# Programmable Precision References

The TL431, A, B integrated circuits are three–terminal programmable shunt regulator diodes. These monolithic IC voltage references operate as a low temperature coefficient zener which is programmable from  $V_{ref}$  to 36 V with two external resistors. These devices exhibit a wide operating current range of 1.0 mA to 100 mA with a typical dynamic impedance of 0.22  $\Omega$ . The characteristics of these references make them excellent replacements for zener diodes in many applications such as digital voltmeters, power supplies, and op amp circuitry. The 2.5 V reference makes it convenient to obtain a stable reference from 5.0 V logic supplies, and since the TL431, A, B operates as a shunt regulator, it can be used as either a positive or negative voltage reference.

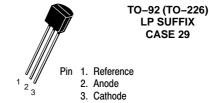
#### **Features**

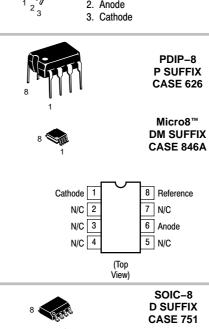
- Programmable Output Voltage to 36 V
- Voltage Reference Tolerance: ±0.4%, Typ @ 25°C (TL431B)
- Low Dynamic Output Impedance, 0.22 Ω Typical
- Sink Current Capability of 1.0 mA to 100 mA
- Equivalent Full–Range Temperature Coefficient of 50 ppm/°C Typical
- Temperature Compensated for Operation over Full Rated Operating Temperature Range
- Low Output Noise Voltage
- Pb-Free Packages are Available

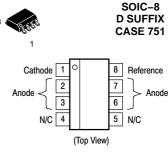


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This is an internally modified SOIC–8 package. Pins 2, 3, 6 and 7 are electrically common to the die attach flag. This internal lead frame modification increases power dissipation capability when appropriately mounted on a printed circuit board. This modified package conforms to all external dimensions of the standard SOIC–8 package.

#### **ORDERING INFORMATION**

See detailed ordering and shipping information in the package dimensions section on page 13 of this data sheet.

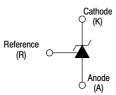
#### **DEVICE MARKING INFORMATION**

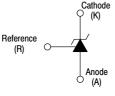
See general marking information in the device marking section on page 15 of this data sheet.

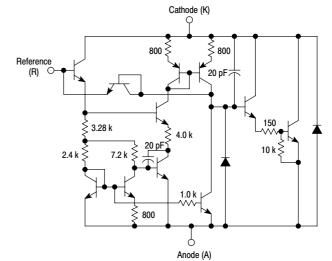
#### **Symbol**

### **Representative Schematic Diagram**

Component values are nominal







### Representative Block Diagram

Reference Cathode (R) (K) 2.5 V<sub>ref</sub>

This device contains 12 active transistors.

#### MAXIMUM RATINGS (Full operating ambient temperature range applies, unless otherwise noted.)

Rating	Symbol	Value	Unit
Cathode to Anode Voltage	V <sub>KA</sub>	37	V
Cathode Current Range, Continuous	I <sub>K</sub>	-100 to +150	mA
Reference Input Current Range, Continuous	I <sub>ref</sub>	-0.05 to +10	mA
Operating Junction Temperature	TJ	150	°C
Operating Ambient Temperature Range TL431I, TL431AI, TL431BI TL431C, TL431AC, TL431BC NCV431AI, TL431BV	T <sub>A</sub>	-40 to +85 0 to +70 -40 to +125	°C
Storage Temperature Range	T <sub>stg</sub>	-65 to +150	°C
Total Power Dissipation @ T <sub>A</sub> = 25°C  Derate above 25°C Ambient Temperature  D, LP Suffix Plastic Package P Suffix Plastic Package  DM Suffix Plastic Package	PD	0.70 1.10 0.52	W
Total Power Dissipation @ T <sub>C</sub> = 25°C  Derate above 25°C Case Temperature  D, LP Suffix Plastic Package  P Suffix Plastic Package	PD	1.5 3.0	W

Maximum ratings are those values beyond which device damage can occur. Maximum ratings applied to the device are individual stress limit values (not normal operating conditions) and are not valid simultaneously. If these limits are exceeded, device functional operation is not implied, damage may occur and reliability may be affected.

NOTE: ESD data available upon request.

#### RECOMMENDED OPERATING CONDITIONS

Condition	Symbol	Min	Max	Unit
Cathode to Anode Voltage	V <sub>KA</sub>	$V_{ref}$	36	V
Cathode Current	Ι <sub>Κ</sub>	1.0	100	mA

#### THERMAL CHARACTERISTICS

Characteristic	Symbol	D, LP Suffix Package	P Suffix Package	DM Suffix Package	Unit
Thermal Resistance, Junction-to-Ambient	$R_{\theta JA}$	178	114	240	°C/W
Thermal Resistance, Junction-to-Case	$R_{ heta JC}$	83	41	ı	°C/W

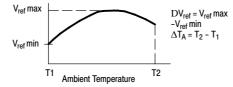
#### **ELECTRICAL CHARACTERISTICS** (T<sub>A</sub> = 25°C, unless otherwise noted.)

