Application Report
SWRA420–February 2013

CC256x QFN PCB Guidelines

ABSTRACT

This application report presents the printed circuit board (PCB) design guidelines, including placement and layout, for the TI CC256x QFN integrated circuit (IC).

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1 Introduction

The CC256x QFN IC is a Bluetooth RF radio device targeted to operate on commercial platforms. The CC256x QFN reference design is based on the following requirements:

- A single-sided assembly
- A low manufacturing cost for PCBs

When designing a PCB for the CC256x QFN device, design rules and layout guidelines must be considered to achieve optimum performance.

This application report complements, and does not replace, the CC256x QFN product data sheet. TI advises the design engineer to use the data sheet and available application notes for the system design.

TI recommends that designers and users follow the design rules described in this application report to achieve performance similar to that obtained with the TI reference design. Following the given design rules minimizes the number of board spins required to achieve optimal performance.

If you have questions or cannot completely meet the guidelines described in this document, consult with your TI representative before releasing layout files to production.

1.1 Reference Documents

- CC256x QFN Datasheet (SWRS121)
- CC256x QFN EM Reference Design (SWRR117)
- CC256x System Design Guide
- *QFN/SON PCB Attachment Application Report* (SLUA271)
- Application Notes for Surface Mount Assembly of Amkor's Dual Row MicroLeadFrame (MLF) Packages

2 PCB Stack-Up

The recommendations in this document refer to a four-layer, CC256x QFN PCB based on standard flame-retardant 4 (FR4) material, which is the technology commonly used for commercial applications (see Figure 1).

The layout recommendations for the CC256x QFN device described in this document are based on two routable layers and two solid ground layers (layer 2 and layer 4).

The pads of the QFN package are 0.6-mm pitch; therefore, when using standard routing rules to fan out, all required signals must be used. The design is based on a single-side assembly to reduce complexity and fabrication cost.

Figure 1. PCB Stack-Up

The following layout rules apply to this design:

- The board is 0.062 inch thick.
- The smallest trace width is 5 mils.
- The smallest clearance from pad to trace is 4 mils.

- The via size is 0.018 inch with a finished hole side of 0.008 inch.

3 Placement Guidelines

Figure 2 shows the reference design schematic for the CC256x QFN IC.

Figure 3 shows the component placement recommended by TI. The PCB area for the CC256x QFN reference circuit is 0.65 inch x 0.65 inch and fits inside a standard RF shield can (Laird Technologies, BMI-S-202-F/BMI-S-202-C). The total area is 0.65 inch × 0.65 inch (0.42 inch^2) or 16.5 mm × 16.5 mm (272 mm^2).

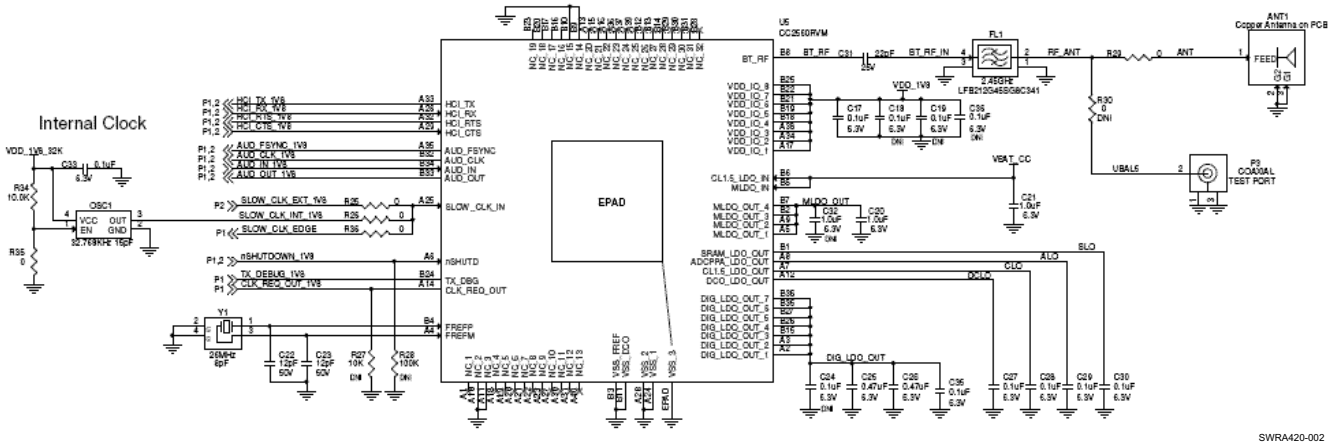


Figure 2. Reference Design Schematic

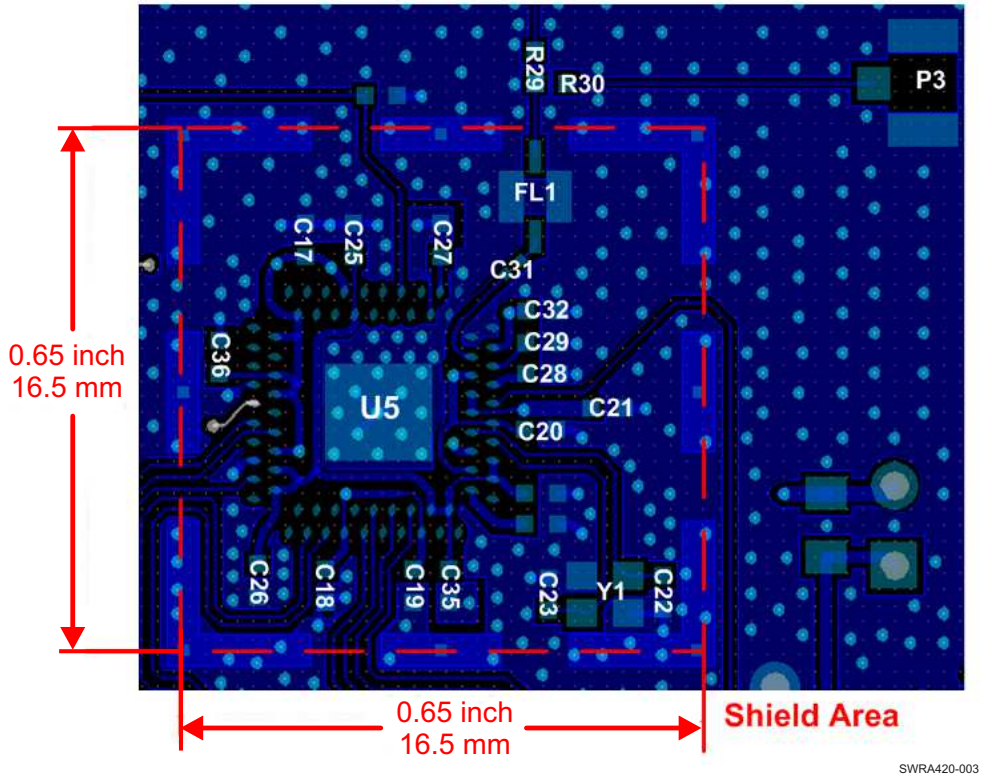


Figure 3. Component Placement

3.1 External Components

The passive components must be placed as close as possible to the CC256x QFN device. Traces must be as short as possible: The general trace width guideline is approximately 10 mils (.010 inch or 0.254 mm).

