

#### ISL29035

Integrated Digital Light Sensor with Interrupt

FN8371 Rev 3.00 December 12, 2016

The <u>ISL29035</u> is an integrated ambient and infrared light-to-digital converter with I<sup>2</sup>C (SMBus compatible) Interface. Its advanced self-calibrated photodiode array emulates human eye response with excellent IR rejection. The on-chip ADC is capable of rejecting 50Hz and 60Hz flicker caused by artificial light sources. The Lux range select feature allows users to program the Lux range for optimized counts/Lux.

For ambient light sensing, an internal 16-bit ADC has been designed based upon the charge-balancing technique. The ADC conversion time is nominally 105ms and is user adjustable from 11 $\mu$ s to 105ms, depending on oscillator frequency and ADC resolution. In normal operation, typical current consumption is 57 $\mu$ A. In order to further minimize power consumption, two power-down modes have been provided. If polling is chosen over continuous measurement of light, the auto power-down function shuts down the whole chip after each ADC conversion for the measurement. The other power-down mode is controlled by software via the I<sup>2</sup>C interface. The power consumption can be reduced to less than 0.3 $\mu$ A when powered down.

The ISL29035 supports a software brownout condition detection. The device powers up with the brownout bit asserted until the host clears it through the  $\rm I^2C$  interface.

The ISL29035 supports a software and hardware interrupt that remains asserted until the host clears it through the  $I^2C$  interface. Function of ADC conversion continues without stopping after interrupt is asserted.

Designed to operate on supplies from 2.25V to 3.63V with an  $I^2C$  supply from 1.7V to 3.63V, the ISL29035 is specified for operation across the -40°C to +85°C ambient temperature range.

#### **Features**

• Resolution	16-bit ADC
Wide dynamic range	1:4,200,000
$\bullet \ \ \text{Integrated noise reduction} \ \ldots$	50/60Hz
Close to human eye response w	ith excellent IR/UV rejection
• Shutdown modes	software and automatic
Programmable interrupt thresh	nold with persistence filter
• Supply current (typical)	57μΑ
• Shutdown current (maximum)	0.51μA
• I <sup>2</sup> C (SMB compatible) power su	upply 1.7V to 3.63V
• Sensor power supply	2.25V to 3.63V
Operating temperature range.	40°C to +85°C
• Small form factor package	6 Ld 1.5x1.6x0.75 ODFN

## **Applications**

- · Mobile devices: smart phone, PDA, GPS
- · Computing devices: notebook PC, MacBook, tablets
- · Consumer devices: LCD-TV, digital camera
- · Industrial and medical light sensing

#### **Related Literature**

 AN1591, "Evaluation Hardware/Software Manual for ALS and Proximity Sensor"

TABLE 1. KEY DIFFERENCES BETWEEN FAMILY OF PARTS

PART NUMBER	ALS SENSING	INTERRUPT PIN	NUMBER OF PINS
ISL29034	Yes	No	4 Ld
ISL29035	Yes	Yes	6 Ld

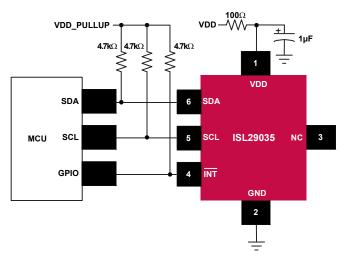


FIGURE 1. ISL29035 TYPICAL APPLICATION DIAGRAM

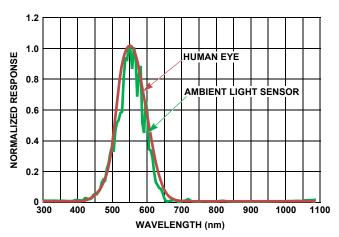


FIGURE 2. NORMALIZED SPECTRAL RESPONSE FOR AMBIENT LIGHT SENSING

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## **Block Diagram**

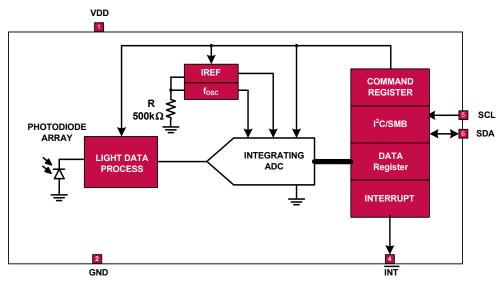
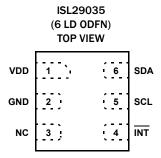


FIGURE 3. BLOCK DIAGRAM

## **Pin Configuration**



## **Pin Descriptions**

PIN NUMBER	PIN NAME	DESCRIPTION
1	VDD	Positive supply
2	GND	Ground pin
3	NC	No connect
4	ĪNT	Interrupt pin; LOW for interrupt alarming. INT pin is an open-drain. INT remains asserted until the interrupt status bit is reset.
5	SCL	I <sup>2</sup> C serial clock
6	SDA	I <sup>2</sup> C serial data

## **Ordering Information**

PART NUMBER (Notes 1, 2, 3)	TEMP RANGE (°C)	TAPE AND REEL (UNITS)	PACKAGE (RoHS COMPLIANT)	PKG. DWG. #
ISL29035IROZ-T7	-40 to +85	3k units	6 Ld ODFN	L6.1.5x1.6
ISL29035IROZ-T7A	-40 to +85	250 units	6 Ld ODFN	L6.1.5x1.6
ISL29035EVAL1Z	Evaluation Board			

- 1. Please refer to  $\underline{\mathsf{TB347}}$  for details on reel specifications.
- 2. These Intersil Pb-free plastic packaged products employ special Pb-free material sets; molding compounds/die attach materials and NiPdAu-Ag plate e4 termination finish, which is RoHS compliant and compatible with both SnPb and Pb-free soldering operations. Intersil Pb-free products are MSL classified at Pb-free peak reflow temperatures that meet or exceed the Pb-free requirements of IPC/JEDEC J STD-020.
- 3. For Moisture Sensitivity Level (MSL), please see product information page for <u>ISL29035</u>. For more information on MSL, please see tech brief <u>TB477</u>.