			TL431I			TL431C		
Characteristic	Symbol	Min	Тур	Max	Min	Тур	Max	Unit
Reference Input Voltage (Figure 1) $V_{KA} = V_{ref}, I_K = 10 \text{ mA}$ $T_A = 25^{\circ}\text{C}$ $T_A = T_{low} \text{ to } T_{high} \text{ (Note 1)}$	V <sub>ref</sub>	2.44 2.41	2.495 -	2.55 2.58	2.44 2.423	2.495 –	2.55 2.567	V
Reference Input Voltage Deviation Over Temperature Range (Figure 1, Notes 1, 2) V <sub>KA</sub> = V <sub>ref</sub> , I <sub>K</sub> = 10 mA	$\Delta V_{ref}$	_	7.0	30	-	3.0	17	mV
Ratio of Change in Reference Input Voltage to Change in Cathode to Anode Voltage $I_K = 10$ mA (Figure 2), $\Delta V_{KA} = 10$ V to $V_{ref}$ $\Delta V_{KA} = 36$ V to 10 V	$rac{\Delta V_{ m ref}}{\Delta V_{ m KA}}$	- -	-1.4 -1.0	-2.7 -2.0	- -	-1.4 -1.0	-2.7 -2.0	mV/V
Reference Input Current (Figure 2) $I_K = 10 \text{ mA, } R1 = 10 \text{ k, } R2 = \infty$ $T_A = 25^{\circ}\text{C}$ $T_A = T_{low} \text{ to } T_{high} \text{ (Note 1)}$	I <sub>ref</sub>	- -	1.8	4.0 6.5	- -	1.8 -	4.0 5.2	μΑ
Reference Input Current Deviation Over Temperature Range (Figure 2, Note 1, 4) I <sub>K</sub> = 10 mA, R1 = 10 k, R2 = ∞	$\Delta I_{ref}$	_	0.8	2.5	-	0.4	1.2	μΑ
Minimum Cathode Current For Regulation V <sub>KA</sub> = V <sub>ref</sub> (Figure 1)	I <sub>min</sub>	_	0.5	1.0	-	0.5	1.0	mA
Off–State Cathode Current (Figure 3) V <sub>KA</sub> = 36 V, V <sub>ref</sub> = 0 V	I <sub>off</sub>	-	20	1000	_	20	1000	nA
Dynamic Impedance (Figure 1, Note 3) $V_{KA} = V_{ref},  \Delta I_K = 1.0  \text{mA to } 100  \text{mA}$ $f \leq 1.0  \text{kHz}$	Z <sub>KA</sub>	-	0.22	0.5	-	0.22	0.5	Ω

= -40°C for TL431AIP TL431AIP, TL431IP, TL431IP, TL431BID, TL431BIP, TL431BIP, TL431AIDM, TL431DM, TL431BIDM; 0°C for TL431ACP, TL431ACLP, TL431CP, TL431CLP, TL431CD, TL431ACD, TL431BCD, TL431BCP, TL431BCLP, TL431CDM, TL431ACDM, TL431BCDM

 $T_{high} = +85^{\circ}C$  for TL431AIP, TL431AIP, TL431IP, TL431IP, TL431BID, TL431BIP, TL431BIP, TL431BIDM, TL431AIDM, TL431AIDM, TL431AIDM +70°C for TL431ACP, TL431ACLP, TL431CP, TL431ACD, TL431BCD, TL431BCP, TL431BCLP, TL431CDM, TL431ACDM, TL431BCDM

2. The deviation parameter  $\Delta V_{ref}$  is defined as the difference between the maximum and minimum values obtained over the full operating ambient temperature range that applies.



The average temperature coefficient of the reference input voltage,  $\alpha V_{\text{ref}}$  is defined as:

$$V_{ref} \stackrel{ppm}{\overset{\circ}{C}} = \frac{\left(\frac{\Delta V_{ref}}{V_{ref} @ 25^{\circ}C}\right) \times 10^{6}}{\Delta T_{A}} = \frac{\Delta V_{ref} \times 10^{6}}{\Delta T_{A} (V_{ref} @ 25^{\circ}C)}$$

 $\alpha V_{ref}$  can be positive or negative depending on whether  $V_{ref}$  Min or  $V_{ref}$  Max occurs at the lower ambient temperature. (Refer to Figure 6.)

Example :  $\Delta V_{ref}=8.0$  mV and slope is positive,  $V_{ref} @ 25^{\circ}C=2.495 \ V, \Delta T_{A}=70^{\circ}C$ 

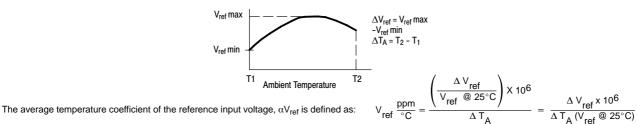
$$V_{ref} = 8.0 \text{ mV}$$
 and slope is positive,  
 $V_{ref} @ 25^{\circ}\text{C} = 2.495 \text{ V}, \Delta T_{A} = 70^{\circ}\text{C}$   $\alpha V_{ref} = \frac{0.008 \times 10^{6}}{70 (2.495)} = 45.8 \text{ ppm/}^{\circ}\text{C}$ 

3. The dynamic impedance  $Z_{KA}$  is defined as:  $|Z_{KA}| = \frac{\Delta V_{KA}}{\Delta I_{K}}$ . When the device is programmed with two external resistors, R1 and R2, (refer to Figure 2) the total dynamic impedance of the circuit is defined as:  $|Z_{KA}'| \approx |Z_{KA}| \left(1 + \frac{R1}{R2}\right)$ 