Pay special attention to the routing for these signals:

- Input/output (I/O) signals to the on-chip power-supply regulators
- Crystal I/O signals
- RF signals

3.2 Low-Dropout Capacitors

Follow these guidelines for the low-dropout (LDO) capacitors (see Figure 4):

- Ensure that the decoupling capacitors and traces to the capacitors that connect the LDO outputs are as short as possible.
- Ensure that the traces are wide.
- Keep the device and the capacitors together on the top side.
- Connect each capacitor ground connection directly to a solid ground layer (layer 2).
- Place the digital LDO decoupling capacitors (C25, C26, and C35) as close as possible to U5, the CC256x QFN IC relevant ball. For C35, create a small ground plane that is isolated from the ground of layer 1 and connect the capacitor ground pad to the solid ground (layer 2).
- Place the decoupling capacitor of MLDO_OUT (C20) as close as possible to U5.
- Using a via, directly connect each capacitor ground to the solid ground (layer 2). In addition, add many vias to increase the coupling to ground.

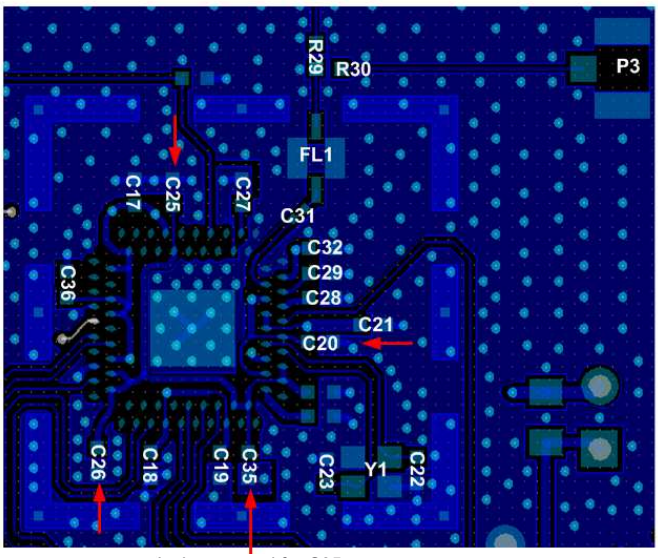


Figure 4. LDO Capacitor Placement

3.3 DCO_LDO_OUT

Follow these guidelines for DCO_LDO_OUT, the digital control oscillator signal (see Figure 5):

- Connect capacitor (C27) directly to ball A12 (DCO_LDO_OUT).
- Using a via, isolate the capacitor (C27) ground from the layer 1 ground and connect the capacitor (C27) ground directly to the layer 2 solid ground.

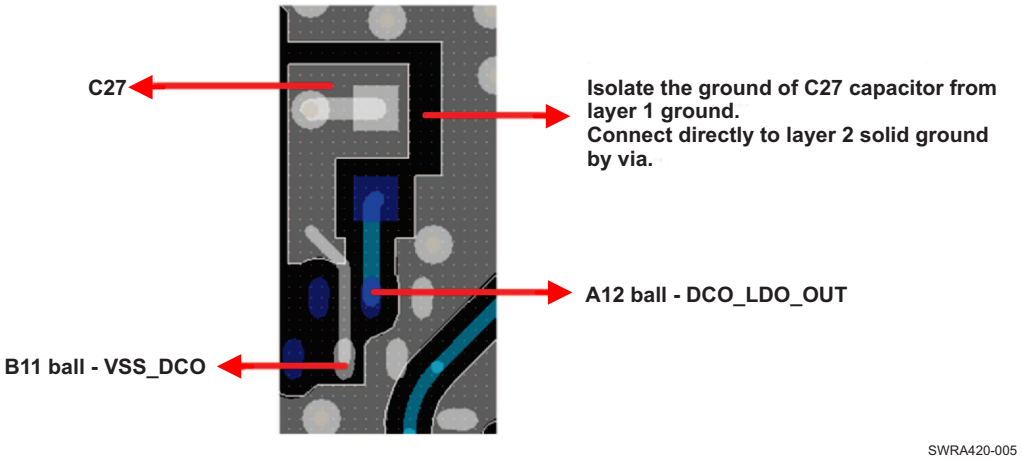


Figure 5. DCO_LDO_OUT Instructions

4 Routing Guidelines

4.1 RF Interface

When designing the RF path, follow the placement and layout rules described in this section.

4.1.1 RF Trace

Route the Bluetooth RF path on the top layer (component side) and keep traces as short as possible. The Bluetooth RF trace must be a 50-Ω, impedance-controlled trace with reference to solid ground (layer 2). Figure 6 shows the RF path. BT_RF (ball B8) is an inner ball. The RF line must be routed between B9 and B10, which are ground balls.

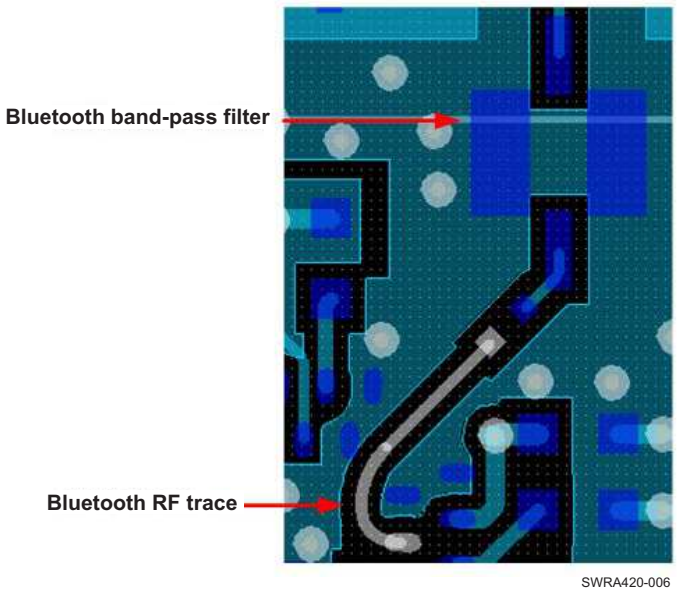


Figure 6. Bluetooth RF Path

4.1.2 Band Pass Filter

The area beneath the filter pads must be filled with ground on layer 1 and layer 2; however, the RF_IN and RF_OUT pads must be kept clear of any ground fill. The local ground beneath the filter area can be strengthened with vias to solid ground (layer 2).

Because the Bluetooth RF output has a DC level, a serial capacitor (C31) must be placed between the CC256x QFN RF port and the filter. Alternatively, a filter with a DC block can be used.

The BPF must meet the following minimum attenuation values to pass regulatory certification. Table 1 lists the band-pass filter (BPF) rejection requirements over frequency for the CC256x QFN solution. Figure 7 shows the specification data for the BPF used in the reference design.

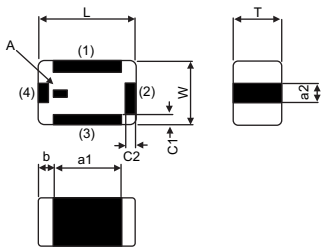
Table 1. Bluetooth BPF Requirements

Harmonic	Center Frequency (MHz)	Min BPF Attenuation (dB)
Fundamental	2450	1.5
2	4900	26
3	7350	26
4	9800	19
5	12250	12
6	14700	9

1. Characteristics (at 25°C)

Tentative part number	LFB212G45SG8C341TEMP
Nominal characteristics impedance	50 ohm
Nominal Center Frequency	2450.00 MHz
Pass Band Range (BW)	fo ± 50.00 MHz
Insertion Loss in BW	1.42 dB typ. at 25 °C
Attenuation (absolute value)	31.9 dB typ. at 1200.00 ~1300.00 MHZ 15.7 dB typ. at 2000.00 MHZ 17.1 dB typ. at 3000.00 MHZ 28.6 dB typ. at 3600.00 ~3800.00 MHZ 34.8 dB typ. at 4800.00 ~ 5000.00 MHZ 36.8 dB typ. 7200.00 ~ 7500.00 MHz
V.S.W.R. in BW	1.98 typ.
Power capacity	500 mW max.

