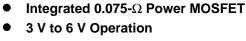


# UCC2918/81510 LOW ON-RESISTANCE HOT SWAP POWER MANAGER

SLUS457C - JULY 2000 - REVISED APRIL 2001



- External Analog Control of Fault Current From 0 A to 4 A
- Independent Analog Control of Current Limit up to 5 A
- Fast Overload Protection
- Unidirectional Switch
- Minimal External Components
- 1-μA I<sub>CC</sub> When Disabled
- Programmable On Time
- Programmable Start Delay
- Fixed 3% Duty Cycle

#### (TOP VIEW) VIN $\square$ 16 ☐ VOUT 2 VIN 15 □ VOUT 14 ☐ VOUT VIN $\Box$ 13 GND\* $\Box$ ☐ GND\* 5 12 GND\* ☐ GND\* 6 11 □ NC **FAULT** 10 SHTDWN □ ст **IFAULT** 9 □ IMAX

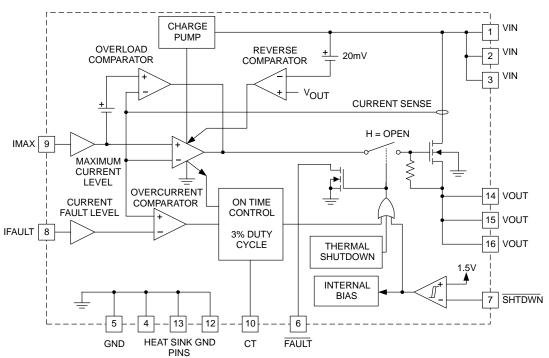
DP PACKAGE

Pin 5 serves as the lowest impedance to the electrical ground. Pins 4, 12, and 13, serve as heat sink/ground. These pins should be connected to large etch PCB areas to help dissipate heat.

# description

The UCC2918 low on-resistance hot swap power manager provides complete power management, hot swap capability, and circuit breaker functions. The only components needed to operate the device, other than supply bypassing, are a timing capacitor and two programming resistors. All control and housekeeping functions are integrated and externally programmable. These include the fault current level, maximum output sourcing current, maximum fault time, and startup delay. In the event of a constant fault, the internal fixed 3% duty cycle ratio limits the average output power. The IFAULT pin allows linear programming of the fault level current from 0 A to 4 A.

# functional block diagram





Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.



UDG-00101

# UCC2918/81510 LOW ON-RESISTANCE HOT SWAP POWER MANAGER

SLUS457C - JULY 2000 - REVISED APRIL 2001

# description (continued)

Fast overload protection is accomplished by an additional overload comparator. Its threshold is internally set above the maximum sourcing current limit setting. In the event of a short circuit or extreme current condition, this comparator is tripped, shutting down the output. This function is needed since the maximum sourcing current limit loop has a finite bandwidth.

When the output current is below the fault level, the output MOSFET is switched on with a nominal resistance of 0.075  $\Omega$ . When the output current exceeds the fault level or the maximum sourcing level, the output remains on, but the fault timer starts charging a capacitor connected to the CT pin (C<sub>T</sub>). Once C<sub>T</sub> charges to a preset threshold, the switch is turned off, and remains off for 30 times the programmed fault time. When the output current reaches the maximum sourcing level, the MOSFET transitions from a switch to a constant current source.

The UCC2918 is designed for unidirectional current flow, emulating an ideal diode in series with the power switch. This feature is particularly attractive in applications where many devices are powering a common bus, such as with SCSI termintation power (Termpwr). The UCC2918 can also be put into the sleep mode, drawing only 1  $\mu$ A of supply current.

Other features include an open-drain fault output indicator, thermal shutdown, undervoltage lockout, 3 V to 6 V operation, and a low thermal resistance small-outline power package.