### **ELECTRICAL CHARACTERISTICS** ( $T_A = 25^{\circ}C$ , unless otherwise noted.)

		TL431	AI / NC	/431AI	Т	L431A0	;	TL431	BI / TL4	31BV	
Characteristic	Symbol	Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Unit
Reference Input Voltage (Figure 1) $V_{KA} = V_{ref}, \ I_K = 10 \ mA$ $T_A = 25^{\circ}C$ $T_A = T_{low} \ to \ T_{high}$	V <sub>ref</sub>	2.47 2.44	2.495 –	2.52 2.55	2.47 2.453	2.495 –	2.52 2.537	2.483 2.475	2.495 2.495	2.507 2.515	V
Reference Input Voltage Deviation Over Temperature Range (Figure 1, Notes 4, 5) V <sub>KA</sub> = V <sub>ref</sub> , I <sub>K</sub> = 10 mA	$\Delta V_{ref}$	_	7.0	30	-	3.0	17	-	3.0	17	mV
Ratio of Change in Reference Input Voltage to Change in Cathode to Anode Voltage $I_K = 10$ mA (Figure 2), $\Delta V_{KA} = 10$ V to $V_{ref}$ $\Delta V_{KA} = 36$ V to 10 V	$\frac{\Delta V_{ref}}{\Delta V_{KA}}$	_ _	-1.4 -1.0	-2.7 -2.0	- -	-1.4 -1.0	-2.7 -2.0	- -	-1.4 -1.0	-2.7 -2.0	mV/V
Reference Input Current (Figure 2) $I_K = 10 \text{ mA, } R1 = 10 \text{ k, } R2 = \infty$ $T_A = 25^{\circ}\text{C}$ $T_A = T_{low} \text{ to } T_{high} \text{ (Note 4)}$	I <sub>ref</sub>	- -	1.8	4.0 6.5	- -	1.8	4.0 5.2	<u>-</u>	1.1	2.0 4.0	μΑ
Reference Input Current Deviation Over Temperature Range (Figure 2, Note 4) I <sub>K</sub> = 10 mA, R1 = 10 k, R2 = ∞	$\Delta I_{ref}$	-	0.8	2.5	-	0.4	1.2	-	0.8	2.5	μΑ
Minimum Cathode Current For Regulation V <sub>KA</sub> = V <sub>ref</sub> (Figure 1)	I <sub>min</sub>	_	0.5	1.0	-	0.5	1.0	-	0.5	1.0	mA
Off–State Cathode Current (Figure 3) V <sub>KA</sub> = 36 V, V <sub>ref</sub> = 0 V	I <sub>off</sub>	_	20	1000	_	20	1000	-	0.23	500	nA
Dynamic Impedance (Figure 1, Note 6) $V_{KA} = V_{ref}$ , $\Delta I_{K} = 1.0$ mA to 100 mA $f \le 1.0$ kHz	Z <sub>KA</sub>	_	0.22	0.5	_	0.22	0.5	_	0.14	0.3	Ω

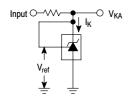
- 4.  $T_{low} = -40^{\circ}\text{C}$  for TL431AIP TL431AIP, TL431IP, TL431IP, TL431BID, TL431BIP, TL431BI TL431BIDM, NCV431AIDMR2, NCV431AIDR2
  - = 0°C for TL431ACP, TL431ACP, TL431CP, TL431CP, TL431CD, TL431ACD, TL431BCD, TL431BCP, TL431BCP, TL431CDM, TL431ACDM, TL431BCDM
  - $T_{high} = +85^{\circ}C$  for TL431AIP, TL431AIP, TL431IP, TL431IP, TL431BID, TL431BIP, TL431BIP, TL431BIDM, TL431AIDM, TL431AIDM, TL431AIDM = +70°C for TL431ACP, TL431ACLP, TL431CP, TL431ACD, TL431BCD, TL431BCP, TL431BCLP, TL431CDM, TL431ACDM, TL431BCDM
    - +125°C TL431BV, NCV431AIDMR2, NCV431AIDR2
- 5. The deviation parameter  $\Delta V_{ref}$  is defined as the difference between the maximum and minimum values obtained over the full operating ambient temperature range that applies.



 $\alpha V_{ref}$  can be positive or negative depending on whether  $V_{ref}$  Min or  $V_{ref}$  Max occurs at the lower ambient temperature. (Refer to Figure 6.)

Example : 
$$\Delta V_{ref} = 8.0$$
 mV and slope is positive, 
$$V_{ref} @ 25^{\circ}C = 2.495 \text{ V}, \Delta T_{A} = 70^{\circ}C \\ \alpha V_{ref} = \frac{0.008 \times 10^{6}}{70 \ (2.495)} = 45.8 \text{ ppm/}^{\circ}C$$

- 6. The dynamic impedance  $Z_{KA}$  is defined as  $|Z_{KA}| = \frac{\Delta V_{KA}}{\Delta I_{K}}$  When the device is programmed with two external resistors, R1 and R2, (refer
- to Figure 2) the total dynamic impedance of the circuit is defined as:  $|Z_{KA}'| \approx |Z_{KA}| \left(1 + \frac{R1}{R2}\right)$ 7. NCV431AIDMR2, NCV431AIDR2  $T_{low} = -40^{\circ}C$ ,  $T_{high} = +125^{\circ}C$ . Guaranteed by design. NCV prefix is for automotive and other applications requiring site and change control.