2. Construction, Dimensions and marking



Mark	Meaning
A	Directional Input Mark

(in mm)

Mark	Dimension	Mark	Dimension
L	2.00±0.15	a2	0.30+0, 10/–0,15
W	1.25±0.10	b	0.20±0.16
T	0.95±0.10	c1	0.3+0.1/–0.2
a1	1.60+0.10/–0_16	c2	0.3+0.1/–0.2

TERMINAL COFIGURATION

Terminal No.	Terminal Name	Terminal No.	Terminal Name
(1)	GND	(3)	GND
(2)	OUT	(4)	IN

3. Land pattern

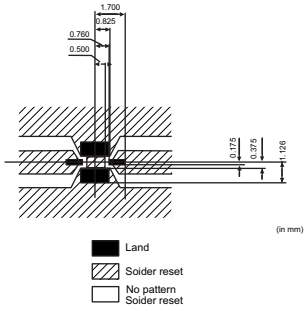


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Figure 7. BPF Specification Data

4.1.3 Antenna

The antenna used in the reference design is a PCB-type inverted F antenna (see Figure 8). For more information on implementing the inverted F antenna design, see *Design Note DN007 2.4 GHz Inverted F Antenna* (SWRU120).

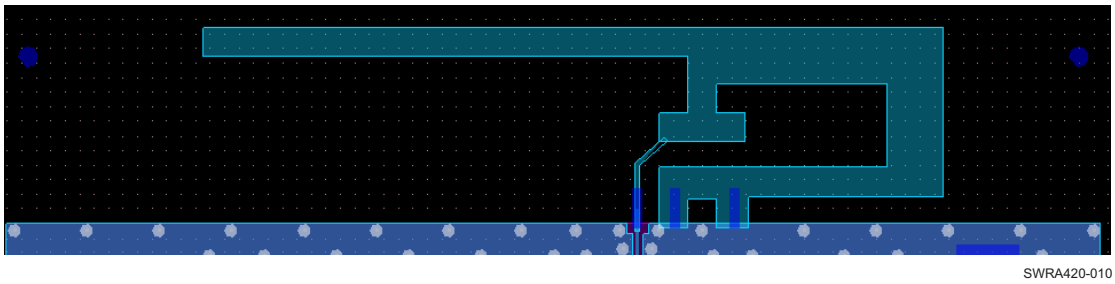

Figure 8. Inverted F Antenna on CC256x QFN Reference Design

4.1.3.1 Description of the Inverted F Antenna Design

The impedance of the inverted F antenna is matched directly to 50 Ω; thus, external matching components are not required. However, TI recommends leaving the matching network placeholder preparation.

4.1.3.2 Implementation of the Inverted F Antenna

For optimum performance, create an exact copy of the antenna dimensions (see Figure 9 and Table 2). To implement the antenna in a PCB CAD tool, the easiest approach is to import the antenna layout from a Gerber file or a .DXF file. If the antenna is implemented on a PCB that is wider than the antenna, avoid placing components or having a ground plane close to the end points of the antenna.

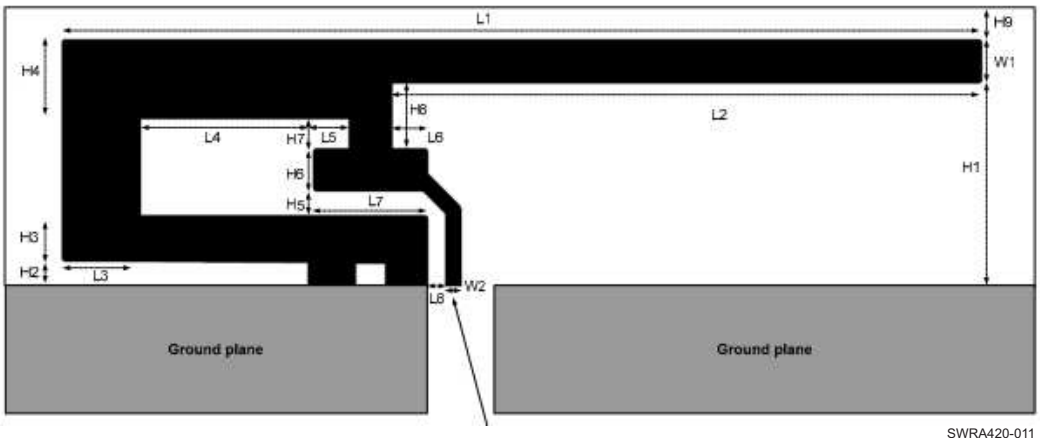

Figure 9. Inverted F Antenna Dimensions

Table 2. Inverted F Antenna Dimensions

Dimension	Measurement (mm)	Dimension	Measurement (mm)
H1	5.70 mm	W2	0.46 mm
H2	0.74 mm	L1	25.58 mm
H3	1.29 mm	L2	16.40 mm
H4	2.21 mm	L3	2.18 mm
H5	0.66 mm	L4	4.80 mm
H6	1.21 mm	L5	1.00 mm
H7	0.80 mm	L6	1.00 mm

Table 2. Inverted F Antenna Dimensions (continued)

Dimension	Measurement (mm)	Dimension	Measurement (mm)
H8	1.80 mm	L7	3.20 mm
H9	0.61 mm	L8	0.45 mm
W1	1.21 mm	–	–

4.1.4 RF Shielding Recommendations

The CC256x QFN reference design accepts RF shield with side dimensions of 0.65 inch x 0.65 inch. Although not required, the RF shield is recommended because the shield protects the CC256x Bluetooth from RF interference and the fast-clock crystal oscillator (26 MHz) from VHF and UHF interference. This protection optimizes the fast-clock phase noise as well as RF spurious content. The RF shield chosen for the CC256x QFN reference design is from Laird Technologies, BMI-S-202-F/BMI-S-202-C.

4.2 Clocks

Clock signal routing directly influences RF performance due to the signal trace susceptibility to noise. Therefore, the clock signal lines must be as short as possible. Clock signal traces must have a ground plane on each side of the signal trace to reduce undesired signal coupling. To reduce capacitive coupling of undesired signals into the clock line, do not route clock traces above or below other signals (especially digital signals).

4.2.1 Fast Clock – Crystal

If a crystal oscillator is used, the parasitic characteristics of the clock trace influence the oscillation. Traces must be kept as short as possible. Traces that are too wide can cause excessive capacitance and traces that are too narrow can cause parasitic inductance of the clock trace. For short clock traces, use a trace width of approximately 10 mils (0.010 inch or 0.254 mm). Keep the crystal tune capacitors (C22 and C23) close to the crystal pads.

TI recommends that you avoid crossing the crystal lines on the adjacent layers. Nevertheless, TI recommends keeping the ground plane under the crystal line to improve the return path. Figure 10 shows the crystal trace routing.