#### **Absolute Maximum Ratings**

VDD to GND	+4.0V
I <sup>2</sup> C Bus (SCL, SDA) and INT Pin Voltage	0.2V to 4.0V
I <sup>2</sup> C Bus (SCL, SDA) and INT Pin Current	<10mA
Input Voltage Slew Rate (Maximum)	0.1V/µs
ESD Ratings	
Human Body Model	3kV

#### **Thermal Information**

Thermal Resistance (Typical)	$\theta_{JA}(^{\circ}C/W)$
6 Ld ODFN Package (Note 4)	210
Maximum Junction Temperature (TJ <sub>MAX</sub> )	+90°C
Storage Temperature Range40	0°C to +100°C
Operating Temperature	40°C to +85°C
Pb-Free Reflow Profile	see <u>TB477</u>

CAUTION: Do not operate at or near the maximum ratings listed for extended periods of time. Exposure to such conditions may adversely impact product reliability and result in failures not covered by warranty.

#### NOTE:

 θ<sub>JA</sub> is measured in free air with the component mounted on a high effective thermal conductivity test board with "direct attach" features. See Tech Brief <u>TB379</u>.

## **Electrical Specifications** $V_{DD} = 3.0V$ , $T_A = +25$ °C, 16-bit ADC operation, unless otherwise specified.

DESCRIPTION	PARAMETER	TEST CONDITIONS	MIN (Note 7)	TYP	MAX (Note 7)	UNIT
Power Supply Range	V <sub>DD</sub>		2.25		3.63	V
Supply Current	I <sub>DD</sub>			57	85	μΑ
Supply Current when Powered Down	I <sub>DD1</sub>	Software disabled or auto power-down		0.24	0.51	μΑ
Supply Voltage Range for I <sup>2</sup> C Interface	V <sub>I2C</sub>		1.7		3.63	V
ADC Integration/Conversion Time	t <sub>int</sub>	16-bit ADC data		105		ms
I <sup>2</sup> C Clock Rate Range	F <sub>I</sub> <sup>2</sup> <sub>C</sub>			400		kHz
Count Output when Dark	DATA_0	E = 0 Lux, Range 0 (1k Lux)		1	5	Counts
Full Scale ADC Code	DATA_F				65535	Counts
Part-to-Part Variation (3σ Population)	%/Value	E = 300 Lux, cold white LED Range 0 (1k Lux)		±5		%
Light Count Output with LSB of 0.015 Lux/Count	ADC <sub>R0</sub>	E = 300 Lux, cold white LED (Note 5), ALS Range 0 (1k Lux)		20473		Counts
Light Count Output with LSB of 0.06 Lux/Count	ADC <sub>R1</sub>	E = 300 Lux, cold white LED (Note 5), ALS Range 1 (4k Lux)		5100		Counts
Light Count Output with LSB of 0.24 Lux/Count	ADC <sub>R2</sub>	E = 300 Lux, cold white LED (Note 5), ALS Range 2 (16k Lux)		1400		Counts
Light Count Output with LSB of 0.96 Lux/Count	ADC <sub>R3</sub>	E = 300 Lux, cold white LED (Note 5), ALS Range 3 (64k Lux)		366		Counts
Infrared Count Output (Note 6)	ADC_IR <sub>RO</sub>	Range 0 (1k Lux)	1402	1997	2598	Counts
Infrared Count Output (Note 6)	ADC_IR <sub>R1</sub>	Range 1 (4k Lux)		481		Counts
Infrared Count Output (Note 6)	ADC_IR <sub>R2</sub>	Range 2 (16k Lux)		148		Counts
Infrared Count Output (Note 6)	ADC_IR <sub>R3</sub>	Range 3 (64k Lux)		42		Counts
SDA Current Sinking Capability	I <sub>SDA</sub>		4	5		mA
INT Current Sinking Capability	I <sub>INT</sub>		4	5		mA

- 5. 550nm green LED is used in production test. The 550nm LED irradiance is calibrated to produce the same DATA count against an illuminance level of 300 Lux Cold White LED.
- 6. 850nm IR LED is used in production test. The 850nm LED irradiance is calibrated to produce the same DATA\_IR count against and illuminance level of 210 lux sunlight at sea level.
- 7. Compliance to datasheet limits is assured by one or more methods: production test, characterization and/or design.



# $\textbf{l^2C Interface Specifications} \quad \textit{V}_{DD} = 3.0 \textit{V}, \textit{T}_{A} = +25\,^{\circ}\textit{C}, \ \textbf{16-bit ADC operation, unless otherwise specified}.$