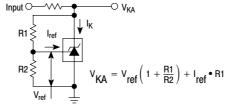


Figure 2. Test Circuit for  $V_{KA} > V_{ref}$ 

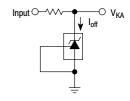


Figure 3. Test Circuit for Ioff

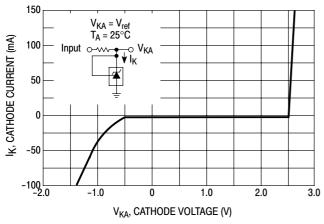


Figure 4. Cathode Current versus Cathode Voltage

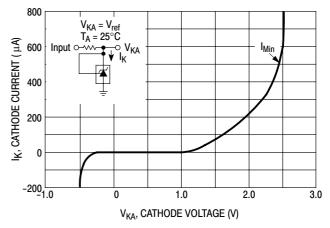


Figure 5. Cathode Current versus Cathode Voltage

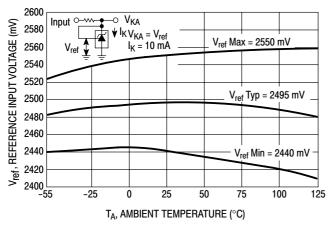


Figure 6. Reference Input Voltage versus Ambient Temperature

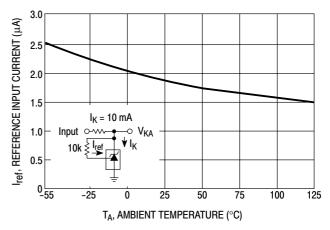


Figure 7. Reference Input Current versus
Ambient Temperature

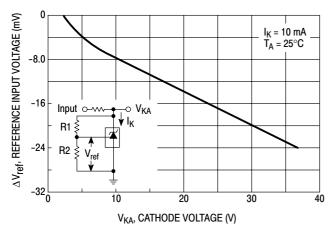


Figure 8. Change in Reference Input Voltage versus Cathode Voltage

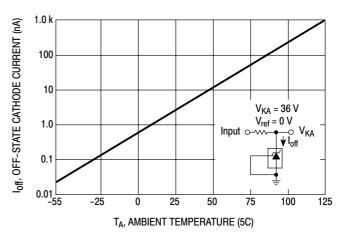


Figure 9. Off-State Cathode Current versus Ambient Temperature

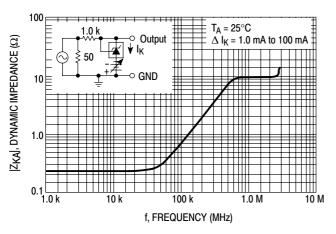


Figure 10. Dynamic Impedance versus Frequency

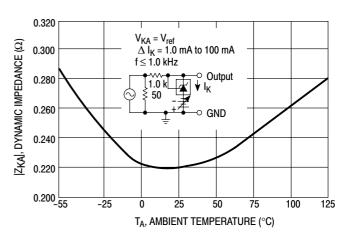


Figure 11. Dynamic Impedance versus Ambient Temperature

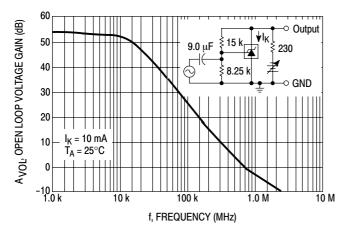


Figure 12. Open-Loop Voltage Gain versus Frequency

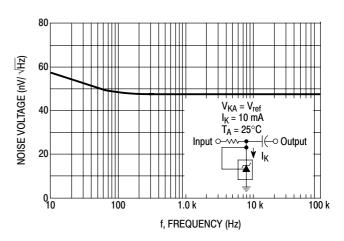


Figure 13. Spectral Noise Density

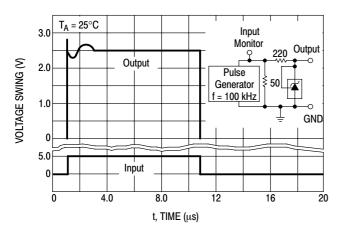


Figure 14. Pulse Response

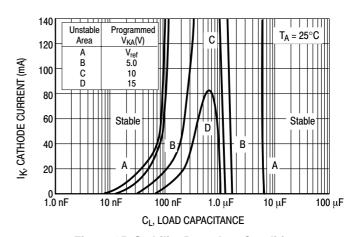


Figure 15. Stability Boundary Conditions

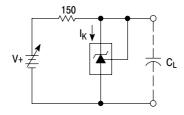


Figure 16. Test Circuit For Curve A of Stability Boundary Conditions

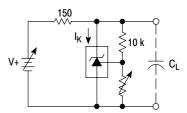


Figure 17. Test Circuit For Curves B, C, And D of Stability Boundary Conditions

#### **TYPICAL APPLICATIONS**

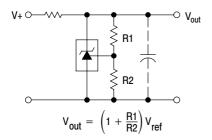


Figure 18. Shunt Regulator

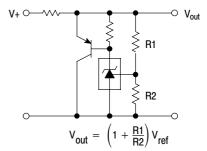


Figure 19. High Current Shunt Regulator

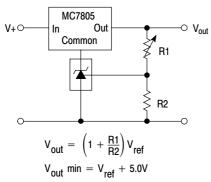


Figure 20. Output Control for a

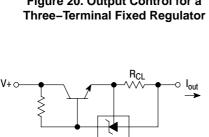


Figure 22. Constant Current Source

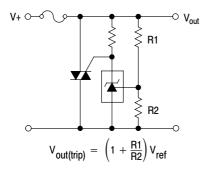


Figure 24. TRIAC Crowbar

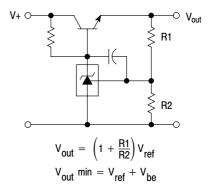


Figure 21. Series Pass Regulator

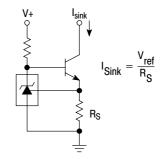


Figure 23. Constant Current Sink

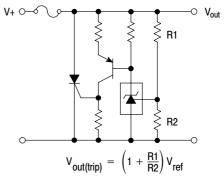
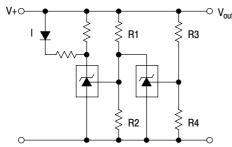


Figure 25. SRC Crowbar



L.E.D. indicator is 'on' when V+ is between the upper and lower limits.

$$\begin{aligned} \text{Lower Limit} &= \left(1 + \frac{R1}{R2}\right) V_{ref} \\ \text{Upper Limit} &= \left(1 + \frac{R3}{R4}\right) V_{ref} \end{aligned}$$

