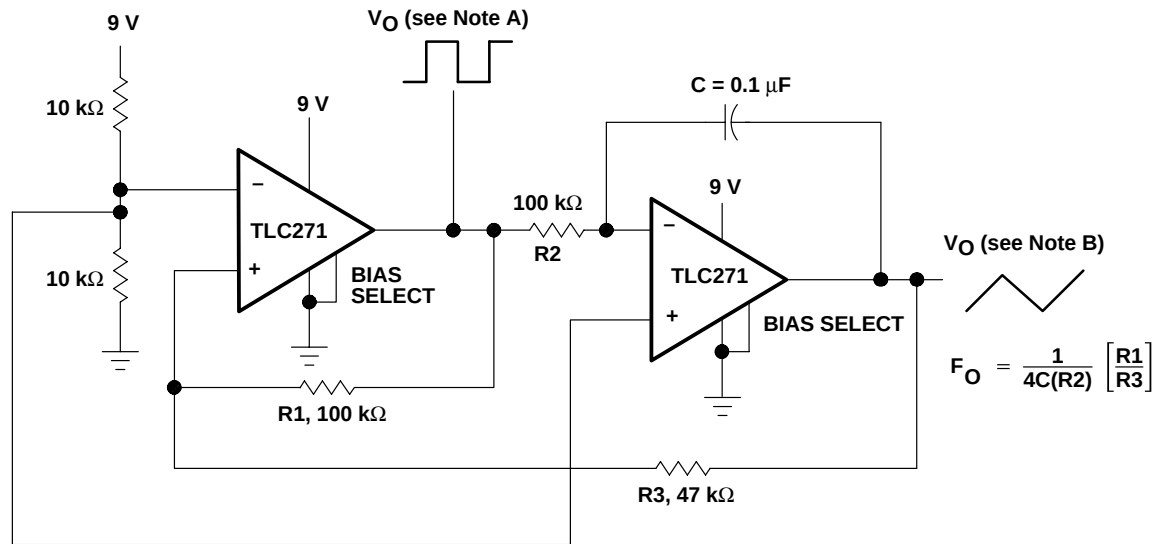


NOTE B: d = damping factor, I/O

Figure 114. State-Variable Filter

APPLICATION INFORMATION

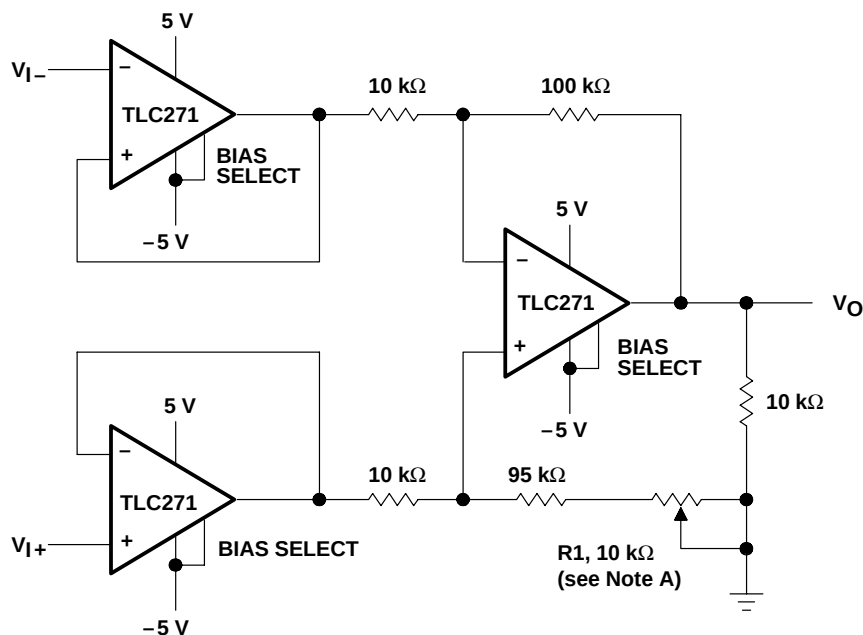
output characteristics (continued)



NOTES: A. $V_{O(PP)} = 8 \text{ V}$
 B. $V_{O(PP)} = 4 \text{ V}$

Figure 115. Single-Supply Function Generator

APPLICATION INFORMATION (HIGH-BIAS MODE)



NOTE A: CMRR adjustment must be noninductive.

Figure 116. Low-Power Instrumentation Amplifier

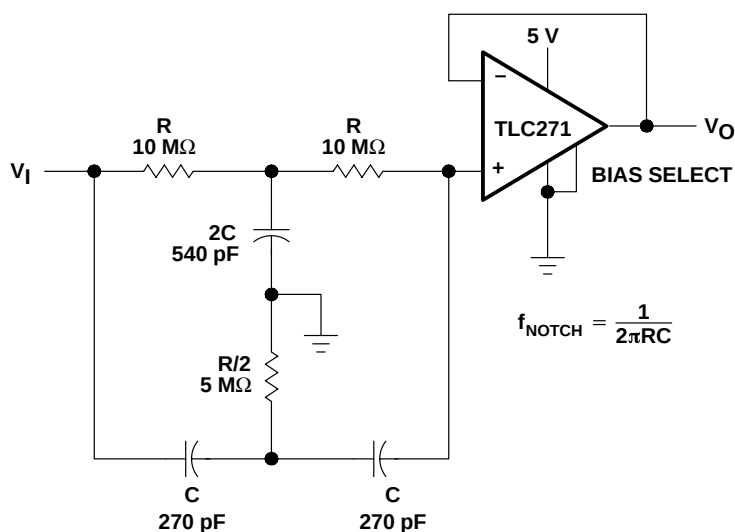
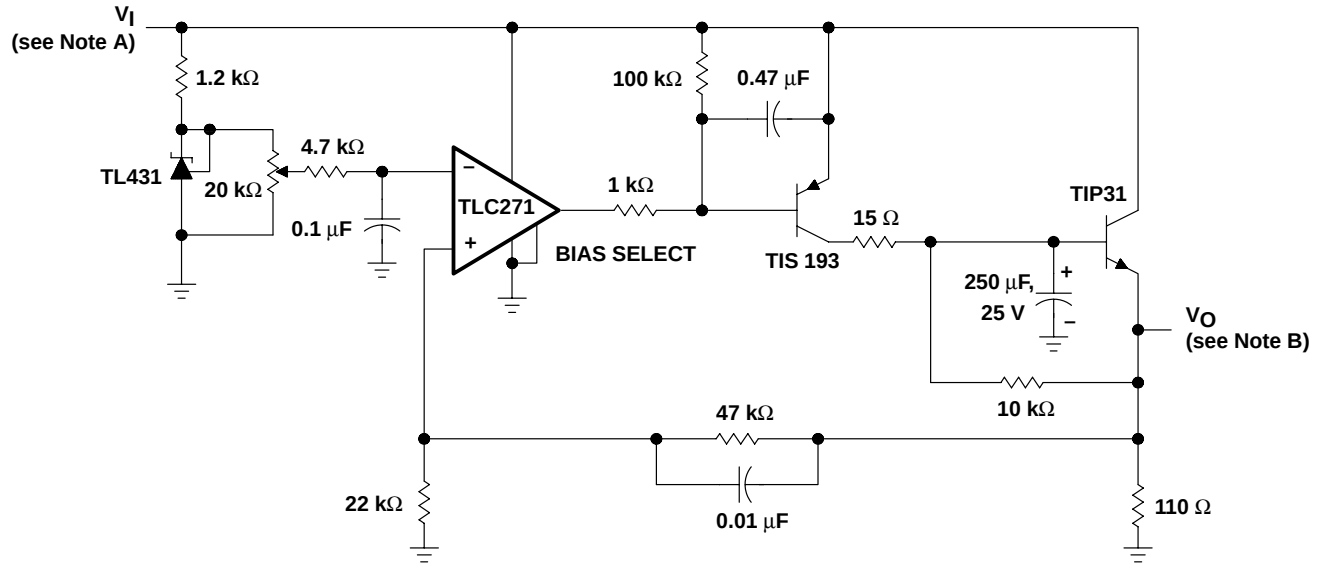