PARAMETER	SYMBOL	TEST CONDITIONS	MIN ( <u>Note 7</u> )	TYP	MAX (Note 7)	UNIT
SDA and SCL Input Buffer LOW Voltage	$V_{IL}$				0.55	V
SDA and SCL Input Buffer HIGH Voltage	V <sub>IH</sub>		1.25			V
SDA and SCL Input Buffer Hysteresis	V <sub>Hys</sub> ( <u>Note 8</u> )			0.05 x V <sub>DD</sub>		٧
SDA Output Buffer LOW Voltage (Open-Drain), Sinking 4mA	V <sub>OL</sub> ( <u>Note 8</u> )		0	0.06	0.4	٧
SDA and SCL Pin Capacitance	C <sub>PIN</sub> ( <u>Note 8</u> )	$T_A = +25 ^{\circ}\text{C}, f = 1\text{MHz}, V_{DD} = 5\text{V}, V_{IN} = 0\text{V}, V_{OUT} = 0\text{V}$			10	pF
SCL Frequency	f <sub>SCL</sub>				400	kHz
Pulse Width Suppression Time at SDA and SCL Inputs	t <sub>IN</sub>	Any pulse narrower than the maximum specification is suppressed			50	ns
SCL Falling Edge to SDA Output Data Valid	t <sub>AA</sub>				900	ns
Time the Bus Must be Free before the Start of a New Transmission	t <sub>BUF</sub>		1300			ns
Clock LOW Time	t <sub>LOW</sub>		1300			ns
Clock HIGH Time	t <sub>HIGH</sub>		600			ns
START Condition Set-Up Time	t <sub>SU:STA</sub>		600			ns
START Condition Hold Time	t <sub>HD:STA</sub>		600			ns
Input Data Set-Up Time	t <sub>SU:DAT</sub>		100			ns
Input Data Hold Time	t <sub>HD:DAT</sub>		30			ns
STOP Condition Set-Up Time	t <sub>SU:STO</sub>		600			ns
STOP Condition Hold Time	t <sub>HD:STO</sub>		600			ns
Output Data Hold Time	t <sub>DH</sub>		0			ns
SDA and SCL Rise Time	t <sub>R</sub> ( <u>Note 8</u> )		20 + 0.1 x Cb			ns
SDA and SCL Fall Time	t <sub>F</sub> ( <u>Note 8</u> )		20 + 0.1 x Cb			ns
Capacitive Loading of SDA or SCL	C <sub>b</sub> ( <u>Note 10</u> )	Total on-chip and off-chip			400	pF
SDA and SCL Bus Pull-Up Resistor Off-Chip	R <sub>PU</sub> ( <u>Note 8</u> )	Maximum is determined by $t_R$ and $t_F$ . For Cb = 400pF, max is about $2k\Omega \sim 2.5k\Omega$ For Cb = 40pF, max is about $15k\Omega \sim 20k\Omega$	1			k?

- 8. Limits should be considered typical and are not production tested.
- 9. These are I<sup>2</sup>C specific parameters and are not tested, however, they are used to set conditions for testing devices to validate specification.
- 10.  $C_b$  is the capacitance of the bus in pF.



# **SDA vs SCL Timing**

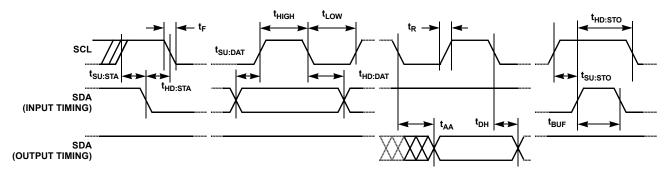


FIGURE 4. I<sup>2</sup>C BUS TIMING

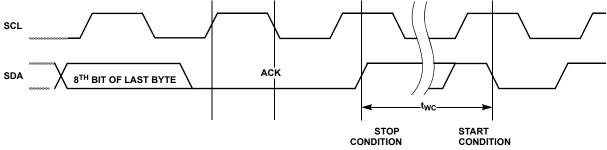


FIGURE 5. I<sup>2</sup>C WRITE CYCLE TIMING

# **Typical Performance Curves**

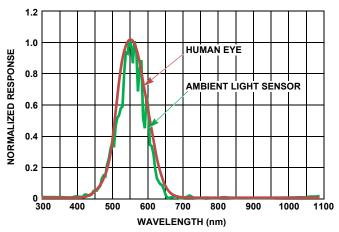


FIGURE 6. NORMALIZED SPECTRAL RESPONSE FOR AMBIENT LIGHT SENSING

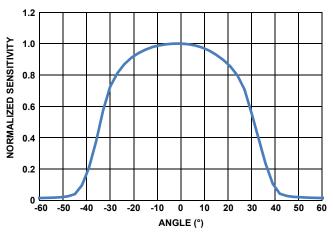


FIGURE 7. NORMALIZED RADIATION PATTERN

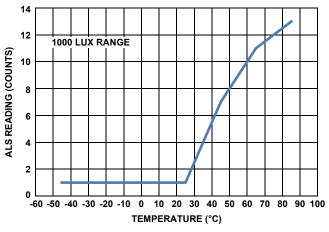


FIGURE 8. TEMPERATURE TEST IN DARK CONDITION

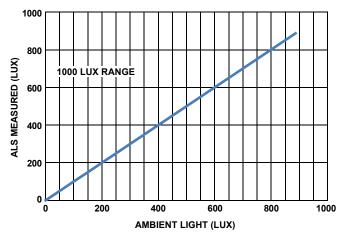


FIGURE 9. ALS TRANSFER FUNCTION

## **Principles of Operation**

#### **Photodiodes and ADC**

The ISL29035 contains two photodiode arrays, which convert light into current. The spectral response for ambient light sensing is shown in <u>Figure 6 on page 7</u>. After light is converted to current during the light signal process, the current output is converted to digital by a built-in 16-bit Analog-to-Digital Converter (ADC). An I<sup>2</sup>C command reads the ambient light or IR intensity in counts.

The converter is a charge-balancing integrating type 16-bit ADC. The chosen method for conversion is best for converting small current signals in the presence of an AC periodic noise. A 105ms integration time, for instance, highly rejects 50Hz and 60Hz power line noise simultaneously.

The integration time of the built-in ADC is determined by the internal oscillator, and the n-bit (n = 4, 8, 12, 16) counter inside the ADC. A good balancing act of integration time and resolution (depending on the application) is required for optimal results.

The ADC has I<sup>2</sup>C programmable range select to dynamically accommodate various lighting conditions. For very dim conditions, the ADC can be configured at its lowest range (Range 0) in the ambient light sensing.

#### **Low-Power Operation**

The ISL29035 initial operation is at the power-down mode after a supply voltage is provided. The data registers contain the default value of 0. When the ISL29035 receives an I<sup>2</sup>C command to do a one-time measurement from an I<sup>2</sup>C master, it will start ADC conversion with light sensing. The ISL29035 will go to the power-down mode automatically after one conversion is finished and keep the conversion data available for the master to fetch anytime afterwards. When receiving an I<sup>2</sup>C command of continuous measurement, the device will continuously do ADC conversions with light sensing and will continuously update the data registers with the latest conversion data. The device will go into power-down mode after receiving the power-down I<sup>2</sup>C command.