# absolute maximum ratings over operating free-air temperature (unless otherwise noted)†‡

Input voltage	8 V
SOIC power dissipation	
Fault output sink current	50 mA
Fault output voltage	VIN
Output current (dc)	Internally Limited
Input voltage SHTDWN, IFAULT, IMAX	–0.3 V to VIN
Storage temperature range, T <sub>stq</sub>	–65°C to 150°C
Operating virtual junction temperature, T <sub>J</sub>	–55°C to 150°C
Lead temperature (soldering, 10 seconds)	300°C

<sup>†</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.



<sup>‡</sup> Unless otherwise indicated, voltages are reference to ground and currents are positive into, negative out of the specified terminal. Pulsed is defined as a less than 10% duty cycle with a maximum duration of 500 μs. Consult *Packaging Section* of Databook for thermal limitations and considerations of package.

# electrical characteristics at T<sub>A</sub> = -40°C to 85°C, VIN = 5 V, R<sub>IMAX</sub> = 42.2 kΩ, R<sub>IFAULT</sub> = 52.3 kΩ, SHTDWN = 2.4 V, C<sub>T</sub> = 0.1 μF, T<sub>A</sub> = T<sub>J</sub> (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNITS
Supply Section					
Voltage input range, VIN		3	5	6	V
VDD supply current	No load		1	2	mA
Sleep mode current	SHTDWN = 0.2 V		0.5	5	μΑ
Output Section					
	$I_{OUT} = 1 \text{ A to } 4 \text{ A},  VIN = 5 \text{ V},  T_{A} = 25^{\circ}\text{C}$		0.075	0.095	Ω
	$I_{OUT} = 1 \text{ A to } 4 \text{ A},  VIN = 3 \text{ V},  T_{A} = 25^{\circ}\text{C}$		0.09	0.116	Ω
RDS(on)	I <sub>OUT</sub> = 1 A to 4 A, VIN = 5 V		0.075	0.125	Ω
	I <sub>OUT</sub> = 1 A to 4 A, VIN = 3 V		0.09	0.154	Ω
Reverse leakage current	SHTDWN = 0 V, VIN = 0 V V <sub>OUT</sub> = 5 V			20	μΑ
Initial start-up time	See Note 1		100		μs
Thermal shutdown	See Note 1		170		°C
Thermal hysteresis	See Note 1		10		°C
Output leakage	SHTDWN = 0.2 V			20	μΑ
	RIFAULT = 105 kΩ	0.75	1	1.25	Α
Tita comment	$R_{IFAULT} = 52.3 \text{ k}\Omega$	1.7	2	2.3	Α
Trip current	$R_{IFAULT} = 34.8 \text{ k}\Omega$	2.5	3	3.5	Α
	$R_{IFAULT} = 25.5 \text{ k}\Omega$	3.3	4	4.7	Α
	$R_{\text{IMAX}} = 118 \text{ k}\Omega$	0.3	1	1.7	Α
	$R_{\text{IMAX}} = 60.4 \text{ k}\Omega$	1	2	3	Α
Maximum output current	$R_{\text{IMAX}} = 42.2 \text{ k}\Omega$	2	3	4	Α
	$R_{\text{IMAX}} = 33.2 \text{ k}\Omega$	2.5	3.8	5.1	Α
	$R_{\text{IMAX}} = 27.4 \text{ k}\Omega$	3.0	4.6	6.2	Α
Fault Section					
C <sub>T</sub> charge current	V <sub>CT</sub> = 1 V	-50	-36	-22	μΑ
C <sub>T</sub> discharge current	V <sub>CT</sub> = 1 V	0.5	1.2	2.0	μΑ
Output duty cycle	VOUT = 0 V	1.5	3	6	%
C <sub>T</sub> fault threshold		0.8	1.3	1.8	V
C <sub>T</sub> reset threshold		0.25	0.5	0.75	V
Shutdown Section					
Shutdown threshold		1.1	1.5	2.0	V
Shutdown hysteresis			100		mV
Input low current	SHTDWN = 0 V	-500	0	500	nA
Input high current	SHTDWN = 2 V	-2	-1	-0.5	μΑ
Open Drain Fault Output Section					
High level output current				1	μΑ
Low level output voltage	I <sub>OUT</sub> = 1 mA		0.4	0.9	V

NOTE 1: Ensured by design. Not production tested.