Figure 117. Single-Supply Twin-T Notch Filter

APPLICATION INFORMATION (HIGH-BIAS MODE)



NOTES: A. $V_I = 3.5$ to 15 V
 B. $V_O = 2.0$ V, 0 to 1 A

Figure 118. Logic-Array Power Supply

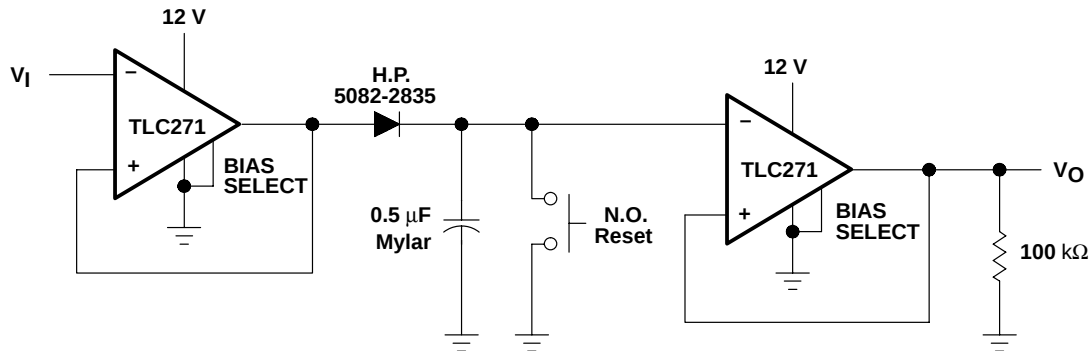
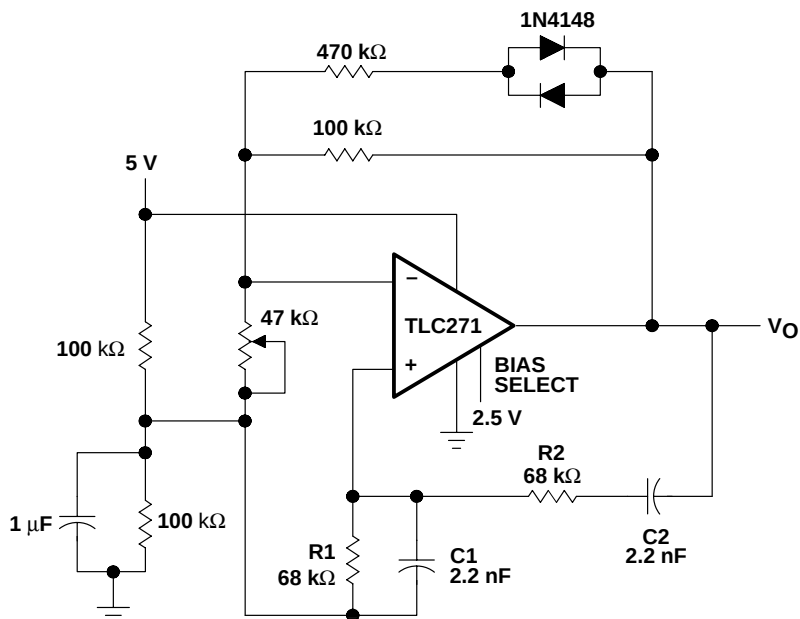


Figure 119. Positive-Peak Detector

APPLICATION INFORMATION (MEDIUM-BIAS MODE)



NOTES: A. $V_{O(pp)} = 2\text{ V}$

B. $f_0 = \frac{1}{2\pi \sqrt{R1R2C1C2}}$

Figure 120. Wein Oscillator

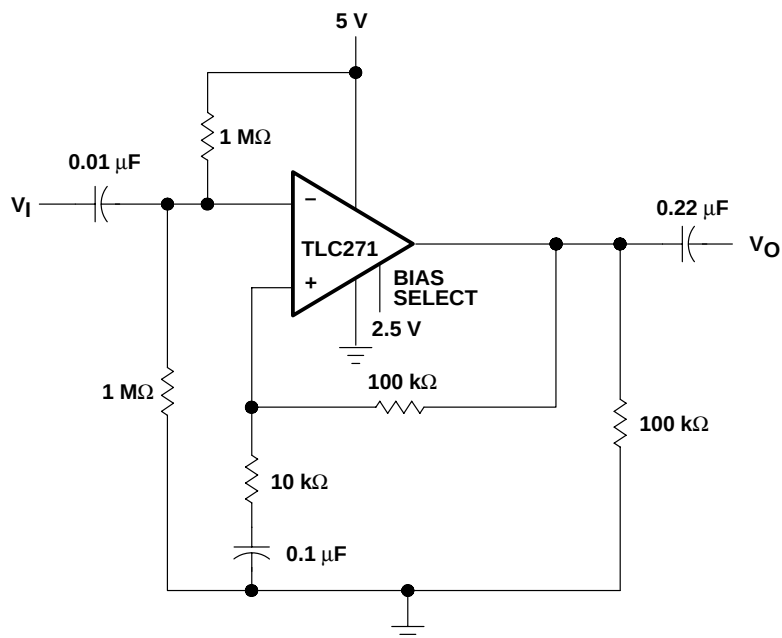
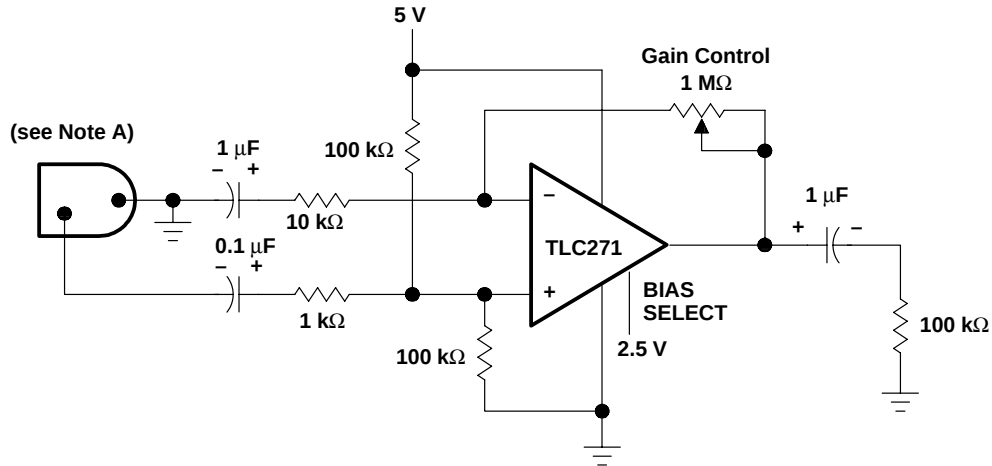