Figure 26. Voltage Monitor

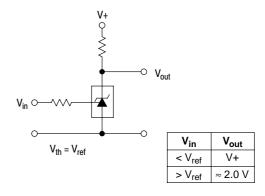


Figure 27. Single–Supply Comparator with Temperature–Compensated Threshold

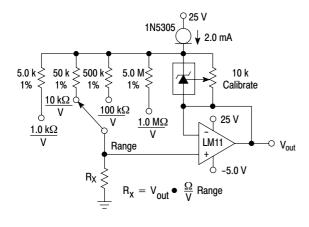


Figure 28. Linear Ohmmeter

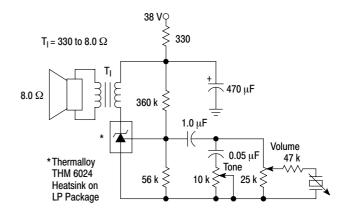


Figure 29. Simple 400 mW Phono Amplifier

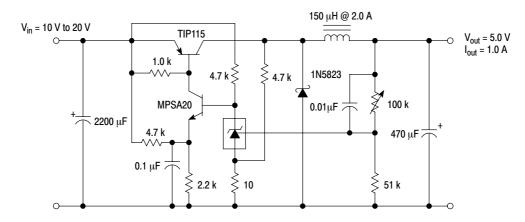


Figure 30. High Efficiency Step-Down Switching Converter

Test	Conditions	Results
Line Regulation	$V_{in} = 10 \text{ V to } 20 \text{ V}, I_0 = 1.0 \text{ A}$	53 mV (1.1%)
Load Regulation	$V_{in} = 15 \text{ V}, I_{o} = 0 \text{ A to } 1.0 \text{ A}$	25 mV (0.5%)
Output Ripple	$V_{in} = 10 \text{ V}, I_0 = 1.0 \text{ A}$	50 mVpp P.A.R.D.
Output Ripple	$V_{in} = 20 \text{ V}, I_0 = 1.0 \text{ A}$	100 mVpp P.A.R.D.
Efficiency	$V_{in} = 15 \text{ V}, I_0 = 1.0 \text{ A}$	82%

#### **APPLICATIONS INFORMATION**

The TL431 is a programmable precision reference which is used in a variety of ways. It serves as a reference voltage in circuits where a non-standard reference voltage is needed. Other uses include feedback control for driving an optocoupler in power supplies, voltage monitor, constant current source, constant current sink and series pass regulator. In each of these applications, it is critical to maintain stability of the device at various operating currents and load capacitances. In some cases the circuit designer can estimate the stabilization capacitance from the stability boundary conditions curve provided in Figure 15. However, these typical curves only provide stability information at specific cathode voltages and at a specific load condition. Additional information is needed to determine the capacitance needed to optimize phase margin or allow for process variation.

A simplified model of the TL431 is shown in Figure 31. When tested for stability boundaries, the load resistance is 150  $\Omega$ . The model reference input consists of an input transistor and a dc emitter resistance connected to the device anode. A dependent current source, Gm, develops a current whose amplitude is determined by the difference between the 1.78 V internal reference voltage source and the input transistor emitter voltage. A portion of Gm flows through compensation capacitance,  $C_{P2}$ . The voltage across  $C_{P2}$  drives the output dependent current source, Go, which is connected across the device cathode and anode.

Model component values are:

 $V_{ref} = 1.78 \text{ V}$ 

 $Gm = 0.3 + 2.7 \exp(-I_C/26 \text{ mA})$ 

where I<sub>C</sub> is the device cathode current and Gm is in mhos

Go = 1.25 (
$$V_{cp}$$
2) µmhos.

Resistor and capacitor typical values are shown on the model. Process tolerances are  $\pm 20\%$  for resistors,  $\pm 10\%$  for capacitors, and  $\pm 40\%$  for transconductances.

An examination of the device model reveals the location of circuit poles and zeroes:

P1 = 
$$\frac{1}{2\pi R_{GM} C_{P1}}$$
 =  $\frac{1}{2\pi * 1.0 M * 20 pF}$  = 7.96 kHz

$$P2 = \frac{1}{2\pi R_{p2}C_{p2}} = \frac{1}{2\pi * 10 M * 0.265 pF} = 60 \text{ kHz}$$

$$Z1 = \frac{1}{2\pi R_{71}C_{P1}} = \frac{1}{2\pi * 15.9 k * 20 pF} = 500 kHz$$

In addition, there is an external circuit pole defined by the load:

$$P_L = \frac{1}{2\pi R_I C_I}$$

Also, the transfer dc voltage gain of the TL431 is:

$$G = G_M R_{GM} GoR_L$$

Example 1:

 $\rm I_{\mbox{\scriptsize C}} = 10\,m\mbox{\scriptsize mA}, R_{\mbox{\scriptsize L}} = \,230~\Omega, C_{\mbox{\scriptsize L}} = \,0.$  Define the transfer gain .

The DC gain is:

$$G = G_M R_{GM} GoR_L =$$
(2.138)(1.0 M)(1.25  $\mu$ )(230) = 615 = 56 dB

Loop gain = 
$$G \frac{8.25 \text{ k}}{8.25 \text{ k} + 15 \text{ k}} = 218 = 47 \text{ dB}$$

The resulting transfer function Bode plot is shown in Figure 32. The asymptotic plot may be expressed as the following equation:

$$Av = 615 \frac{\left(1 + \frac{jf}{500 \text{ kHz}}\right)}{\left(1 + \frac{jf}{8.0 \text{ kHz}}\right)\left(1 + \frac{jf}{60 \text{ kHz}}\right)}$$

The Bode plot shows a unity gain crossover frequency of approximately 600 kHz. The phase margin, calculated from the equation, would be 55.9 degrees. This model matches the Open–Loop Bode Plot of Figure 12. The total loop would have a unity gain frequency of about 300 kHz with a phase margin of about 44 degrees.