SLUS457C - JULY 2000 - REVISED APRIL 2001

# pin descriptions

**CT:** A capacitor connected to this pin sets the maximum fault time. The maximum time must be greater than the time to charge external load capacitance. The nominal fault time is defined as:

$$T_{\text{FAULT}} = 22.2 \times 10^3 \times C_{\text{T}} \tag{1}$$

Once the fault time is reached, the output shuts down for a time given by:

$$T_{SD} = 0.667 \times 10^6 \times C_T$$
 (2)

This equates to a 3% duty cycle. The recommended minimum value for the C<sub>T</sub> capacitor is 0.1 μF.

**FAULT:** Open-drain output, which pulls low on any condition that causes the output to open; fault, thermal shutdown, shutdown, and maximum sourcing current greater than the fault time.

**GND:** This is the most negative voltage in the circuit. All 4 ground pins should be used, and properly heat sunk on the PCB.

**IFAULT:** A resistor connected from this pin to ground sets the fault threshold. The resistor versus fault current is set by the formula:

$$R_{\mathsf{FAULT}} = \frac{105 \, \mathsf{k}\Omega}{\mathsf{I}_{\mathsf{TRIP}}} \tag{3}$$

**IMAX:** A resistor connected from this pin to ground sets the maximum sourcing current. The resistor vs the output sourcing current is set by the formula:

$$R_{\text{IMAX}} = \frac{126 \text{ k}\Omega}{\text{Maximum Sourcing Current}}$$
(4)

**SHTDWN**: When this pin is brought low, the IC is put into sleep mode. The input threshold is hysteretic, allowing the user to program a startup delay with an external RC circuit.

**VIN:** This is the input voltage to the UCC2918. The recommended operating voltage range is 3V to 6V. All VIN pins should be connected together and to the power source.

**VOUT:** Output voltage for the circuit breaker. When switched the output voltage will be approximately:

$$V_{OUT} = V_{IN} - 0.075 \Omega \times I_{OUT}. \tag{5}$$

All VOUT pins should be connected together and to the load.



#### APPLICATION INFORMATION

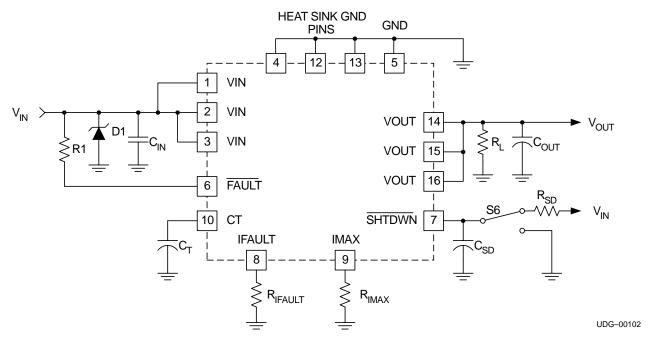


Figure 1. Typical Application

# protecting the UCC2918 from voltage transients

The parasitic inductance associated with the power distribution can cause a voltage spike at  $V_{IN}$  if the load current is suddenly interrupted by the UCC2918. It is important to limit the peak of this spike to less than 6 V to prevent damage to the UCC2918. This voltage spike can be minimized by:

- Reducing the power distribution inductance (e.g., twist the positive + and negative leads of the power supply feeding V<sub>IN</sub>, locate the power supply close to the UCC2918 or use a PCB ground plane).
- Decoupling  $V_{IN}$  with a capacitor,  $C_{IN}$  (refer to Figure 1), located close to the VIN pin. This capacitor is typically less than 1  $\mu$ F to limit the inrush current.
- Clamping the voltage at V<sub>IN</sub> below 6 V with a Zener diode, D1 (refer to Figure 1), located close to the VIN pin.