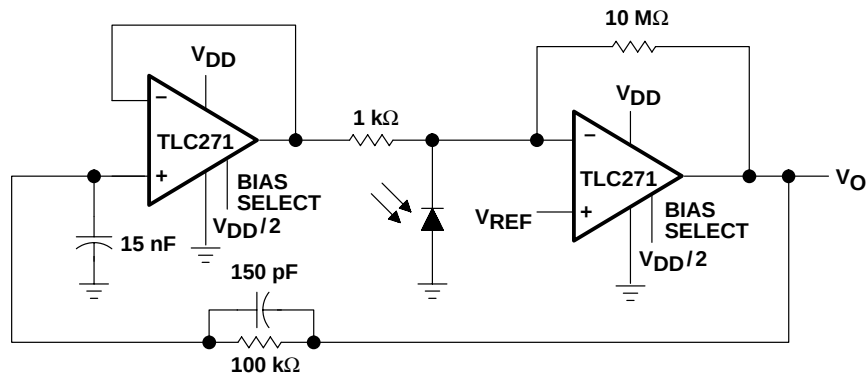
Figure 121. Single-Supply AC Amplifier

APPLICATION INFORMATION (MEDIUM-BIAS MODE)



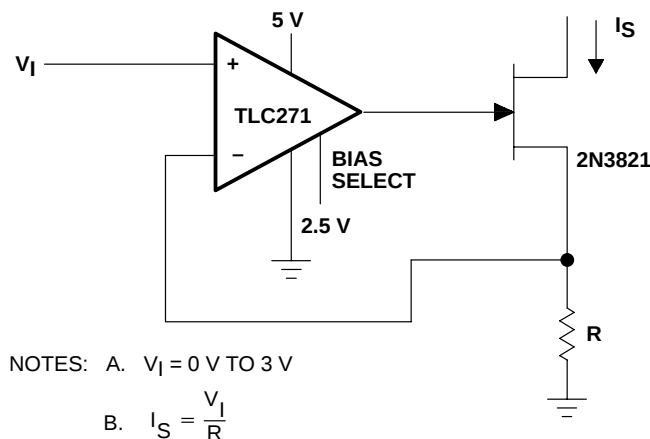
NOTE A: Low to medium impedance dynamic mike

Figure 122. Microphone Preamplifier



NOTES: A. $V_{DD} = 4 \text{ V to } 15 \text{ V}$
 B. $V_{ref} = 0 \text{ V to } V_{DD} - 2 \text{ V}$

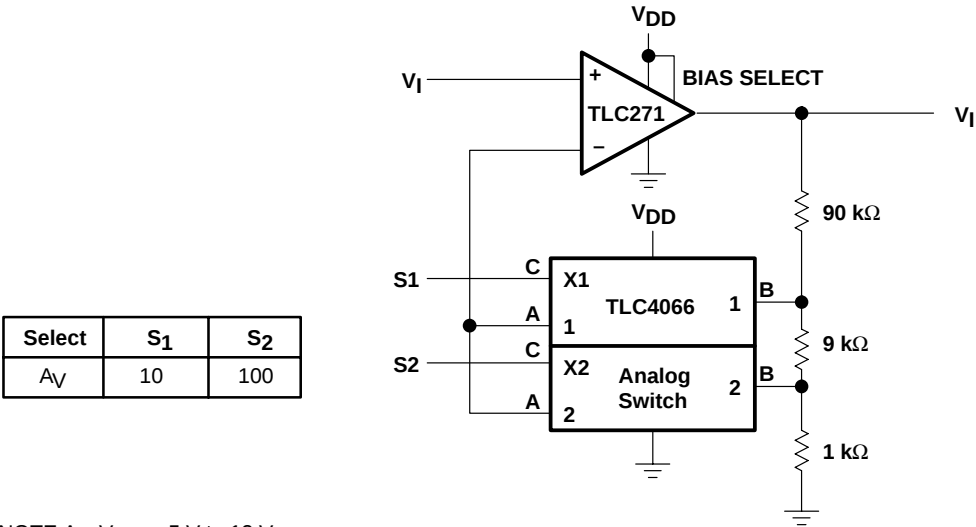
Figure 123. Photo-Diode Amplifier With Ambient Light Rejection



NOTES: A. $V_I = 0 \text{ V to } 3 \text{ V}$
 B. $I_S = \frac{V_I}{R}$

Figure 124. Precision Low-Current Sink

APPLICATION INFORMATION (LOW-BIAS MODE)



NOTE A: V_{DD} = 5 V to 12 V

Figure 125. Amplifier With Digital Gain Selection

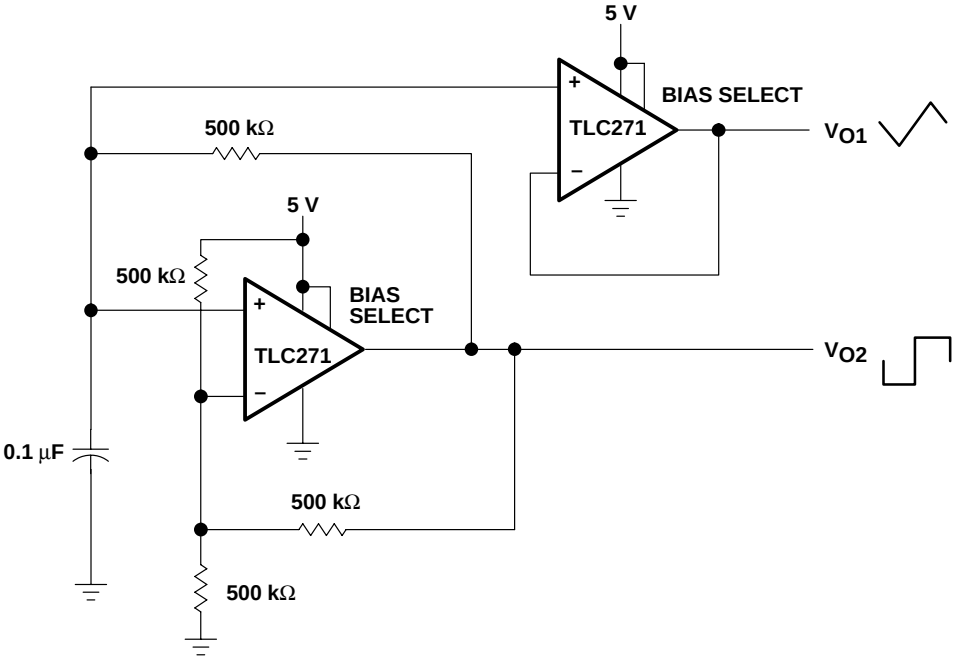
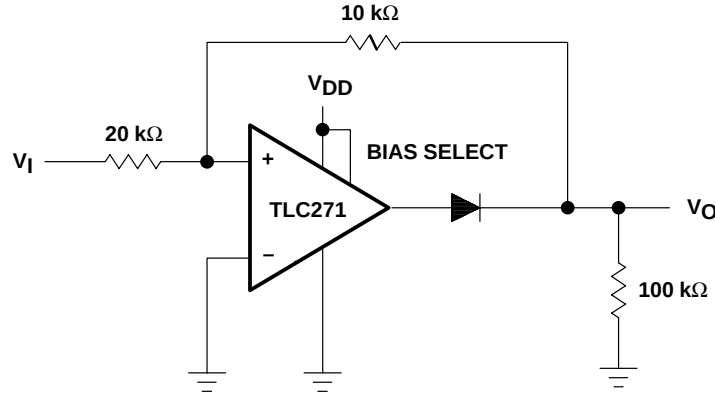