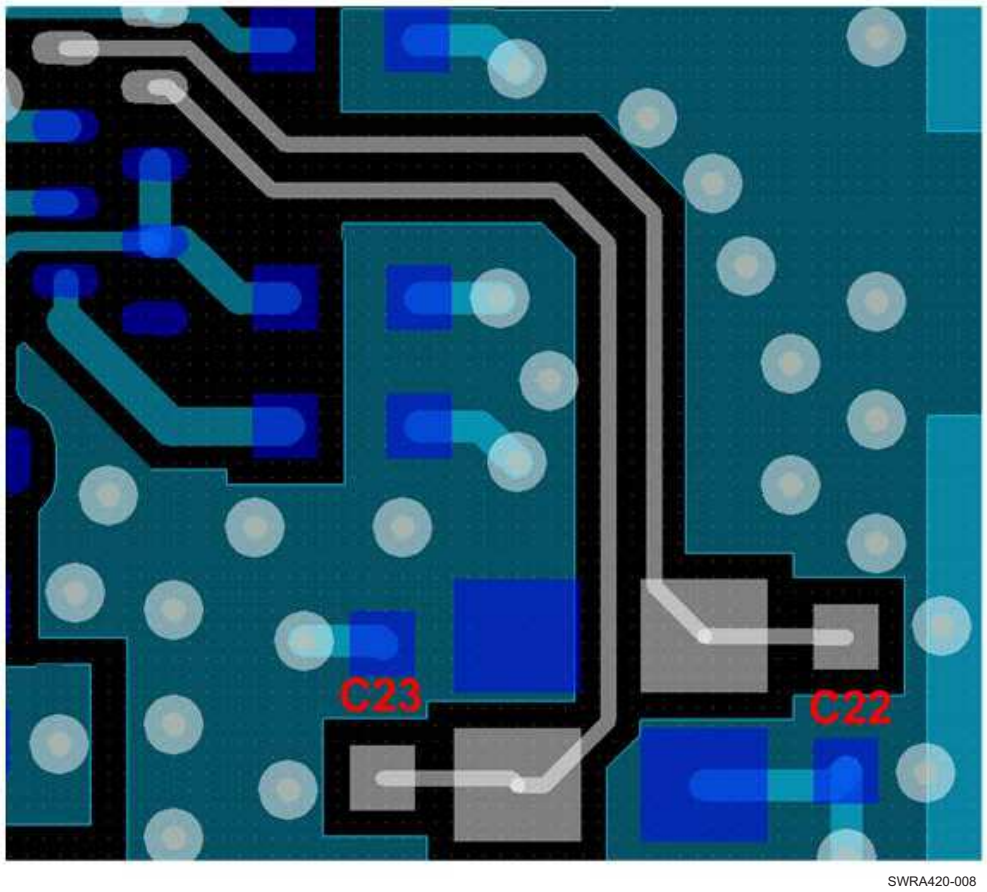

Figure 10. Crystal Trace Routing

4.2.2 Slow Clock

The slow clock signal lines must be as short as possible. Traces of slow clock signals should have a ground plane on each side of the signal trace to reduce undesired signal coupling. To reduce capacitive coupling of undesired signals into the clock line, slow clock traces must not be routed above or below other signals (especially digital signals).

4.2.3 Digital Clock Signals

The audio clock (AUD_CLK, ball B32) signal is a digital clock signal and can be source of interference for adjacent signals. Avoid laying this type of signal trace next to sensitive signals (for example, the RF path, DCO_LDO, and other LDO capacitors). Keep the traces of these signals as short as possible and maintain a maximum clearance value from other traces. Run the lines with ground on the adjacent layer to improve the return path and isolation. TI recommends running these lines as a bus interface.

4.3 Interfaces

4.3.1 Audio Lines Routing Guidelines

The digital audio lines (pulse-code modulation [PCM]) are high-speed digital lines in which the four wires (AUD_CLK, AUD_FSYNC, AUD_IN, and AUD_OUT) must be roughly the same length. These lines are high-speed digital and must be separated from DC supply lines, RF lines, and sensitive fast-clock lines and circuitry. Run the lines with ground on the adjacent layer to improve the return path and isolation. Figure 11 shows the routing for the audio interface.

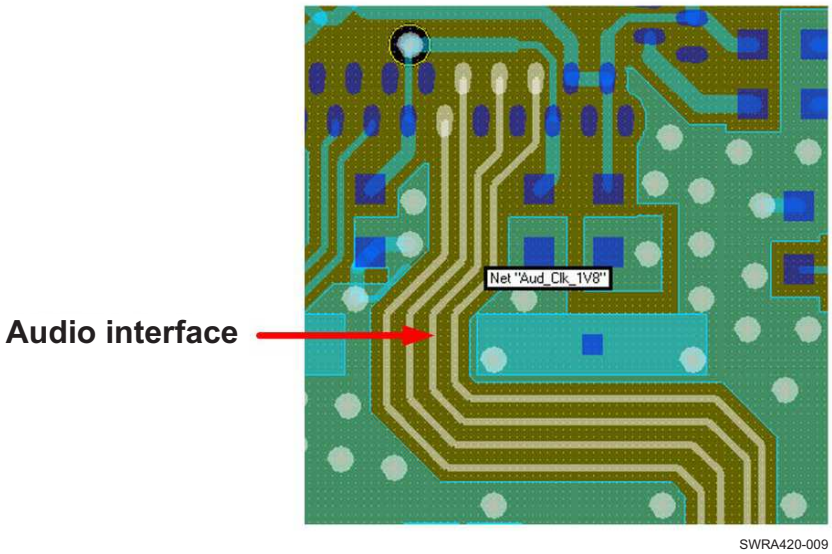

Figure 11. Audio Interface Routing

4.3.2 Universal Asynchronous Receiver-Transmitter (UART) Signal Lines

The UART used for the CC256x is a 4-wire UART connection (H4) with hardware flow control (HCI_TX, HCI_RX, HCI_RTS, and HCI_CTS). The default baud rate is 115.2 kbps; the device supports up to 4 Mbps.

4.4 Power Supply

Adhere to the following requirements to supply power to the CC256x QFN device:

- Star pattern format to supply power to the different pads of the device
- Typical width of about 10 mils (.010 inch or 0.254 mm) for wide traces
- Short power supply trace length
- Decoupling capacitor placed as close as possible to the device

The most common ground return issues occur when ground returning currents have a longer return path due to placement of the DC bypass capacitor to ground.

4.5 Ground

The common ground is the solid ground plane of layer 2 (see Figure 12).

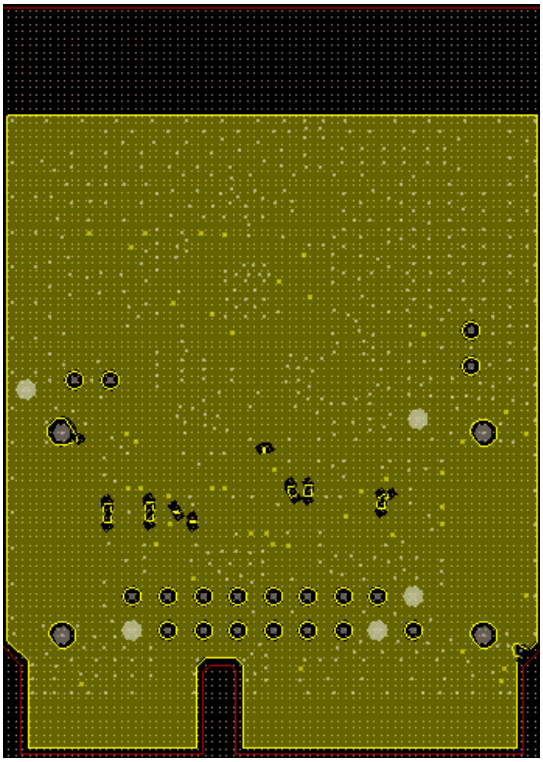

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Figure 12. Solid Ground Plane (Layer 2)

4.5.1 Key VSS Ball

To optimize isolation from digital noise, avoid connecting ball B3 (VSS_FREF) to the system-on-a-chip (SoC) thermal ground. Instead, connect B3 to solid ground (layer 2) directly using a via, as shown in Figure 13.