# estimating maximum load capacitance

For circuit breaker applications, the rate at which the total output capacitance can be charged depends on the maximum output current available and the nature of the load. For a constant-current current-limited circuit breaker, the output comes up if the load requires less than the maximum available short-circuit current.

To ensure recovery of a duty-cycle of the current-limited circuit breaker from a short-circuited load condition, there is a maximum total output capacitance that can be charged for a given unit ON time (fault time). The design value of ON or fault time can be adjusted by changing the timing capacitor  $C_T$ .

#### **APPLICATION INFORMATION**

# estimating maximum load capacitance

For worst-case constant-current load of value just less than the trip limit, C<sub>OUT(max)</sub> can be estimated from:

$$C_{OUT(max)} \approx \left(I_{MAX} - I_{LOAD}\right) \left(\frac{22 \times 10^3 \times C_T}{V_{OUT}}\right)$$
 (6)

Where V<sub>OUT</sub> is the output voltage and I<sub>MAX</sub> is the maximum sourcing current.

For a resistive load of value R<sub>LOAD</sub>, the value of C<sub>OUT(max)</sub> can be estimated from:

$$C_{OUT(max)} \approx \frac{22 \times 10^{3} \times C_{T}}{R_{LOAD} \times \ell n \left(\frac{1}{1 - \frac{V_{OUT}}{I_{MAX} \times R_{LOAD}}}\right)}$$
(7)

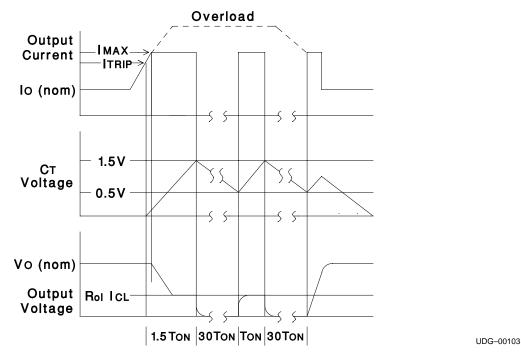


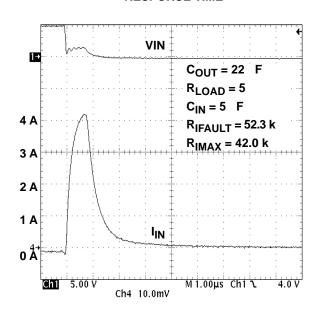
Figure 2. Load Curent, Timing Capacitor Voltage, and Output Voltage of the UCC2918 Under Fault



#### TYPICAL CHARACTERISTICS

# REVERSE VOLTAGE COMPARATOR RESPONSE TIME

#### **FAULT TIMING WAVEFORMS**



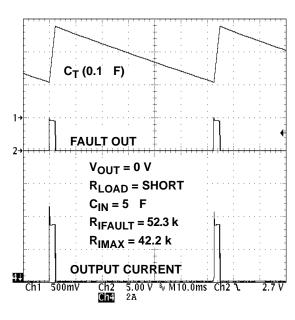


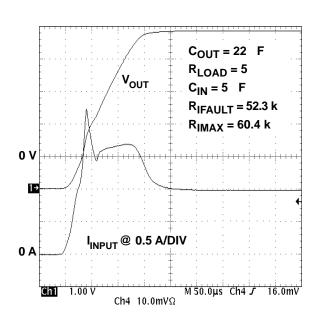
Figure 3

Figure 4

NOTE: In Figure 3 the input driven with a pulse generator shows C<sub>OUT</sub> discharging through R<sub>L</sub> and conducting through UCC81510 FET in the reverse direction.