#### **Ambient Light and IR Sensing**

There are four operational modes in ISL29035: Programmable ALS once with auto power-down, programmable IR sensing once with auto power-down, programmable continuous ALS sensing and programmable continuous IR sensing. These four modes can be programmed in series to fulfill the application needs. The detailed program configuration is listed in "Command-I Register (Address: 0x00)" on page 11.

When the part is programmed for ambient light sensing, the ambient light wavelength within the "Ambient Light Sensing" spectral response curve in Figure 15 is converted into current. With ADC, the current is converted to an unsigned n-bit (up to 16 bits) digital output.

When the part is programmed for Infrared (IR) sensing, the IR light wavelength within the "IR Sensing" spectral response curve in <u>Figure 15</u> is converted into current. With ADC, the current is converted to an unsigned n-bit (up to 16 bits) digital output.

#### **Interrupt Function**

The active low interrupt pin is an open-drain pull-down configuration. The interrupt pin serves as an alarm or monitoring function to determine whether the ambient light level exceeds the upper threshold or goes below the lower threshold. It should be noted that the function of ADC conversion continues without stopping after interrupt is asserted. If the user needs to read the ADC count that triggers the interrupt, the reading should be done before the data registers are refreshed by the following conversions. The user can also configure the persistence of the interrupt pin. This reduces the possibility of false triggers, such as noise or sudden spikes in ambient light conditions. An unexpected camera flash, for example, can be ignored by setting the persistence to 8 integration cycles.

### **Serial Interface**

The ISL29035 supports the Inter-Integrated Circuit ( $I^2C$ ) bus data transmission protocol. The  $I^2C$  bus is a two-wire serial bidirectional interface consisting of SCL (Clock) and SDA (Data). Both the wires are connected to the device supply via pull-up resistors. The  $I^2C$  protocol defines any device that sends data onto the bus as a transmitter and the receiving device as the receiver. The device controlling the transfer is a master and the device being controlled is the slave. The transmitting device pulls down the SDA line to transmit a "0" and releases it to transmit a "1". The master always initiates the data transfer, only when the bus is not busy, and provides the clock for both transmitting and receiving operations. The ISL29035 operates as a slave device in all applications. The serial communication over the  $I^2C$  interface is conducted by sending the Most Significant Bit (MSB) of each byte of data first.

#### **Start Condition**

During data transfer, the SDA line must remain stable while the SCL line is HIGH. All I<sup>2</sup>C interface operations must begin with a START condition, which is a HIGH to LOW transition of SDA while SCL is HIGH (refer to Figure 12 on page 9). The ISL29035 continuously monitors the SDA and SCL lines for the START condition and does not respond to any command until this condition is met (refer to Figure 12). A START condition is ignored during the power-up sequence.

#### **Stop Condition**

All I<sup>2</sup>C interface operations must be terminated by a STOP condition, which is a LOW to HIGH transition of SDA while SCL is HIGH (refer to Figure 12). A STOP condition at the end of a read/write operation places the device in its standby mode. If a stop is issued in the middle of a Data byte, or before 1 full Data byte + ACK is sent, then the serial communication of the ISL29035 resets itself without performing the read/write. The contents of the array are not affected.

#### **Acknowledge**

An Acknowledge (ACK) is a software convention used to indicate a successful data transfer. The transmitting device releases the SDA bus after transmitting 8 bits. During the ninth clock cycle, the receiver pulls the SDA line LOW to acknowledge the reception of the 8 bits of data (refer to Figure 12). The ISL29035 responds with an ACK after recognition of a START condition followed by a



valid Identification Byte, and once again, after successful receipt of an Address Byte. The ISL29035 also responds with an ACK after receiving a Data byte of a write operation. The master must respond with an ACK after receiving a Data byte of a read operation.

#### **Device Addressing**

Following a START condition, the master must output a Device Address byte. The 7 MSBs of the Device Address byte are known as the device identifier. The device identifier bits of the ISL29035 are internally hard-wired as "1000100". The LSB of the Device Address byte is defined as a Read or Write (R/ $\overline{W}$ ) bit. When this R/ $\overline{W}$  bit is a "1", a read operation is selected and when "0", a write operation is selected (refer to Figure 10). The master generates a START condition followed by Device Address byte 1000100x (x as R/ $\overline{W}$ ) and the ISL29035 compares it with the internal device identifier. Upon a correct comparison, the device outputs an acknowledge (LOW) on the SDA line (refer to Figure 12).

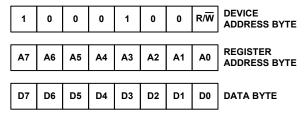


FIGURE 10. DEVICE ADDRESS, REGISTER ADDRESS AND DATA BYTE

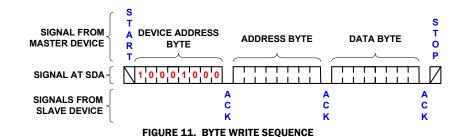
#### **Write Operation**

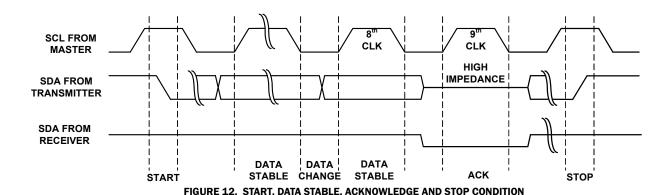
#### **BYTE WRITE**

In a byte write operation, the ISL29035 requires the Device Address byte, Register Address byte and the Data byte. The master starts the communication with a START condition. Upon receipt of the Device Address byte, Register Address byte and the Data byte, the ISL29035 responds with an Acknowledge (ACK). Following the ISL29035 data acknowledge response, the master terminates the transfer by generating a STOP condition. The ISL29035 then begins an internal write cycle of the data to the volatile memory. During the internal write cycle, the device inputs are disabled and the SDA line is in a high impedance state, so the device will not respond to any requests from the master (refer to Figure 11).

#### **BURST WRITE**

The ISL29035 has a burst write operation, which allows the master to write multiple consecutive bytes from a specific address location. It is initiated in the same manner as the byte write operation, but instead of terminating the write cycle after the first Data byte is transferred, the master can write to the whole register array. After the receipt of each byte, the ISL29035 responds with an acknowledge, and the address is internally incremented by one. The address pointer remains at the last address byte written. When the counter reaches the end of the register address list, it "rolls over" and goes back to the first Register Address.