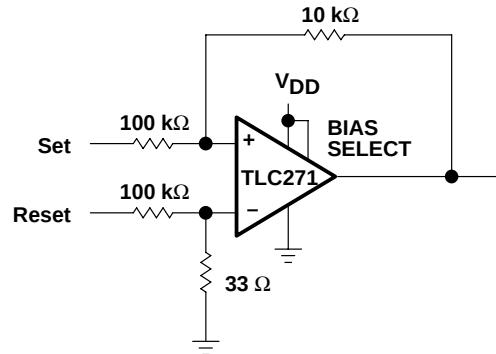
Figure 126. Multivibrator

APPLICATION INFORMATION (LOW-BIAS MODE)



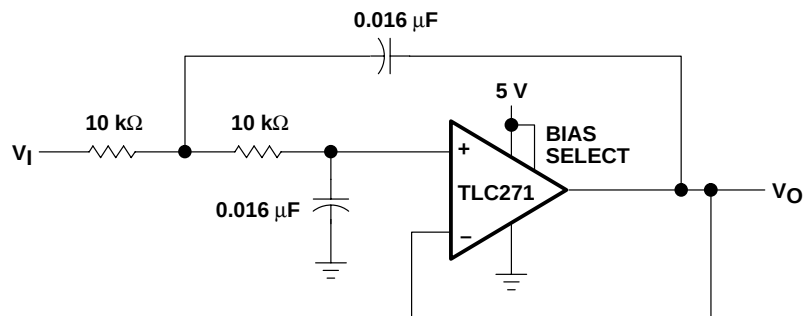
NOTE A: $V_{DD} = 5\text{ V to }16\text{ V}$

Figure 127. Full-Wave Rectifier



NOTE A: $V_{DD} = 5\text{ V to }16\text{ V}$

Figure 128. Set/Reset Flip-Flop



NOTE A: Normalized to $F_C = 1\text{ kHz}$ and $R_L = 10\text{ k}\Omega$

Figure 129. Two-Pole Low-Pass Butterworth Filter

PACKAGING INFORMATION

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Package Qty	Eco Plan ⁽²⁾	Lead/ Ball Finish	MSL Peak Temp ⁽³⁾	Samples (Requires Login)
TLC271ACD	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271ACDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271ACDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271ACDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271ACP	ACTIVE	PDIP	P	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type	
TLC271ACPE4	ACTIVE	PDIP	P	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type	
TLC271ACPSR	ACTIVE	SO	PS	8	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271ACPSRG4	ACTIVE	SO	PS	8	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271AID	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271AIDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271AIDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271AIDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271AIP	ACTIVE	PDIP	P	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type	
TLC271AIPE4	ACTIVE	PDIP	P	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type	
TLC271BCD	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271BCDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271BCDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271BCDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271BCP	ACTIVE	PDIP	P	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type	

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Package Qty	Eco Plan ⁽²⁾	Lead/ Ball Finish	MSL Peak Temp ⁽³⁾	Samples (Requires Login)
TLC271BCPE4	ACTIVE	PDIP	P	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type	
TLC271BID	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271BIDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271BIDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271BIDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271BIP	ACTIVE	PDIP	P	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type	
TLC271BIPE4	ACTIVE	PDIP	P	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type	
TLC271CD	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271CDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271CDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271CDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271CP	ACTIVE	PDIP	P	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type	
TLC271CPE4	ACTIVE	PDIP	P	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type	
TLC271CPSR	ACTIVE	SO	PS	8	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271CPSRG4	ACTIVE	SO	PS	8	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271CPW	ACTIVE	TSSOP	PW	8	150	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271CPWG4	ACTIVE	TSSOP	PW	8	150	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271CPWLE	OBSOLETE	TSSOP	PW	8		TBD	Call TI	Call TI	
TLC271CPWR	ACTIVE	TSSOP	PW	8	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271CPWRG4	ACTIVE	TSSOP	PW	8	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Package Qty	Eco Plan ⁽²⁾	Lead/ Ball Finish	MSL Peak Temp ⁽³⁾	Samples (Requires Login)
TLC271ID	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271IDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271IDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271IDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271IP	ACTIVE	PDIP	P	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type	
TLC271IPE4	ACTIVE	PDIP	P	8	50	Pb-Free (RoHS)	CU NIPDAU	N / A for Pkg Type	
TLC271MDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271MDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	
TLC271MJG	OBSOLETE	CDIP	JG	8		TBD	Call TI	Call TI	

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

⁽³⁾ MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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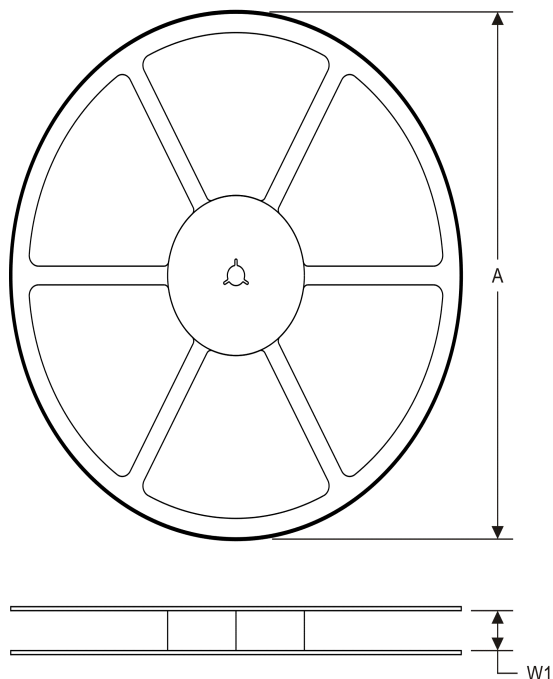
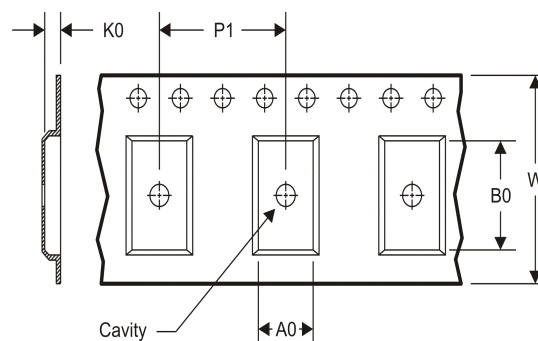
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PACKAGE OPTION ADDENDUM

23-May-2012

continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

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TAPE AND REEL INFORMATION
REEL DIMENSIONS

TAPE DIMENSIONS


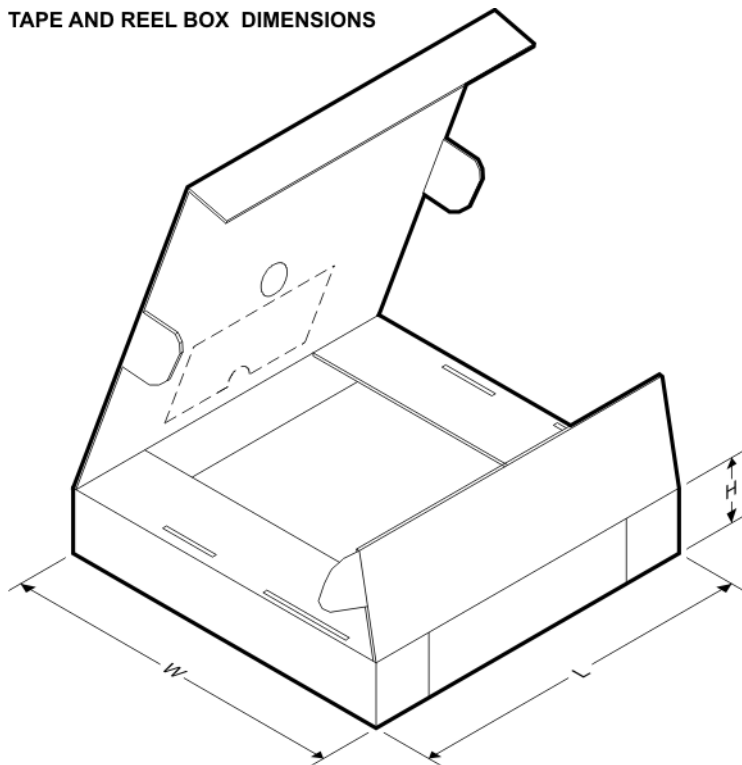
A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