#### **INRUSH CURRENT LIMITING**

# FAULT AND OUTPUT TURN-OFF DELAY FROM CT FAULT THRESHOLD



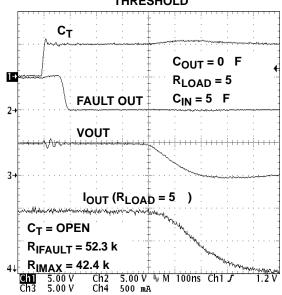


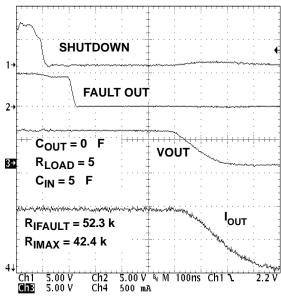
Figure 5

Figure 6

NOTE: In Figure 5 the input is switched on through the external FET. VOLIT shows IMAX linear amplifier limiting the changing current to COLIT.

#### TYPICAL CHARACTERISTICS

# **PROPAGATION DELAY** SHUTDOWN TO FAULT AND **OUTPUT RAMP-DOWN**



# PROPAGATION DELAY **ENABLE TO FAULT AND OUTPUT RAMP-UP**

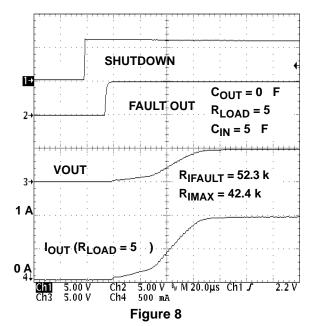




Figure 7

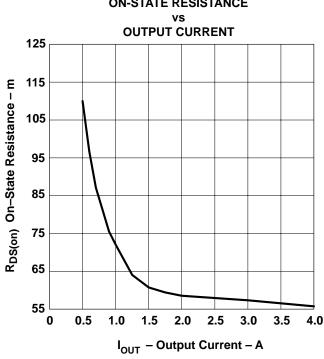


Figure 9

#### **ON-STATE RESISTANCE** vs

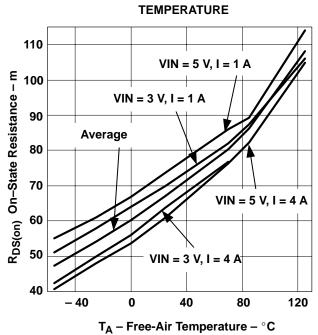


Figure 10

# UCC2918/81510 LOW ON-RESISTANCE HOT SWAP POWER MANAGER

SLUS457C - JULY 2000 - REVISED APRIL 2001

### **APPLICATION INFORMATION**

# safety considerations

Although the UCC2918 is designed to provide system protection for all fault conditions, all integrated circuits can ultimately fall short. For this reason, if the UCC2918 is intended for use in safety critical applications where UL© or some other safety rating is required, a redundant safety device such as a fuse should be placed in series with the power device. The UCC2918 prevents the fuse from blowing for virtually all fault conditions, increasing system reliability and reducing maintenance cost, in addition to providing the hot swap benefits of the device.



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PRODUCT SUPPORT: DEVELOPMENT TOOLS

### UCC2918, 0-4A, 3-6V Low RDSON Single Hot-Swap IC Hi-Side MOSFET, Industrial Temp.

DEVICE STATUS: ACTIVE

Part Number Comments

#### **Recommended Replacement Devices**

I art Number	comments						
<u>UCC3918</u>	The device has SIMILAR FUNCTIONALITY but is not functionally device., UCC2918DB/81510 has a limit on the CT cap of $>0.1$ ul						
PARAMETER NA	ME	UCC2918 UCC3918					
Number of FETs	1	1	1				
Enable/Shutdov	vn	1L	1L				
Ramp		Current	Current				
Vin (max) (V)		6	6				
Vin (min) (V)		3	3				
Current Limit Ty	уре	Adjustable	Adjustable				
Circuit Breaker	Function	Fixed	Fixed				
rDS(on) per FE	Γ (typ) (mOhms)	75	75				
Iq (typ) (uA)		100	100				
Over Current Re	eporting	Yes	Yes				
Over Temp Rep	orting	Yes	Yes				
UVLO		No	No				
Average Power	Limiting	Fixed 3% Duty Cycle	Fixed 3% Duty Cycle				
Nemko Complia	nt	No	No				
UL-Recognized		No	No				
Overcurrent Transient Filter		Programmable	Programmable				
Start Delay		Programmable	Programmable				
Automatic Retry	7	Yes	Yes				