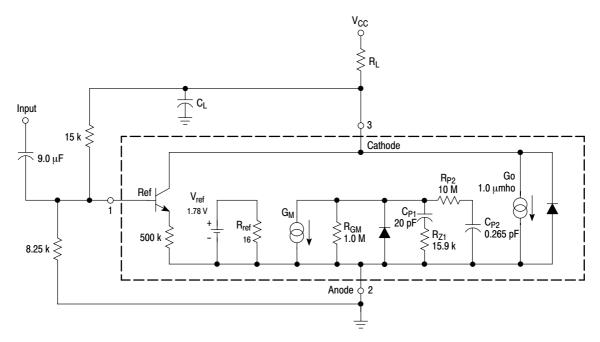


Figure 31. Simplified TL431 Device Model

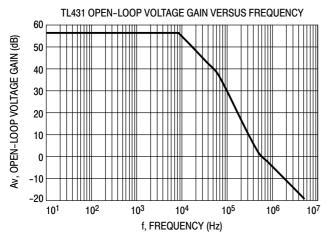


Figure 32. Example 1 Circuit Open Loop Gain Plot Example 2.

 $I_C=7.5$  mA,  $R_L=2.2$  k $\Omega$ ,  $C_L=0.01$   $\mu F$ . Cathode tied to reference input pin. An examination of the data sheet stability boundary curve (Figure 15) shows that this value of load capacitance and cathode current is on the boundary. Define the transfer gain.

The DC gain is:

$$G = G_M R_{GM} GoR_L =$$

 $(2.323)(1.0 \text{ M})(1.25 \mu)(2200) = 6389 = 76 \text{ dB}$ 

The resulting open loop Bode plot is shown in Figure 33. The asymptotic plot may be expressed as the following equation:

$$Av = 615 \frac{\left(1 + \frac{jf}{500 \text{ kHz}}\right)}{\left(1 + \frac{jf}{8.0 \text{ kHz}}\right) \left(1 + \frac{jf}{60 \text{ kHz}}\right) \left(1 + \frac{jf}{7.2 \text{ kHz}}\right)}$$

Note that the transfer function now has an extra pole formed by the load capacitance and load resistance.

Note that the crossover frequency in this case is about 250 kHz, having a phase margin of about -46 degrees. Therefore, instability of this circuit is likely.

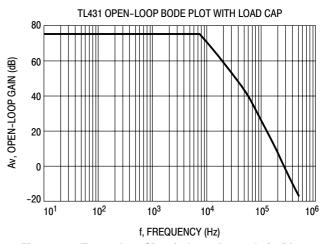


Figure 33. Example 2 Circuit Open Loop Gain Plot

With three poles, this system is unstable. The only hope for stabilizing this circuit is to add a zero. However, that can only be done by adding a series resistance to the output capacitance, which will reduce its effectiveness as a noise filter. Therefore, practically, in reference voltage applications, the best solution appears to be to use a smaller value of capacitance in low noise applications or a very large value to provide noise filtering and a dominant pole rolloff of the system.

#### **ORDERING INFORMATION**

Device	Operating Temperature Range	Package Code	Shipping Information <sup>†</sup>	Tolerance
TL431ACD		SOIC-8		
TL431ACDG	]	SOIC-8 (Pb-Free)		1.0%
TL431BCD		SOIC-8	98 Units / Rail	
TL431BCDG	]	SOIC-8 (Pb-Free)		0.4%
TL431CD		2010.0		2.2%
TL431ACDR2		SOIC-8		
TL431ACDR2G	]	SOIC-8 (Pb-Free)		1.0%
TL431BCDR2		SOIC-8		
TL431BCDR2G	]	SOIC-8 (Pb-Free)	2500 Units / Tape & Reel	0.4%
TL431CDR2		SOIC-8		
TL431CDR2G	]	SOIC-8 (Pb-Free)		2.2%
TL431ACDMR2		Micro8		1.0%
TL431BCDMR2		IVIICIOO		
TL431BCDMR2G		Micro8 (Pb-Free)	4000 Units / Tape & Reel	0.4%
TL431CDMR2		Micro8		
TL431CDMR2G	]	Micro8 (Pb-Free)		2.2%
TL431ACP				1.0%
TL431BCP	0°C to 70°C	PDIP-8		0.4%
TL431CP	1		50 Units / Rail	2.2%
TL431CPG		PDIP-8 (Pb-Free)		
TL431ACLP		TO-92 (TO-226)		
TL431ACLPG		TO-92 (TO-226) (Pb-Free)		1.0%
TL431BCLP		TO-92 (TO-226)	1	
TL431BCLPG		TO-92 (TO-226) (Pb-Free)	2000 Units / Bag	0.4%
TL431CLP		TO-92 (TO-226)	1	
TL431CLPG		TO-92 (TO-226) (Pb-Free)		2.2%
TL431ACLPRA		TO-92 (TO-226)		
TL431ACLPRAG		TO-92 (TO-226) (Pb-Free)		1.0%
TL431BCLPRA		TO-92 (TO-226)		
TL431BCLPRAG		TO-92 (TO-226) (Pb-Free)		0.4%
TL431CLPRA	]	TO-92 (TO-226)	2000 Units / Tape & Reel	
TL431CLPRAG	]	TO-92 (TO-226) (Pb-Free)		2.2%
TL431ACLPRE		TO-92 (TO-226)	1	
TL431ACLPREG		TO-92 (TO-226) (Pb-Free)		1.0%
TL431BCLPRE	1	TO-92 (TO-226)	1	0.4%