TAPE AND REEL INFORMATION

*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLC271ACDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC271ACPSR	SO	PS	8	2000	330.0	16.4	8.2	6.6	2.5	12.0	16.0	Q1
TLC271AIDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC271BCDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC271BIDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC271CDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC271CPSR	SO	PS	8	2000	330.0	16.4	8.2	6.6	2.5	12.0	16.0	Q1
TLC271CPWR	TSSOP	PW	8	2000	330.0	12.4	7.0	3.6	1.6	8.0	12.0	Q1
TLC271IDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLC271MDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1

TAPE AND REEL BOX DIMENSIONS

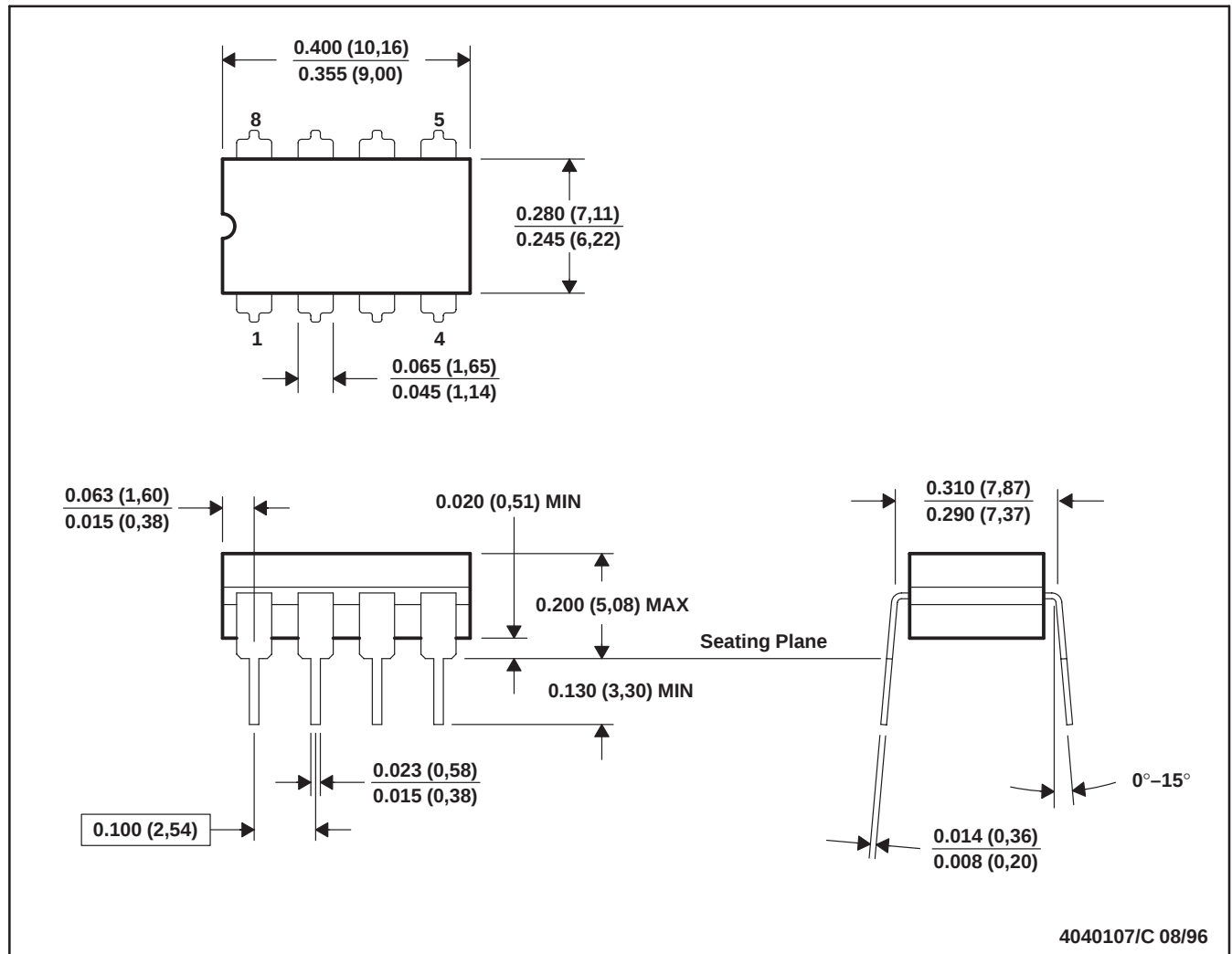


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TLC271ACDR	SOIC	D	8	2500	340.5	338.1	20.6
TLC271ACPSR	SO	PS	8	2000	367.0	367.0	38.0
TLC271AIDR	SOIC	D	8	2500	340.5	338.1	20.6
TLC271BCDR	SOIC	D	8	2500	340.5	338.1	20.6
TLC271BIDR	SOIC	D	8	2500	340.5	338.1	20.6
TLC271CDR	SOIC	D	8	2500	340.5	338.1	20.6
TLC271CPSR	SO	PS	8	2000	367.0	367.0	38.0
TLC271CPWR	TSSOP	PW	8	2000	367.0	367.0	35.0
TLC271IDR	SOIC	D	8	2500	340.5	338.1	20.6
TLC271MDR	SOIC	D	8	2500	367.0	367.0	35.0

JG (R-GDIP-T8)