• Integrated 0.075-  $\Omega$  Power MOSFET

• 3 V to 6 V Operation

**FEATURES** 

- External Analog Control of Fault Current From 0 A to 4 A
- Independent Analog Control of Current Limit up to 5 A
- Fast Overload Protection
- Unidirectional Switch
- Minimal External Components
- 1-uA I<sub>CC</sub> When Disabled
- Programmable On Time
- Programmable Start Delay
- Fixed 3% Duty Cycle

Back to Top

Product Folder: UCC2918, 0-4A, 3-6V Low RDSON Single Hot-Swap IC Hi-Side MOSFET, Industrial Temp.

DESCRIPTION Back to Top

The UCC2918 low on-resistance hot swap power manager provides complete power management, hot swap capability, and circuit breaker functions. The only components needed to operate the device, other than supply bypassing, are a timing capacitor and two programming resistors. All control and housekeeping functions are integrated and externally programmable. These include the fault current level, maximum output sourcing current, maximum fault time, and startup delay. In the event of a constant fault, the internal fixed 3% duty cycle ratio limits the average output power. The IFAULT pin allows linear programming of the fault level current from 0 A to 4 A.

#### TECHNICAL RESOURCES

Back to Top

To view the following documents, Acrobat Reader 4.0 is required.

To download a document to your hard drive, right-click on the link and choose 'Save'.

**DATASHEET** Back to Top

Full datasheet in Acrobat PDF: ucc2918.pdf (182 KB, Rev. C) (Updated: 03/28/2001)

**APPLICATION NOTES** 

Back to Top

Comparing Performance of Current Ramp and Voltage Ramp Hot Swap Controller ICs (SLVA103 - Updated: 07/20/2001)

SAMPLES	<u> ▲Back to Top</u>				
ORDERABLE DEVICE	<u>PACKAGE</u>	PINS	TEMP (°C)	<u>STATUS</u>	<u>SAMPLES</u>
UCC2918DP/81510	<u>D</u>	16	-40 TO 85	ACTIVE	Request Samples

PRICING/AVAILABII		<u> ▲Back to Top</u>					
ORDERABLE DEVICE	PACKAGE	PINS	TEMP (°C)	<u>STATUS</u>	BUDGETARY PRICE USS/UNIT QTY= 1000+	PACK QTY	PRICING/AVAILABILITY/PKG
UCC2918DP	<u>D</u>	16	-40 TO 85	OBSOLETE			
UCC2918DP/81510	<u>D</u>	16	-40 TO 85	ACTIVE	3.64	40	Check stock or order
UCC2918DPTR	<u>D</u>	16	-40 TO 85	OBSOLETE			
UCC2918J	<u>UTR</u>	16	-40 TO 85	OBSOLETE			
UCC2918N	<u>N</u>	16	-40 TO 85	OBSOLETE			
UCC2918PWP	<u>PW</u>	24	-40 TO 85	OBSOLETE			
UCC2918PWPTR	<u>PW</u>	24	-40 TO 85	OBSOLETE			
UCC2918QP	<u>UTR</u>	28	-40 TO 85	OBSOLETE			

DEVELOPMENT TOOLS		<u> ▲Back to Top</u>		
Tool Part Number	Tool Title		Tool Type	
UCC3918EVM	UCC3918EV	M Hot Swap Power Manager Evaluation Module	Evaluation Modules (EVM)	

Table Data Updated on: 4/14/2002