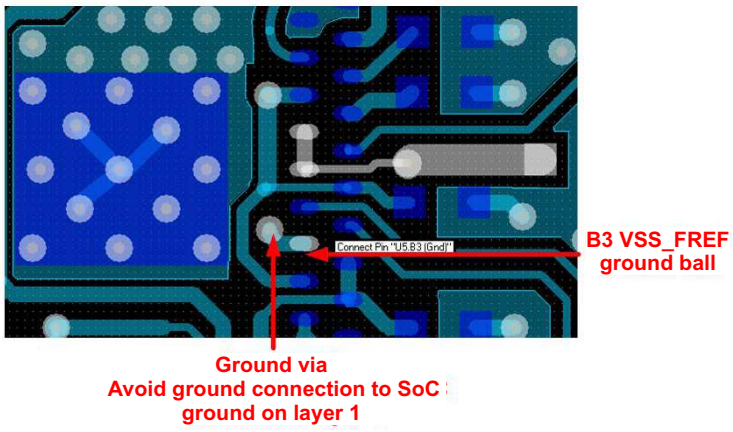

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Figure 13. VSS_FREF Instruction

4.5.2 Thermal Pad Vias

To increase ground coupling, add at least 13 vias to the thermal pads of the SoC directly to the solid ground (see Figure 14).

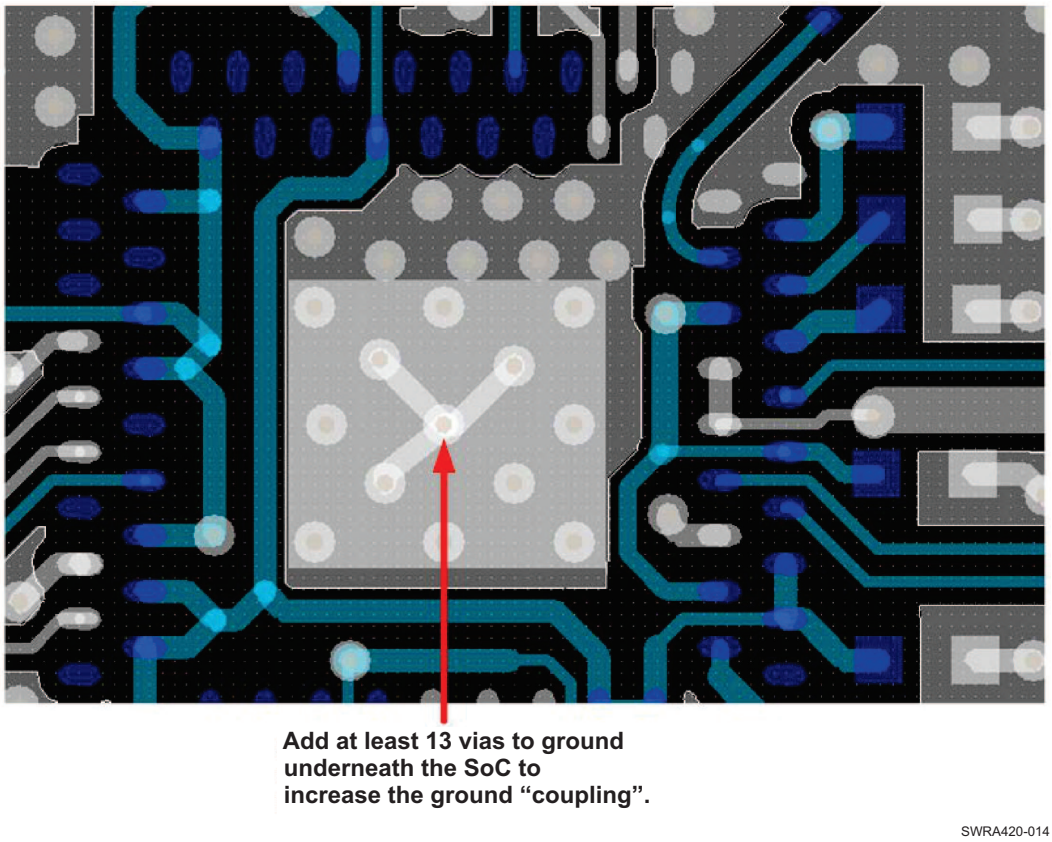

Figure 14. Thermal Pad Vias

5 QFN Package Information

Figure 15 and Figure 16 show the QFN package dimensions and PCB footprint. For more information regarding the QFN mechanical data, see the CC256x QFN datasheet (SWRS121).

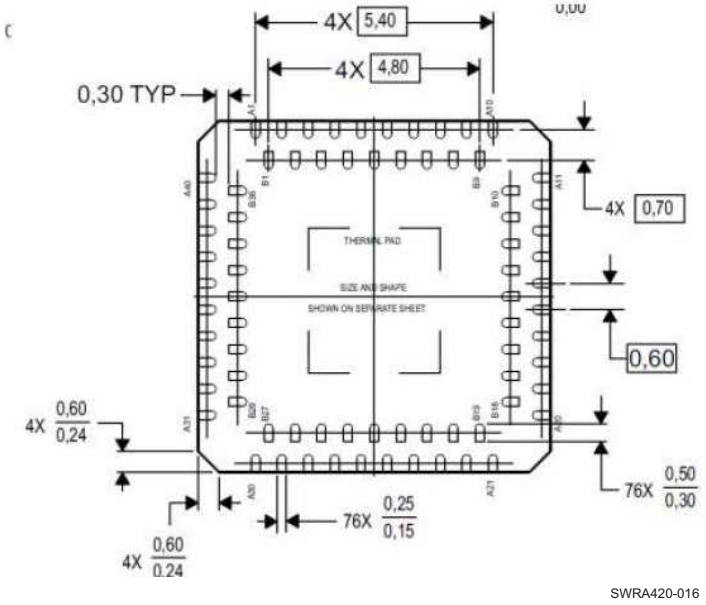

Figure 15. QFN Package Dimensions

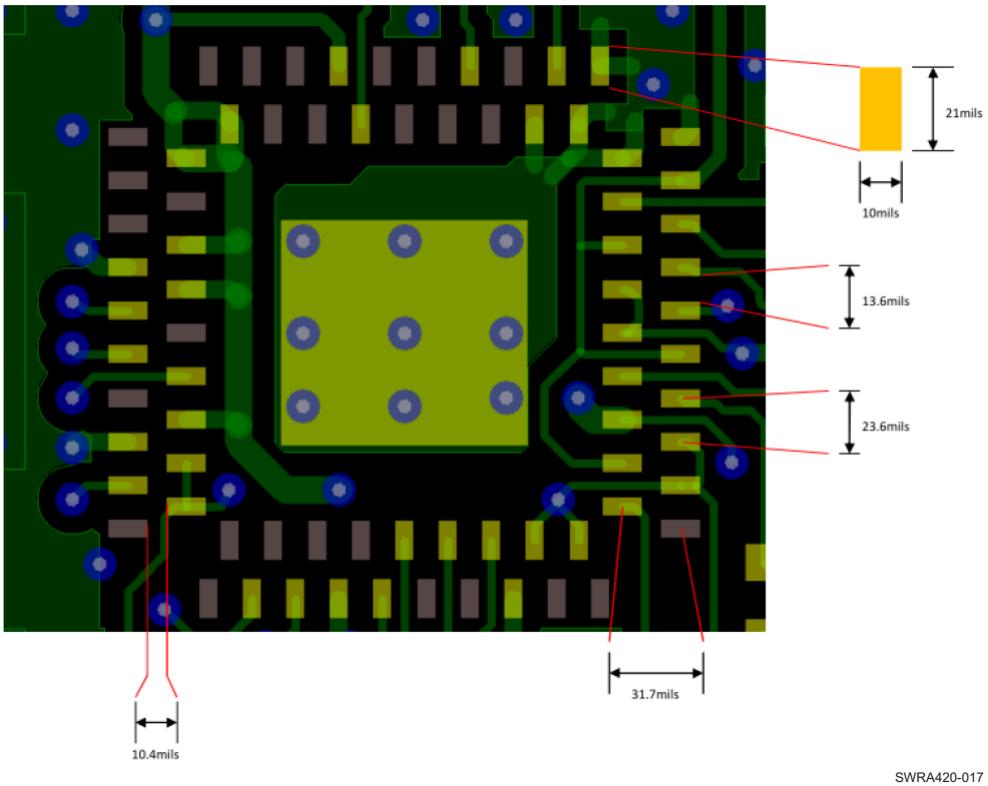

Figure 16. PCB Footprint

6 CC256x QFN EM Reference Design – PCB Layout and Assembly Drawings

Figure 17 through Figure 22 show the PCB layers and assembly drawings for the CC256x QFN EM reference design board.