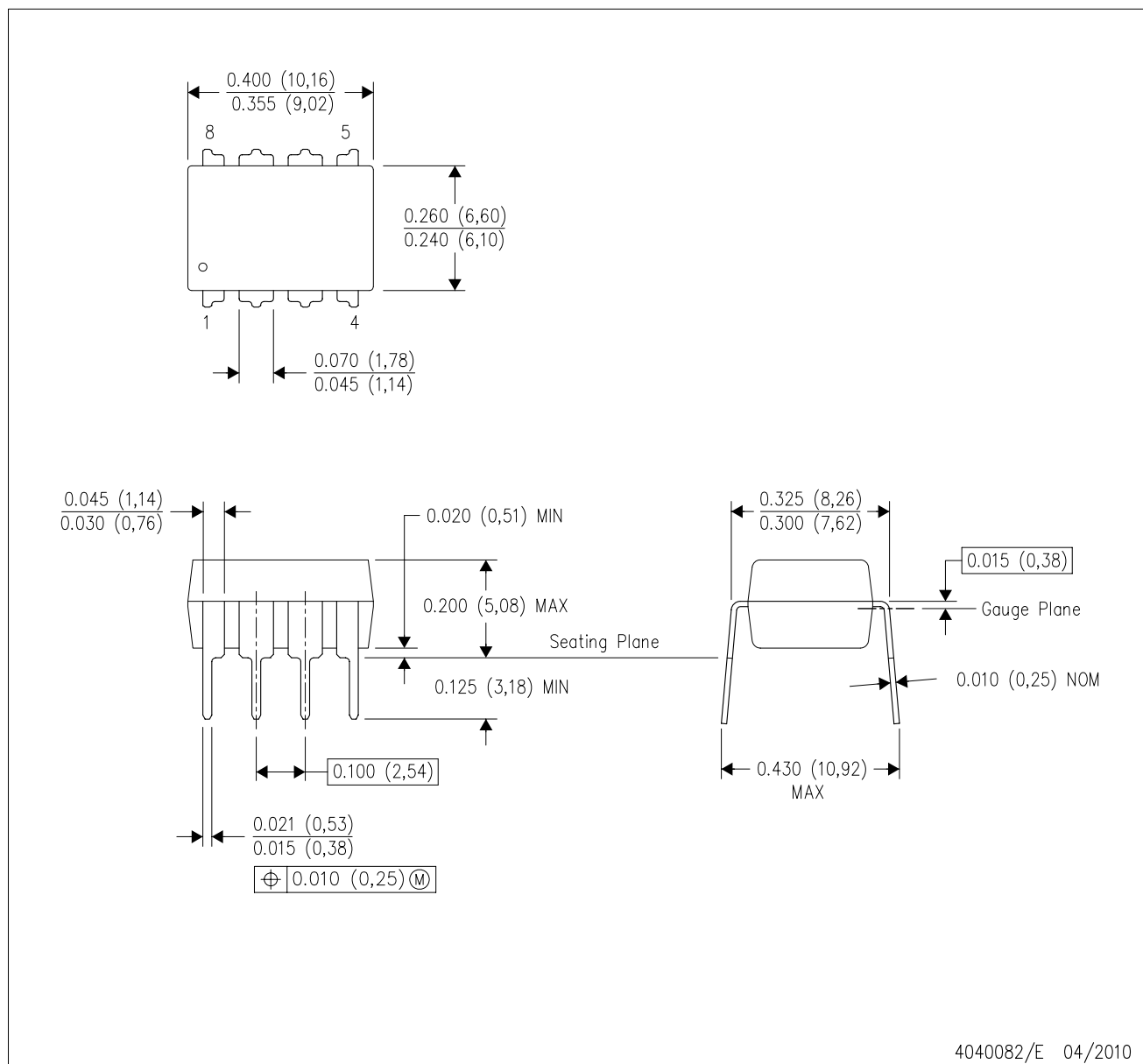
CERAMIC DUAL-IN-LINE



- NOTES:
- A. All linear dimensions are in inches (millimeters).
 - B. This drawing is subject to change without notice.
 - C. This package can be hermetically sealed with a ceramic lid using glass frit.
 - D. Index point is provided on cap for terminal identification.
 - E. Falls within MIL STD 1835 GDIP1-T8

P (R-PDIP-T8)

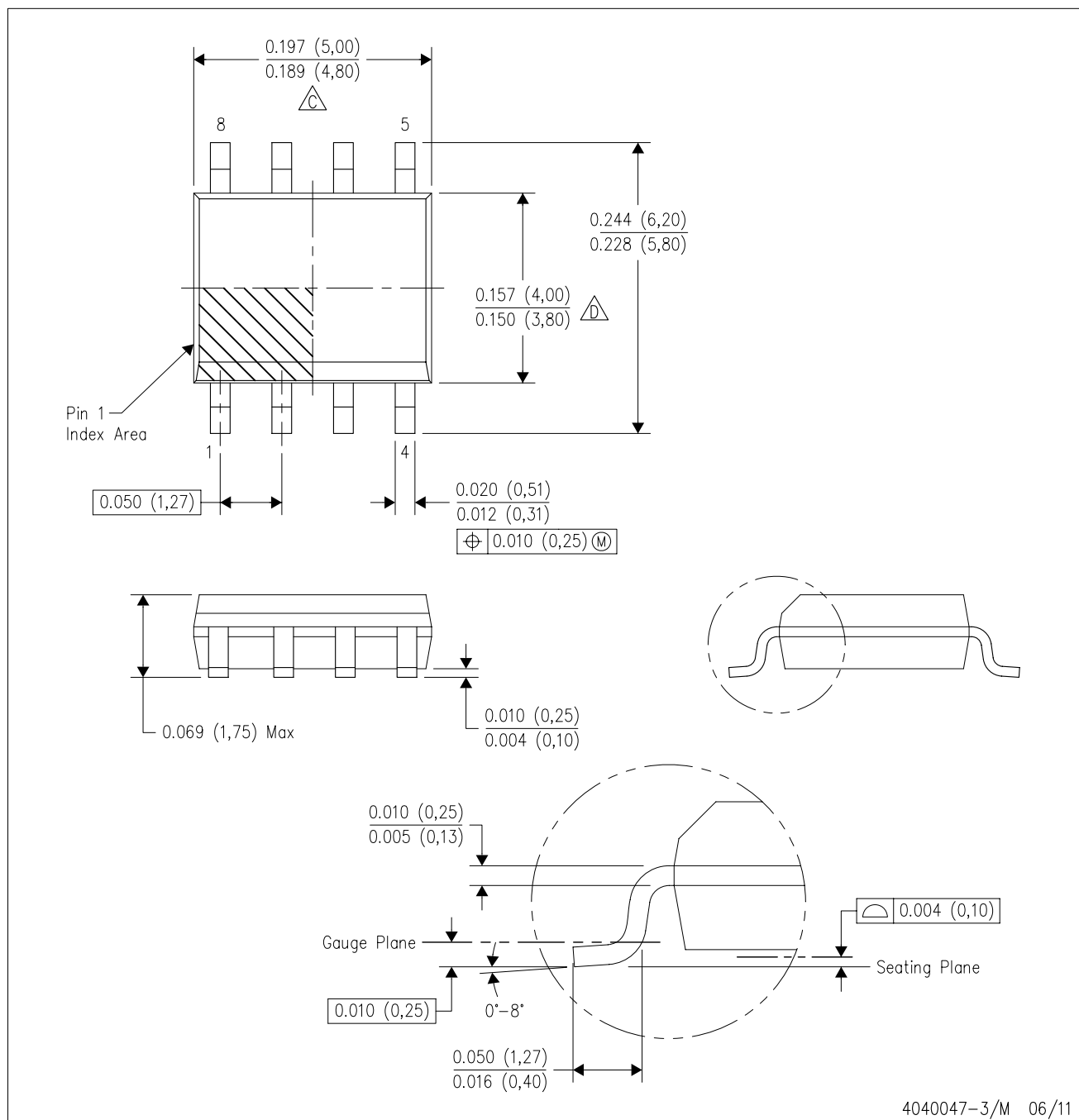
PLASTIC DUAL-IN-LINE PACKAGE



- NOTES:
- A. All linear dimensions are in inches (millimeters).
 - B. This drawing is subject to change without notice.
 - C. Falls within JEDEC MS-001 variation BA.