#### **Read Operation**

The ISL29035 has two basic read operations: Byte Read and Burst Read.

#### **BYTE READ**

Byte read operations allow the master to access any register location in the ISL29035. The Byte read operation is a two step process. The master issues the START condition and the Device Address byte with the  $R/\overline{W}$  bit set to "0", receives an acknowledge, then issues the Register Address byte. After acknowledging receipt of the register address byte, the master immediately issues another START condition and the Device Address byte with the  $R/\overline{W}$  bit set to "1". This is followed by an acknowledge from the device and then by the 8-bit data word. The master terminates the read operation by not responding with an acknowledge and then issuing a stop condition (refer to Figure 13).

#### **BURST READ**

Burst read operation is identical to the Byte Read operation. After the first Data byte is transmitted, the master now responds with an acknowledge, indicating it requires additional data. The device continues to output data for each acknowledge received. The master terminates the read operation by not responding with

an acknowledge but issuing a STOP condition (refer to Figure 14).

For more information about the  $I^2C$  standard, please consult the Phillips<sup>TM</sup>  $I^2C$  specification documents.

#### **Power-On Reset**

The Power-On Reset (POR) circuitry protects the internal logic against powering up in the incorrect state. The ISL29035 will power up into Standby mode after  $V_{DD}$  exceeds the POR trigger level and will power down into Reset mode when  $V_{DD}$  drops below the POR trigger level. This bidirectional POR feature protects the device against 'brown-out' failure following a temporary loss of power.

The POR is an important feature because it prevents the ISL29035 from starting to operate with insufficient voltage, prior to stabilization of the internal bandgap. The ISL29035 prevents communication to its registers and greatly reduces the likelihood of data corruption on power-up.

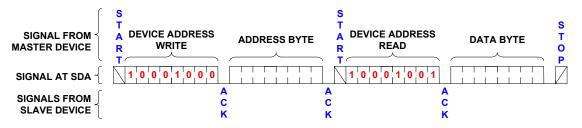


FIGURE 13. BYTE ADDRESS READ SEQUENCE

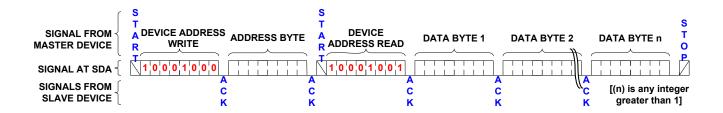


FIGURE 14. BURST READ SEQUENCE

**TABLE 2. REGISTER MAP** 

		STER RESS		REGISTER BITS								
NAME	DEC	HEX	В7	В6	B5	B4	В3	B2	B1	во	DEFAULT	ACCESS
COMMAND-I	0	0x00	0P2	0P1	OP0	RESE	RVED	ĪNT	PRST1	PRST0	0x00	RW
COMMAND-II	1	0x01		RESERVED RES1				RES0	RANGE1	RANGE0	0x00	RW
DATA <sub>LSB</sub>	2	0x02	D7	D6	D5	D4	D3	D2	D1	D0	0x00	RO
DATA <sub>MSB</sub>	3	0x03	D15	D14	D13	D12	D11	D10	D9	D8	0x00	RO
INT_LT_LSB	4	0x04	TL7	TL6	TL5	TL4	TL3	TL2	TL1	TLO	0x00	RW
INT_LT_MSB	5	0x05	TL15	TL14	TL13	TL12	TL11	TL10	TL9	TL8	0x00	RW
INT_HT_LSB	6	0x06	TH7	TH6	TH5	TH4	TH3	TH2	TH1	THO	0xFF	RW
INT_HT_MSB	7	0x07	TH15	TH14	TH13	TH12	TH11	TH10	TH9	TH8	0xFF	RW
ID	15	0x0F	BOUT	RESERVED	1	0	1		RESERVED	•	1x101xxx	RW

## **Register Description**

Following are detailed descriptions of the control registers related to the operation of the ISL29035 ambient light sensor device. These registers are accessed by the  $I^2C$  serial interface. For details on the  $I^2C$  interface, refer to "Serial Interface" on page 8.

All the features of the device are controlled by the registers. The ADC data can also be read. The following sections explain the details of each register bit. All RESERVED bits are Intersil used bits ONLY. The value of the reserved bit can change without notice.

#### **Decimal to Hexadecimal Conversion**

To convert decimal value to hexadecimal value, divide the decimal number by 16 and write the remainder on the side as the least significant digit. This process is continued by dividing the quotient by 16 and writing the remainder until the quotient is 0. When performing the division, the remainders, which will represent the hexadecimal equivalent of the decimal number, are written beginning with the least significant digit (right) and each new digit is written to the next most significant digit (the left) of the previous digit. Consider the number 175 decimal.

**TABLE 3. DECIMAL TO HEXADECIMAL** 

DIVISION	QUOTIENT	QUOTIENT REMAINDER			
175/16	10 = A	15 = F	0xAF		

#### **Command-I Register (Address: 0x00)**

TABLE 4. COMMAND-I REGISTER ADDRESS

	ADDR	REGISTER BITS							DFLT	
NAME	(HEX)	В7	В6	B5	B4	В3	B2	B1	во	(HEX)
COMMAND-I	0x00	0P 2	0P 1	0P 0	RESERVED		ĪNT	PRST 1	PRST 0	0x00

The Command-I register consists of control and status bits. In this register, there are two interrupt persist bits, one interrupt status bit and three operation mode bits. The operation mode bits and the interrupt persist bits are independent of each other. The default register value is 0x00 at power-on.

#### **INTERRUPT PERSIST BITS (B0-B1)**

The interrupt persist bits provide control over when interrupts occur. There are four different selections for this feature. A value of n (where n is 1, 4, 8, and 16) results in an interrupt only if the value remains outside the threshold window for n consecutive integration cycles. For example, if n is equal to 16 and the ADC resolution is set to 16-bits, then the integration time is 105ms. An interrupt is generated whenever the last conversion results in a value outside of the programmed threshold window. The interrupt is active-low and remains asserted until cleared by writing the COMMAND register with the CLEAR bit set. Table 5 lists the possible interrupt persist bits.