<sup>†</sup>For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

#### **ORDERING INFORMATION**

Device	Operating Temperature Range	Package Code	Shipping Information <sup>†</sup>	Tolerance
TL431BCLPREG		TO-92 (TO-226) (Pb-Free)	2000 Units / Tape & Reel	0.4%
TL431ACLPRP		TO-92 (TO-226)		
TL431ACLPRPG		TO-92 (TO-226) (Pb-Free)		1.0%
TL431BCLPRM	0°C to 70°C	TO-92 (TO-226)		
TL431BCLPRMG		TO-92 (TO-226) (Pb-Free)	2000 Units / Fan–Fold	0.4%
TL431CLPRP		TO-92 (TO-226)	1	
TL431CLPRPG		TO-92 (TO-226) (Pb-Free)		2.2%
TL431AID		SOIC-8		1.0%
TL431AIDG		SOIC-8 (Pb-Free)	]	1.0%
TL431BID		SOIC-8		
TL431BIDG		SOIC-8 (Pb-Free)	98 Units / Rail	0.4%
TL431ID		SOIC-8		
TL431IDG		SOIC-8 (Pb-Free)		2.2%
TL431AIDR2		SOIC-8		
TL431AIDR2G		SOIC-8 (Pb-Free)	1	1.0%
TL431BIDR2		SOIC-8		
TL431BIDR2G		SOIC-8 (Pb-Free)	2500 Units / Tape & Reel	0.4%
TL431IDR2		SOIC-8		
TL431IDR2G		SOIC-8 (Pb-Free)		2.2%
TL431AIDMR2		Missis		1.0%
TL431BIDMR2	4000 12 0500	Micro8		
TL431BIDMR2G	—— −40°C to 85°C	Micro8 (Pb-Free)	4000 Units / Tape & Reel	0.4%
TL431IDMR2		Micro8	1	
TL431IDMR2G		Micro8 (Pb-Free)		2.2%
TL431AIP		PDIP-8		
TL431AIPG		PDIP-8 (Pb-Free)	50 Units / Rail	1.0%
TL431BIP		DDID 0	1	0.4%
TL431IP		PDIP-8		2.2%
TL431AILP		TO 02 (TO 226)		1.0%
TL431BILP		TO-92 (TO-226)		
TL431BILPG		TO-92 (TO-226) (Pb-Free)	2000 Units / Box	0.4%
TL431ILP		TO-92 (TO-226)	1	
TL431ILPG		TO-92 (TO-226) (Pb-Free)	]	2.2%
TL431AILPRA TL431AILPRAG		TO-92 (TO-226) TO-92 (TO-226)	2000 Units / Tape & Reel	1.0%
1 L43 IAILFRAG		(Pb–Free)	2000 Office / Tapo & Nooi	1.070

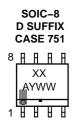
<sup>†</sup>For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

#### **ORDERING INFORMATION**

Device	Operating Temperature Range	Package Code	Shipping Information <sup>†</sup>	Tolerance
TL431BILPRA		TO-92 (TO-226)		
TL431BILPRAG		TO-92 (TO-226) (Pb-Free)	2000 Unito / Tono & Dool	0.4%
TL431ILPRA		TO-92 (TO-226)	2000 Units / Tape & Reel	
TL431ILPRAG	-40°C to 85°C	TO-92 (TO-226) (Pb-Free)	]	2.2%
TL431AILPRM				1.00/
TL431AILPRP		TO-92 (TO-226)	2000 Units / Ammo Pack	1.0%
TL431ILPRP				2.2%
TL431BVD		SOIC-8	98 Units / Rail	
TL431BVDR2		SOIC-8	96 Units / Raii	
TL431BVDMR2	4000 / 40500	Micro8	4000 Units / Tape & Reel	0.4%
TL431BVLP	-40°C to 125°C	TO-92 (TO-226)	2000 Units / Box	
TL431BVP		PDIP-8	50 Units / Rail	
NCV431AIDMR2		Micro8	4000 Units / Tape & Reel	10/
NCV431AIDR2		SOIC-8	2500 Units / Tape & Reel	1%

<sup>†</sup>For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

#### **MARKING DIAGRAMS**







PDIP-8 CASE 626



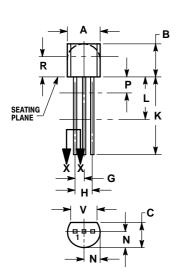
xx = Specific Device Code A = Assembly Location

WL, L = Wafer Lot
YY, Y = Year
WW, W = Work Week
Pb-Free Package

TO-92 (TO-226) CASE 29



#### **PACKAGE DIMENSIONS**



TO-92 (TO-226) LP SUFFIX PLASTIC PACKAGE CASE 29-11 **ISSUE AL** 



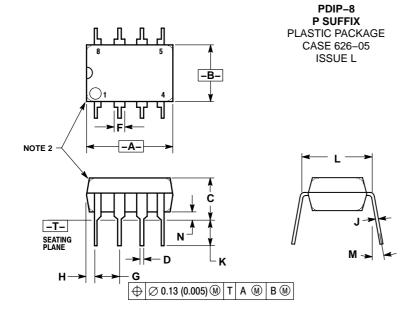
- NOTES:

  1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.

  2. CONTROLLING DIMENSION: INCH.

  3. CONTOUR OF PACKAGE BEYOND DIMENSION R IS UNCONTROLLED.
- LEAD DIMENSION IS UNCONTROLLED IN P AND BEYOND DIMENSION K MINIMUM.