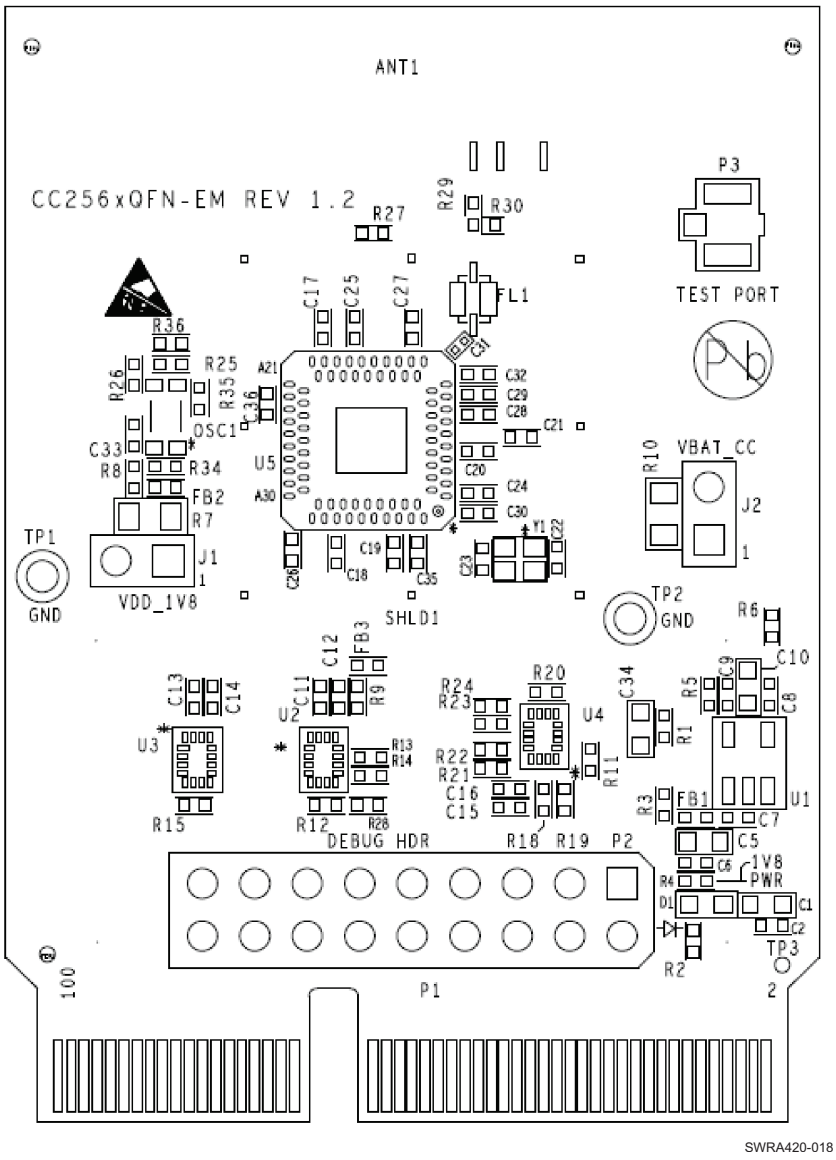

Figure 17. CC256x QFN EM Board – Top Side

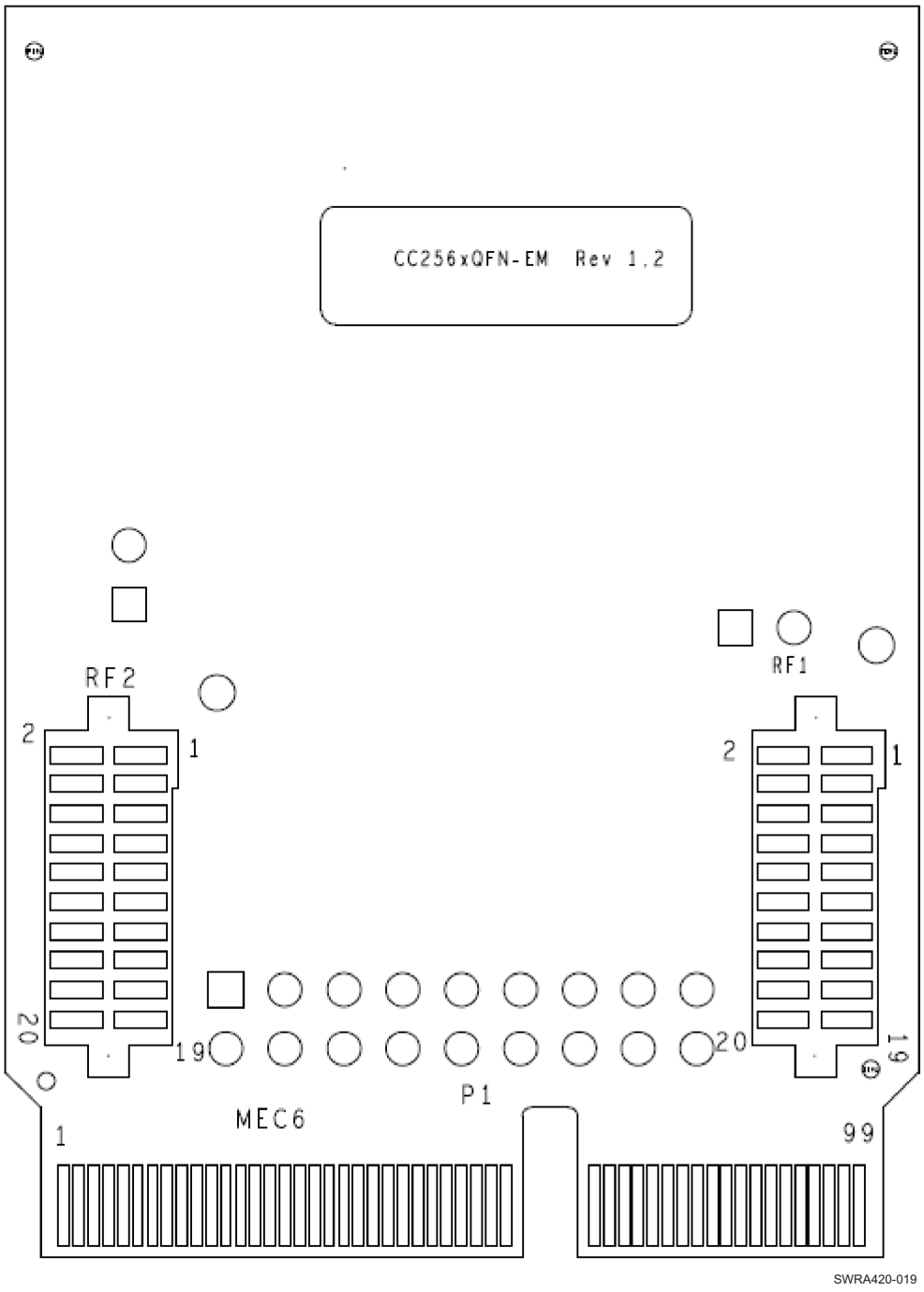

Figure 18. CC256x QFN EM Board – Bottom Side Mirror

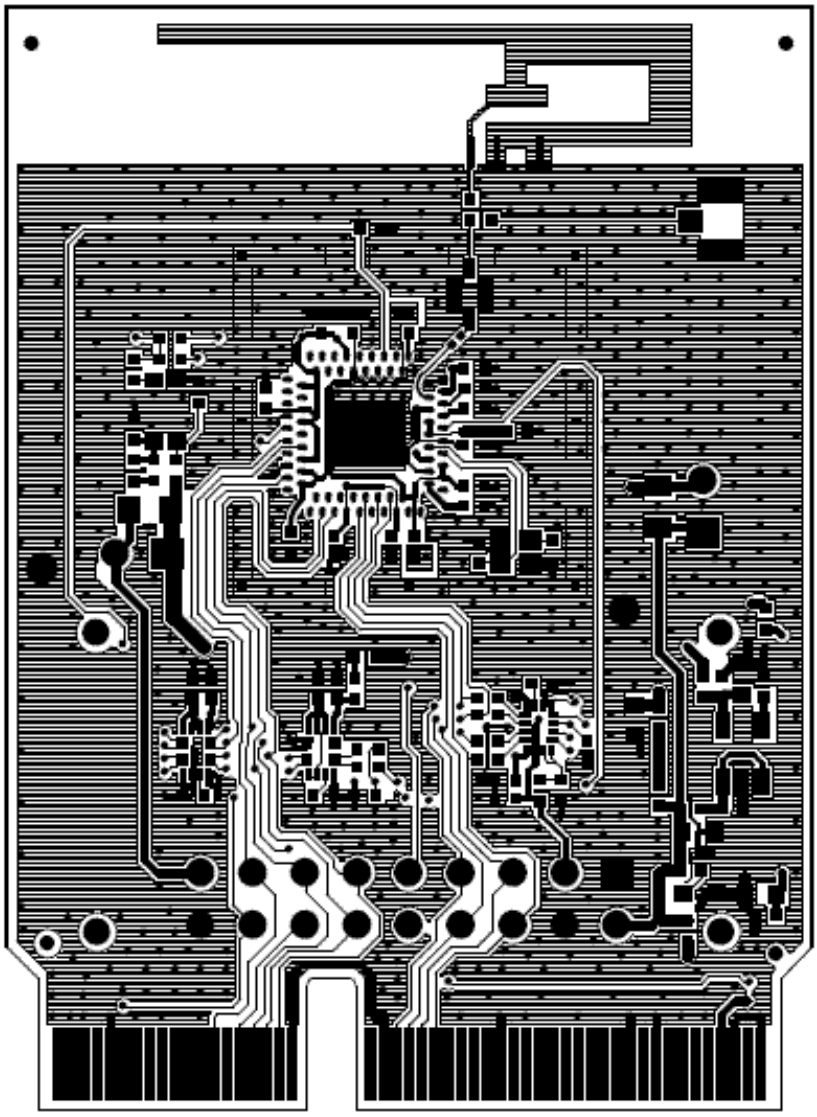

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Figure 19. CC256x QFN EM Board - Layer 1