**TABLE 5. INTERRUPT PERSIST BITS** 

<b>B1</b>	во	NUMBER OF INTEGRATION CYCLES (n)
0	0	1
0	1	4
1	0	8
1	1	16

#### **INTERRUPT STATUS BIT (B2)**

The interrupt status bit ( $\overline{\text{INT}}$ ) is a status bit for light intensity detection. The bit is set to logic HIGH when the light intensity crosses the interrupt thresholds window (register address 0x04 - 0x07), and set to logic LOW when it is within the interrupt thresholds window. Once the interrupt is triggered, the  $\overline{\text{INT}}$  pin goes low and the interrupt status bit goes HIGH until the status bit is polled through the I<sup>2</sup>C read command. Both the  $\overline{\text{INT}}$  pin and the interrupt status bit are automatically cleared at the end of the 8-bit Device Register byte (0x00) transfer. Table 6 shows the interrupt status states.

TABLE 6. INTERRUPT STATUS BIT (INT)

BIT 2	OPERATION
0	Interrupt is cleared or not triggered yet
1	Interrupt is triggered



#### **OPERATION MODE BITS (B5-B7)**

The ISL29035 has different operating modes. These modes are selected by setting B5 - B7 bits on register address 0x00. The device powers up on a disable mode. Table 7 lists the possible operating modes.

**TABLE 7. OPERATING MODES BITS** 

В7	В6	B5	OPERATION
0	0	0	Power-down the device (default)
0	0	1	The device measures ALS only once every integration cycle. This is the lowest operating mode. (Note 11)
0	1	0	IR once
0	1	1	Reserved (Do Not Use)
1	0	0	Reserved (Do Not Use)
1	0	1	Measures ALS continuously
1	1	0	Measures IR continuous
1	1	1	Reserved (Do Not Use)

NOTE:

11. Intersil does not recommend using this mode.

#### **Command-II Register (Address: 0x01)**

**TABLE 8. COMMAND-II REGISTER BITS** 

			REGISTER BITS									
NAME	ADDR (HEX)	В 7	B 6	B 5	B 4	В3	B2	B1	во	DFLT (HEX)		
COMMAND-II	0x01	RESERVED		RES 1	RES 0	RANGE 1	RANGE 0	0x00				

The Command-II register consists of ADC control bits. In this register, there are two range bits and two ADC resolution bits. The default register value is 0x00 at power-on.

#### **FULL SCALE LUX RANGE (B0-B1)**

The full scale Lux range has four different selectable ranges. The range determines the full scale Lux range (1k, 4k, 16k and 64k). Each range has a maximum allowable Lux value. Lower range values offer better resolution. Table 9 lists the possible values of Lux.

**TABLE 9. RANGE REGISTER BITS** 

RANGE SELECTION	B1	во	FULL SCALE LUX RANGE (Lux)
0	0	0	1,000
1	0	1	4,000
2	1	0	16,000
3	1	1	64,000

#### **ADC RESOLUTION (B3-B2)**

B2 and B3 determine the ADC's resolution and the number of clock cycles per conversion. Changing the number of clock cycles does more than just change the resolution of the device; it also

changes the integration time, which is the period the device's Analog-to-Digital (A/D) converter samples the photodiode current signal for a measurement. Table 10 lists the possible ADC resolution. Only 16-bit ADC resolution can reject better 50/60Hz noise flickering light source.

**TABLE 10. ADC RESOLUTION DATA WIDTH** 

В3	B2	NUMBER OF CLOCK CYCLES	n-BIT ADC
0	0	2 <sup>16</sup> = 65,536	16
0	1	2 <sup>12</sup> = 4,096	12
1	0	2 <sup>8</sup> = 256	8
1	1	2 <sup>4</sup> = 16	4

#### **Integration Time**

**TABLE 11. INTEGRATION TIME OF n-BIT ADC** 

n # ADC BITS	INTEGRATION TIME (ms)
4	0.0256
8	0.41
12	6.5
16	105

#### Data Registers (Addresses: 0x02 and 0x03)

**TABLE 12. ADC REGISTER BITS** 

	ADDR	REGISTER BITS								
NAME	(HEX)	В7	В6	B5	В4	В3	B2	<b>B1</b>	во	DFLT (HEX)
DATA <sub>LSB</sub>	0x02	D7	D6	D5	D4	D3	D2	D1	D0	0x00
DATA <sub>MSB</sub>	0x03	D15	D14	D13	D12	D11	D10	D9	D8	0x00

The ISL29035 has two 8-bit read-only registers to hold the upper and lower byte of the ADC value. The upper byte is accessed at address 0x03 and the lower byte is accessed at address 0x02. For 16-bit resolution, the data is from D0 to D15; for 12-bit resolution, the data is from D0 to D11; for 8-bit resolution, the data is from D0 to D7 and for 4-bit resolution, the data is from D0 to D3. The registers are refreshed after every conversion cycle. The default register value is 0x00 at power-on.

**TABLE 13. ADC DATA REGISTERS** 

ADDRESS (HEX)	CONTENTS
0x02	D0 is LSB for 4-, 8-, 12- or 16-bit resolution; D3 is MSB for 4-bit resolution; D7 is MSB for 8-bit resolution
0x03	D15 is MSB for 16-bit resolution; D11 is MSB for 12-bit resolution



# Lower Interrupt Threshold Registers (Address: 0x04 and 0x05)

**TABLE 14. INTERRUPT REGISTER BITS** 

	ADDR	REGISTER BITS								DFLT
NAME	(HEX)	В7	В6	B5	В4	вз	B2	B1	во	(HEX)
INT_LT_LSB	0x04	TL7	TL6	TL5	TL4	TL3	TL2	TL1	TLO	0x00
INT_LT_MSB	0x05	TL1 5	TL1 4	TL1 3	TL1 2	TL1 1	TL1 0	TL9	TL8	0x00

The lower interrupt threshold registers are used to set the lower trigger point for interrupt generation. If the ALS value crosses below or is equal to the lower threshold, an interrupt is asserted on the interrupt pin and the interrupt status. Registers INT\_LT\_LSB (0x04) and INT\_LT\_MSB (0x05) provide the LOW and HIGH bytes, respectively, of the lower interrupt threshold. The HIGH and LOW bytes from each set of registers are combined to form a 16-bit threshold value. The interrupt threshold registers default to 0x00 upon power-up.