	INC	HES	MILLIN	IETERS
DIM	MIN	MAX	MIN	MAX
Α	0.175	0.205	4.45	5.20
В	0.170	0.210	4.32	5.33
С	0.125	0.165	3.18	4.19
D	0.016	0.021	0.407	0.533
G	0.045	0.055	1.15	1.39
Н	0.095	0.105	2.42	2.66
Ĺ	0.015	0.020	0.39	0.50
K	0.500		12.70	
L	0.250		6.35	
N	0.080	0.105	2.04	2.66
Р		0.100		2.54
R	0.115		2.93	
٧	0.135		3.43	

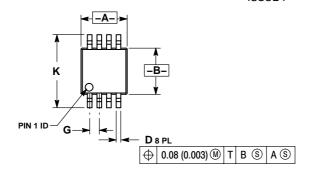


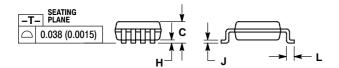
- NOTES:
  1. DIMENSION L TO CENTER OF LEAD WHEN FORMED PARALLEL.
  2. PACKAGE CONTOUR OPTIONAL (ROUND OR
- SQUARE CORNERS).
  3. DIMENSIONING AND TOLERANCING PER ANSI

	MILLIN	IETERS	INC	HES
DIM	MIN	MAX	MIN	MAX
Α	9.40	10.16	0.370	0.400
В	6.10	6.60	0.240	0.260
С	3.94	4.45	0.155	0.175
D	0.38	0.51	0.015	0.020
F	1.02	1.78	0.040	0.070
G	2.54	BSC	0.100	BSC
Н	0.76	1.27	0.030	0.050
7	0.20	0.30	0.008	0.012
K	2.92	3.43	0.115	0.135
L	7.62		0.300	
M		10°		10°
N	0.76	1.01	0.030	0.040

#### PACKAGE DIMENSIONS

Micro8 **DM SUFFIX** PLASTIC PACKAGE CASE 846A-02 ISSUE F





#### NOTES:

- IOTES:

  1. DIMENSIONING AND TOLERANCING PER ANSI
  Y14.5M, 1982.

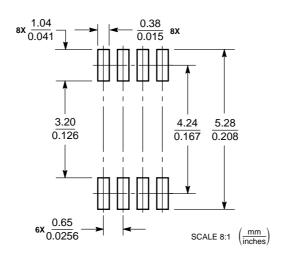
  2. CONTROLLING DIMENSION: MILLIMETER.

  3. DIMENSION A DOES NOT INCLUDE MOLD FLASH,
- 3. DIMENSION A DUES NOT INCLUDE MULD FLASH, PROTRUSIONS OR GATE BURRS. MOLD FLASH, PROTRUSIONS OR GATE BURRS SHALL NOT EXCEED 0.15 (0.006) PER SIDE.

  4. DIMENSION B DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSION. INTERLEAD FLASH OR PROTRUSION. INTERLEAD FLASH OR
- PROTRUSION SHALL NOT EXCEED 0.25 (0.010) PER SIDE.
- 5. 846A-01 OBSOLETE, NEW STANDARD 846A-02.

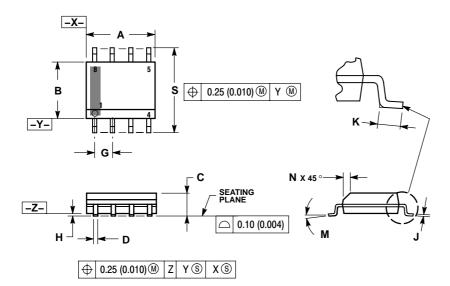
	MILLIMETERS		INC	HES	
DIM	MIN	MAX	MIN	MAX	
Α	2.90	3.10	0.114	0.122	
В	2.90	3.10	0.114	0.122	
C		1.10		0.043	
D	0.25	0.40	0.010	0.016	
G	0.65	BSC	0.026 BSC		
H	0.05	0.15	0.002	0.006	
7	0.13	0.23	0.005	0.009	
K	4.75	5.05	0.187	0.199	
L	0.40	0.70	0.016	0.028	

#### **SOLDERING FOOTPRINT\***



\*For additional information on our Pb-Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

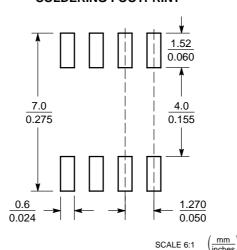
#### SOIC-8 **D SUFFIX** PLASTIC PACKAGE CASE 751-07 **ISSUE AG**



- NOTES:
  1. DIMENSIONING AND TOLERANCING PER
- ANSI Y14.5M, 1982.
  CONTROLLING DIMENSION: MILLIMETER.
  DIMENSION A AND B DO NOT INCLUDE
  MOLD PROTRUSION.
- MAXIMUM MOLD PROTRUSION 0.15 (0.006) PER SIDE
- DIMENSION D DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.127 (0.005) TOTAL IN EXCESS OF THE D DIMENSION AT MAXIMUM MATERIAL CONDITION.
- 751-01 THRU 751-06 ARE OBSOLETE. NEW STANDARD IS 751-07.

	MILLIMETERS		INCHES	
DIM	MIN	MAX	MIN	MAX
Α	4.80	5.00	0.189	0.197
В	3.80	4.00	0.150	0.157
С	1.35	1.75	0.053	0.069
D	0.33	0.51	0.013	0.020
G	1.27 BSC		0.050 BSC	
Н	0.10	0.25	0.004	0.010
J	0.19	0.25	0.007	0.010
K	0.40	1.27	0.016	0.050
M	0 °	8 °	0 °	8 °
N	0.25	0.50	0.010	0.020
S	5.80	6.20	0.228	0.244

#### **SOLDERING FOOTPRINT\***



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