D (R-PDSO-G8)

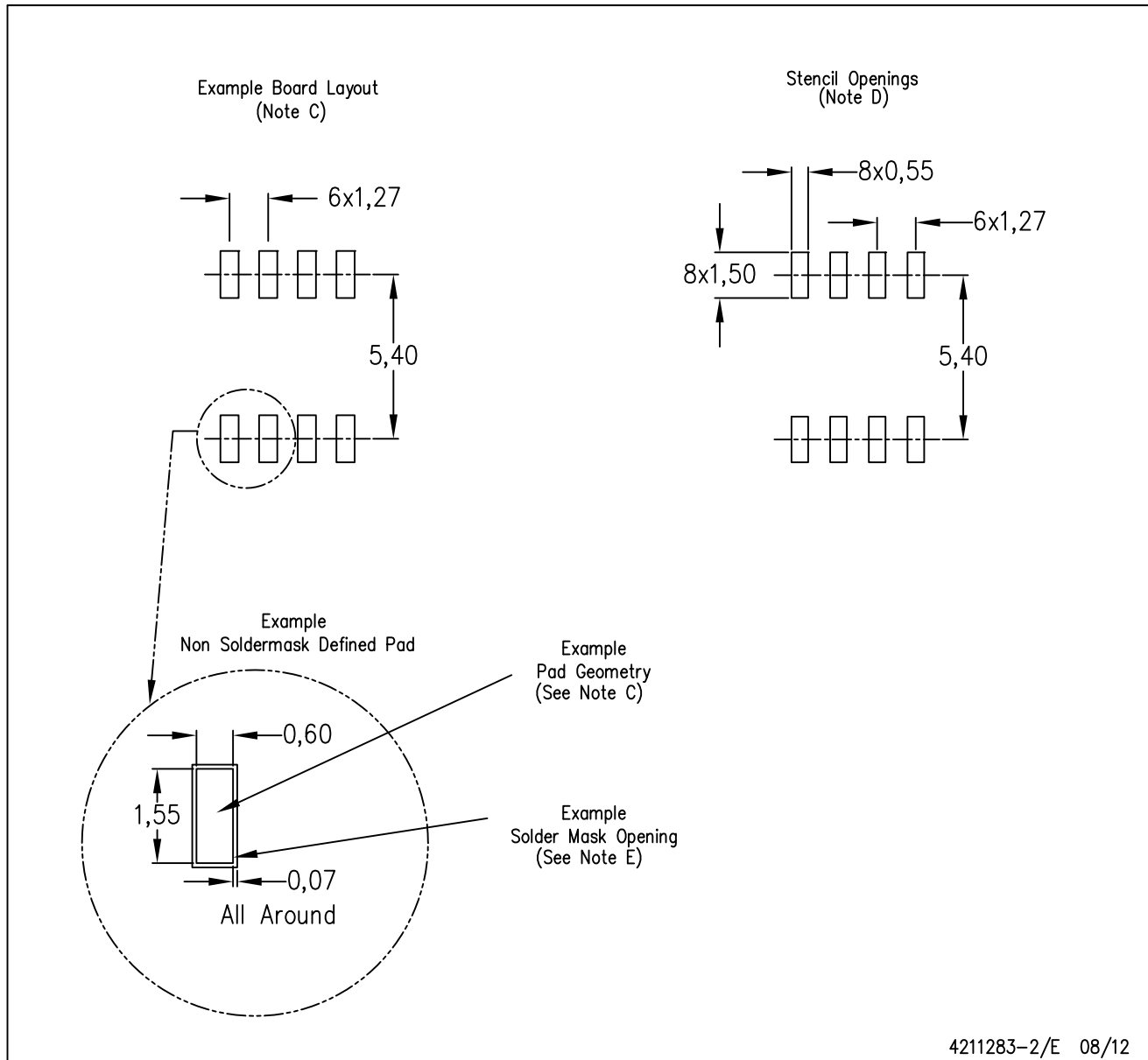
PLASTIC SMALL OUTLINE



- NOTES:
- A. All linear dimensions are in inches (millimeters).
 - B. This drawing is subject to change without notice.
 - $\triangle C$ Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
 - $\triangle D$ Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
 - E. Reference JEDEC MS-012 variation AA.

D (R-PDSO-G8)

PLASTIC SMALL OUTLINE



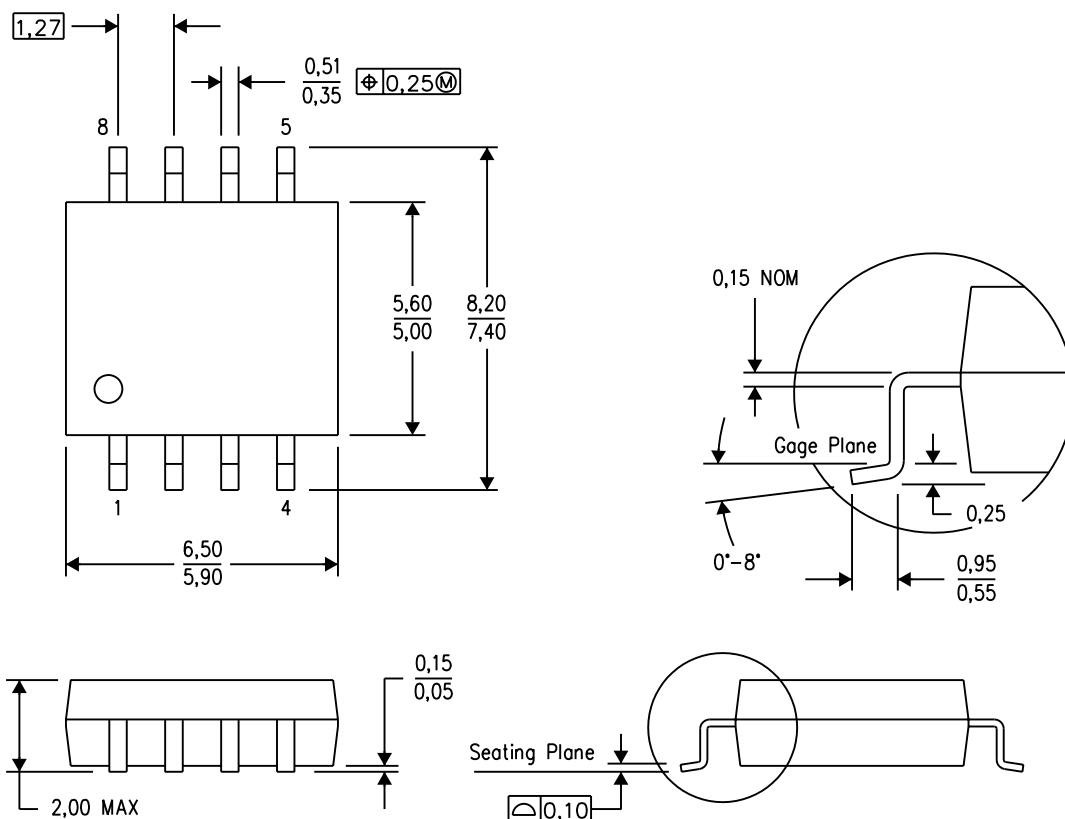
4211283-2/E 08/12

- NOTES:
- A. All linear dimensions are in millimeters.
 - B. This drawing is subject to change without notice.
 - C. Publication IPC-7351 is recommended for alternate designs.
 - D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
 - E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

MECHANICAL DATA

PS (R-PDSO-G8)