# Upper Interrupt Threshold Registers (Address: 0x06 and 0x07)

**TABLE 15. INTERRUPT REGISTER BITS** 

	ADDR	ADDR REGISTER BITS								DFLT
NAME	(HEX)	В7	В6	В5	В4	вз	B2	B1	во	(HEX)
INT_HT_LSB	0x06	TH7	TH6	TH5	TH4	TH3	TH2	TH1	тно	0xFF
INT_HT_MSB	0x07	TH1 5	TH1 4	TH1 3	TH1 2	TH1 1	TH1 0	тн9	TH8	0xFF

The upper interrupt threshold registers are used to set the upper trigger point for interrupt generation. If the ALS value crosses above or is equal to the upper threshold, an interrupt is asserted on the interrupt pin and the interrupt status. Registers INT\_HT\_LSB (0x06) and INT\_HT\_MSB (0x07) provide the LOW and HIGH bytes, respectively, of the upper interrupt threshold. The HIGH and LOW bytes from each set of registers are combined to form a 16-bit threshold value. The interrupt threshold registers default to 0xFF on power-up.

#### ID Register (Address: 0x0F)

**TABLE 16. ID REGISTER BITS** 

	ADDR		REGISTER BITS							
NAME		В7	7 B6		В4	вз	B2	B1	во	DFLT
ID	0x0F	BOUT	RESERVED	1	0	1	RE	RESERVED		1x101xxx

The ID register has three different types of information, which is discussed in the following.

#### **RESERVED BITS (B0-B2 AND B6)**

All RESERVED bits on the ISL29035 are Intersil used bits only. Bit0-Bit2 and Bit6 are RESERVED bits where their value might change without any notification to the user. It is advised when using the identification bits to identify the device in a system, the

software should mask the Bit0-Bit2 and Bit6-Bit7 to properly identify the device.

#### **DEVICE ID BITS (B3-B5)**

The ISL29035 provides 3 bits to identify the device in a system. These bits are located on register address 0x0F, Bit3-Bit5. The identification bit value for the ISL29035 is xx101xxx. The device identification bits are read only bits. It is important to notice that Bit7 is a status bit for Brownout condition (BOUT).

#### **BROWNOUT STATUS BIT - BOUT (B7)**

Bit7 on register address 0x0F is a status bit for Brownout condition (BOUT). The default value of this bit is "BOUT = 1" during the initial power-up, which indicates the device may possibly have gone through a brownout condition. Therefore, the status bit should be reset to "BOUT = 0" by an  $I^2C$  write command during the initial configuration of the device.

The default register value is 0xA8 at power-on.

## **Applications Information**

<u>Figure 15</u> is a normalized spectral response of various types of light sources for reference.

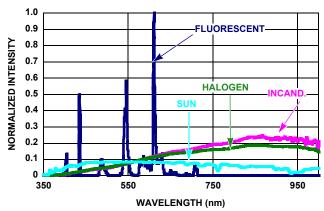


FIGURE 15. NORMALIZED SPECTRAL RESPONSE OF LIGHT SOURCES

#### **Calculating Lux**

The ISL29035's ADC output codes, DATA, are directly proportional to Lux in the ambient light sensing.

$$E_{cal} = \alpha \times DATA$$
 (EQ. 1)

Where  $\mathbf{E}_{\text{cal}}$  is the calculated Lux reading. The constant  $\alpha$  is determined by the full-scale range and the ADC's maximum output counts. The constant is independent of the light sources (fluorescent, incandescent and sunlight) because the light sources' IR component is removed during the light signal process. The constant can also be viewed as the sensitivity (the smallest Lux measurement the device can measure).

$$\alpha = \frac{Range}{Count_{max}}$$
 (EQ. 2)

Where Range is defined in <u>Table 9 on page 12</u>. Count<sub>max</sub> is the maximum output counts from the ADC.



The transfer function used for n-bits ADC becomes Equation 3:

$$E_{cal} = \frac{Range}{2^n} \times DATA$$
 (EQ. 3)

Where n = 4, 8, 12 or 16. This is the number of ADC bits programmed in the command register. 2<sup>n</sup> represents the maximum number of counts possible from the ADC output. Data is the ADC output stored in the data registers (02 hex and 03 hex).

#### **Enhancing EV Accuracy**

The device has an on-chip passive optical filter designed to block (reject) most of the incident Infrared. However, EV measurement may vary under differing IR-content light sources. In order to optimize the measurement variation between differing IR-content light sources, ISL29035 provides IR channel, which is programmed at COMMAND-I (Reg0x0) to measure the IR level of differing IR-content light sources.

The ISL29035's ADC output codes, DATA, are directly proportional to the IR intensity received in the IR sensing.

$$\mathsf{DATA}_{\mathsf{IR}} = \beta \times \mathsf{E}_{\mathsf{IR}} \tag{EQ. 4}$$

Then EV<sub>Accuracy</sub> can be found in Equation 5:

$$EV_{Accuracy} = KxDATA_{EV} + \beta \times DATA_{IR}$$
 (EQ. 5)

Where DATA<sub>EV</sub> is the received ambient light intensity ADC output codes. K is a resolution of visible portion. Its unit is Lux/count. The typical value of K is 0.82. DATA<sub>IR</sub> is the received IR intensity. The constant  $\beta$  changes with the spectrum of background IR, such as A, F2 and D65. The  $\beta$  also changes with the ADC's range and resolution selections. A typical  $\beta$  for Range1 and Range2 is -11292.86 and for Range3 and Range4 it is 2137.14 without IR tinted glass.