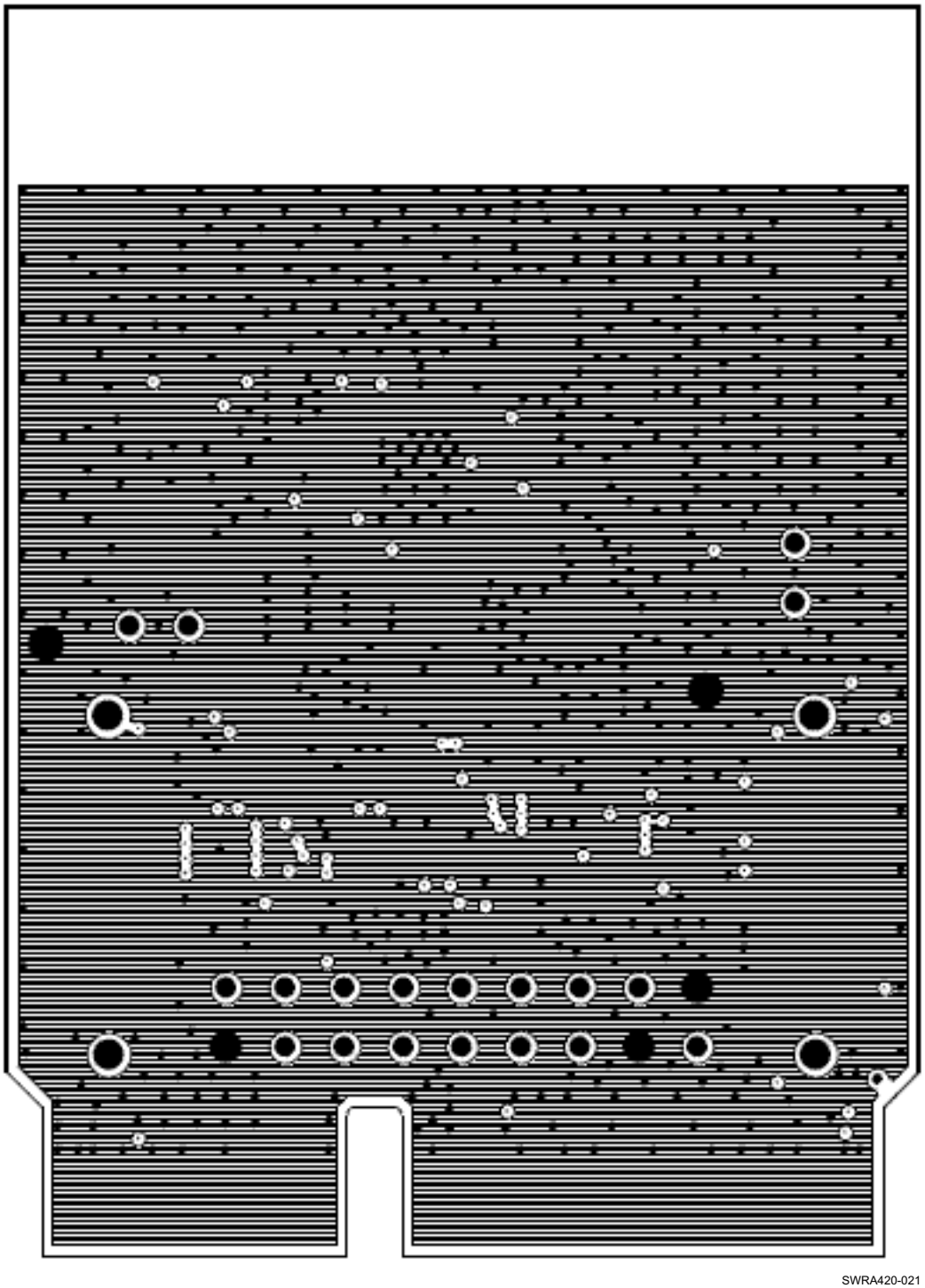

Figure 20. CC256x QFN EM Board - Layer 2

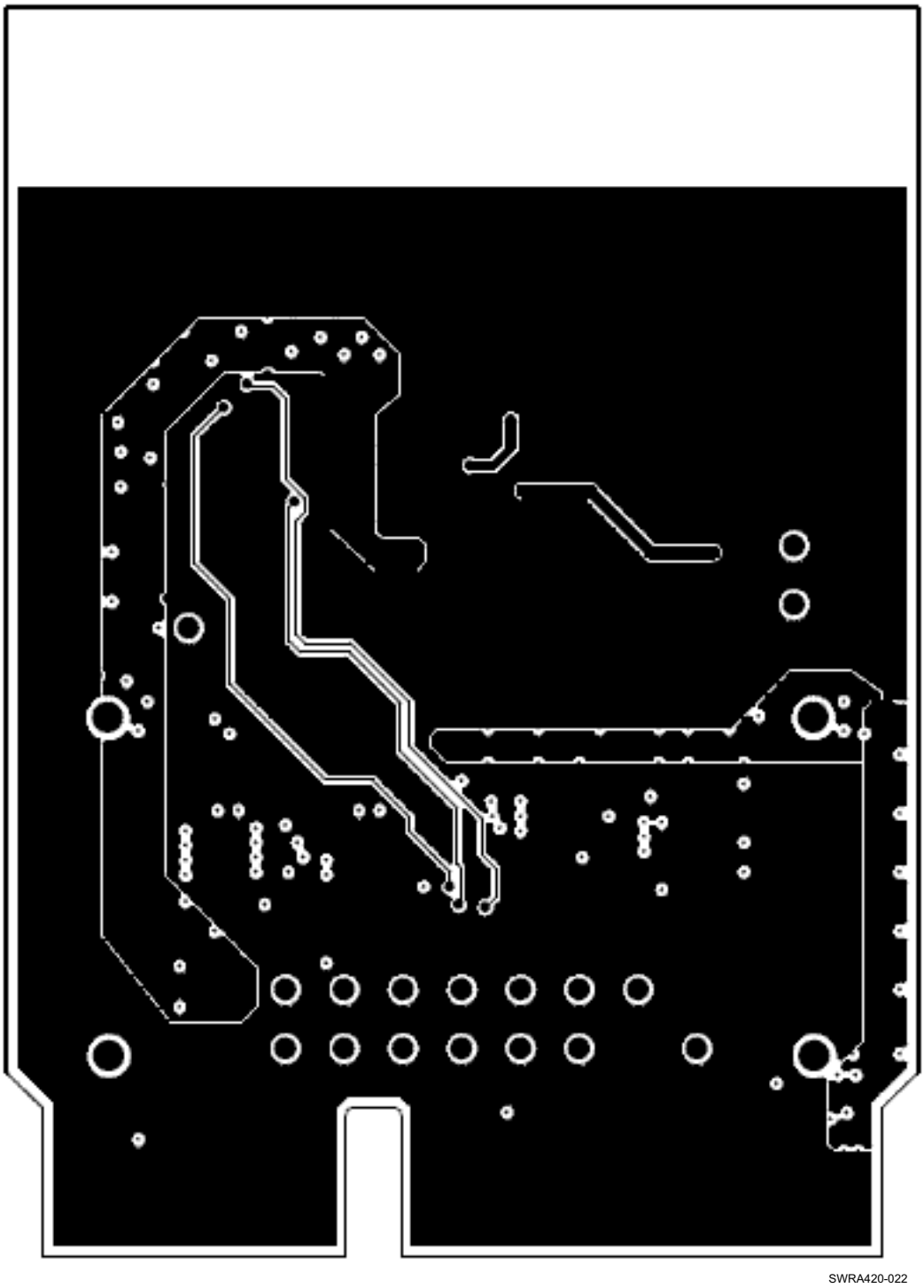

Figure 21. CC256x QFN EM Board - Layer 3