PLASTIC SMALL-OUTLINE PACKAGE

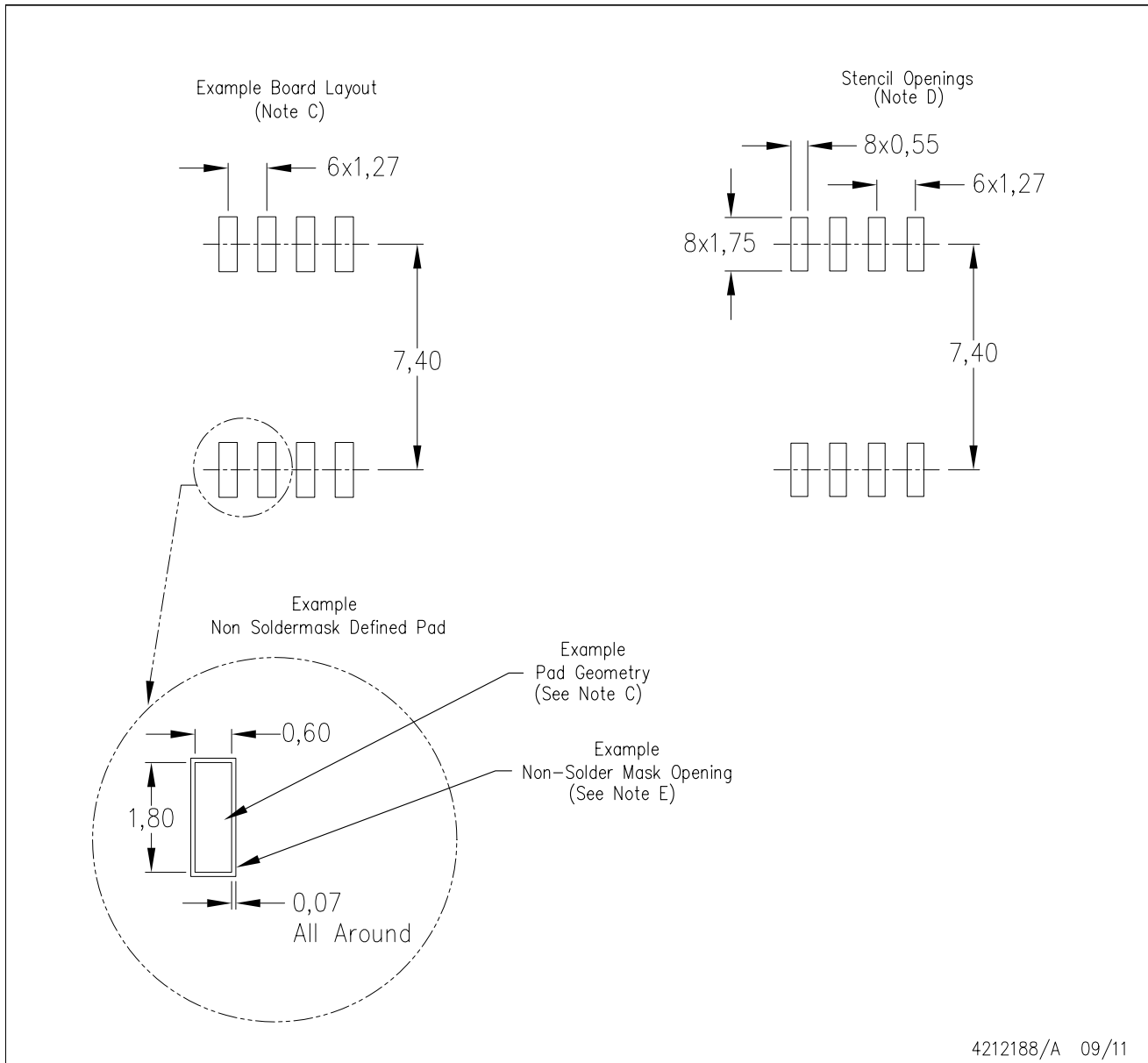


4040063/C 03/03

- NOTES:
- A. All linear dimensions are in millimeters.
 - B. This drawing is subject to change without notice.
 - C. Body dimensions do not include mold flash or protrusion, not to exceed 0,15.

PS (R-PDSO-G8)

PLASTIC SMALL OUTLINE

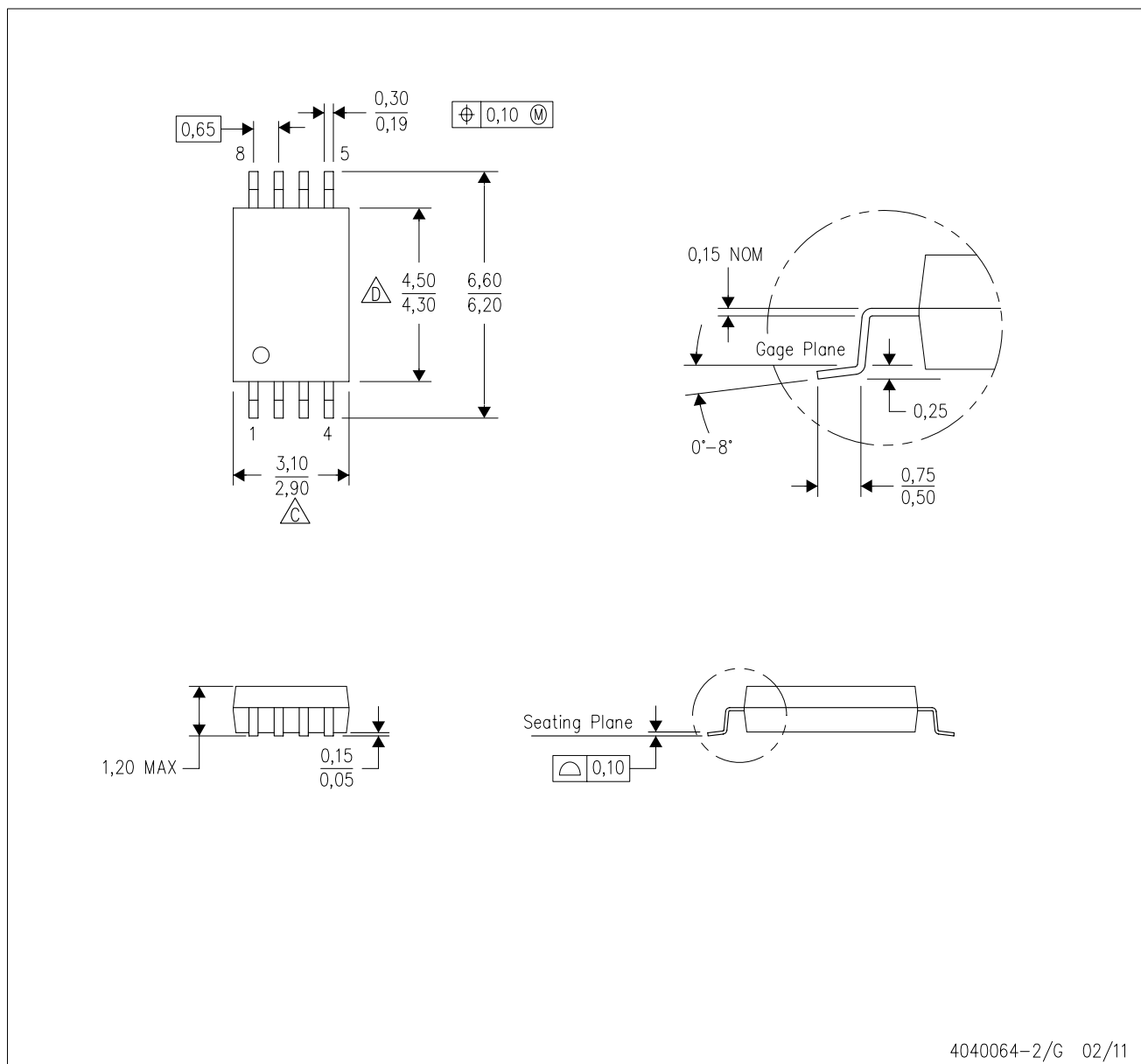


4212188/A 09/11

- NOTES:
- All linear dimensions are in millimeters.
 - This drawing is subject to change without notice.
 - Publication IPC-7351 is recommended for alternate designs.
 - Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
 - Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

PW (R-PDSO-G8)

PLASTIC SMALL OUTLINE



- NOTES:
- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
 - B. This drawing is subject to change without notice.
 - C. Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0,15 each side.
 - D. Body width does not include interlead flash. Interlead flash shall not exceed 0,25 each side.
 - E. Falls within JEDEC MO-153

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