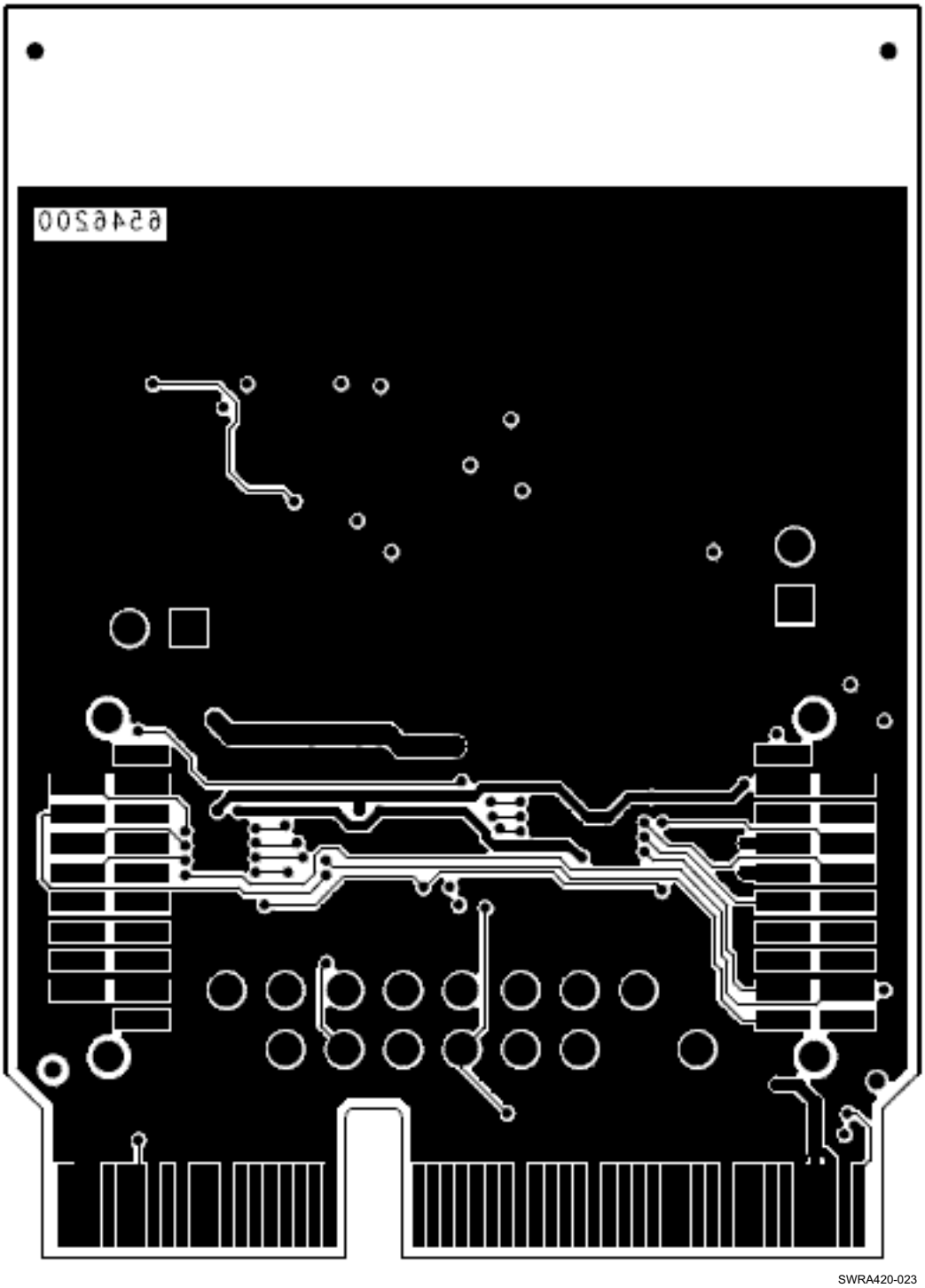

Figure 22. CC256x QFN EM Board - Layer 4

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