#### **Noise Rejection**

Electrical AC power worldwide is distributed at either 50Hz or 60Hz. Artificial light sources vary in intensity at the AC power frequencies. The undesired interference frequencies are infused on the electrical signals. This variation is one of the main sources of noise for the light sensors. Integrating type ADC's have excellent noise-rejection characteristics for periodic noise sources whose frequency is an integer multiple of the conversion rate. By setting the sensor's integration time to an integer multiple of periodic noise signal, the performance of an ambient light sensor can be improved greatly in the presence of noise. In order to reject the AC noise, the integration time of the sensor must be adjusted to match the AC noise cycle. For instance, a 60Hz AC unwanted signal's sum from 0ms to k\*16.66ms  $(k = 1,2...k_i)$  is zero. Similarly, setting the device's integration time to be an integer multiple of the periodic noise signal, greatly improves the light sensor output signal in the presence of noise.

#### **Suggested PCB Footprint**

It is important that users check the <u>TB477</u>, "Surface Mount Assembly Guidelines for Optical Dual Flat Pack No Lead (ODFN) Package" before starting ODFN product board mounting.

#### **Board Mounting Considerations**

For applications requiring the light measurement, the board mounting location should be reviewed. The device uses an Optical Dual Flat Pack No Lead (ODFN) package, which subjects the die to mild stresses when the Printed Circuit (PC) board is heated and cooled, which slightly changes the shape. Because of these die stresses, placing the device in areas subject to slight twisting can cause degradation of reference voltage accuracy. It is normally best to place the device near the edge of a board, or on the shortest side, because the axis of bending is most limited in that location.

#### **Layout Considerations**

The ISL29035 is relatively insensitive to layout. Like other I<sup>2</sup>C devices, it is intended to provide excellent performance even in significantly noisy environments. There are only a few considerations that will ensure best performance.

Route the supply and I<sup>2</sup>C traces as far as possible from all sources of noise. Use two power-supply decoupling capacitors, 1µF and 0.1µF, placed close to the device.

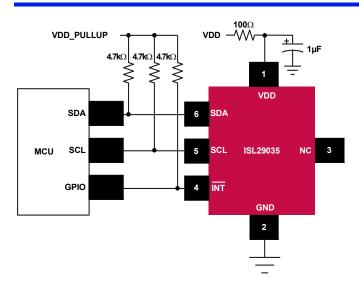
#### **Soldering Considerations**

Convection heating is recommended for reflow soldering; direct infrared heating is not recommended. The plastic ODFN package does not require a custom reflow soldering profile, and is qualified to +260°C. A standard reflow soldering profile with a +260°C maximum is recommended.

#### **Typical Circuit**

A typical application for the ISL29035 is shown in Figure 16 on page 15. The ISL29035's I<sup>2</sup>C address is internally hard-wired as 1000100. The device can be tied onto a system's I<sup>2</sup>C bus, together with other I<sup>2</sup>C compliant devices.





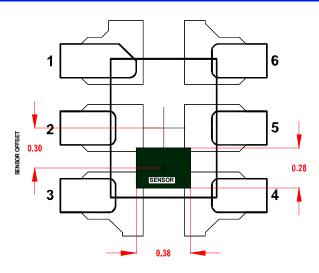


FIGURE 16. ISL29035 TYPICAL CIRCUIT

FIGURE 17. 6 LD ODFN SENSOR LOCATION OUTLINE

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**Revision History** The revision history provided is for informational purposes only and is believed to be accurate, however, not warranted. Please go to web to make sure you have the latest revision.

DATE	REVISION	CHANGE
December 12, 2016	FN8371.3	Updated related literature on page 1.  Added Table 1 on page 1.  Added Note 6 on page 4 and referenced it in EC table.  Removed Note 7 on page 3.  Updated y-axis titles and added labels of Figures 7, 8 and 9. Updated title of Figure 9 (removed under F2 light source) on page 7.
December 24, 2015	FN8371.2	Removed sections "Digital Inputs and Termination" and "Temperature Coefficient" from "Applications Information" on page 13.  Added Related Literature on page 1  Updated Ordering Information table on page 3 by adding T&R quantity column and adding T7A part  Removed IR sensing from Figure 6 on page 7 and from "Photodiodes and ADC" on page 8  Removed 4 notes that were unrelated from Electrical Specifications table on page 4 and removed note referencing removed notes from paragraph after Equation 5 on page 14.  Updated POD L6.1.5x1.6 to most current revision with rev change listed as follows:  Tiebar Note updated  From: Tiebar shown (if present) is a non-functional feature.  To: Tiebar shown (if present) is a non-functional feature and may be located on any of the 4 sides (or ends).
November 12, 2014	FN8371.1	On page 1 updated: 90ms to 105ms and changed Shutdown current from 0.3 to 0.51 and corrected the Labels on Figure 2.  Updated the Ordering Information on page 3.  On page 3 updated pin configuration to show that Pin 1 is the longest pin.  In "Electrical Specifications" on page 4 updated:  - "Light Source Variation" to "Part-to-part Variation".  - test conditions for %/Value, ADC <sub>R0</sub> , ADC <sub>R1</sub> ,ADC <sub>R2</sub> , ADC <sub>R3</sub> , ADC_IR <sub>R0</sub> , ADC_IR <sub>R1</sub> ,ADC_IR <sub>R2</sub> , and ADC_IR <sub>R3</sub> On page 7 corrected the labels on Figure 6.  On page 8 updated:  - under Photodiodes and ADC section Figure reference, changed from 100ms to 105ms, changed from Range 1 to Range 0.  On page 11 under Interrupt Persist Bits (B0-B1) section, updated from 100ms to 105ms  On page 12 added note for Table 7 and verbiage under ADC Resolution (B3-B2) section  On page 14 updated the Temperature Coefficient section.
September 18, 2013	FN8371.0	Initial Release.

### **About Intersil**

Intersil Corporation is a leading provider of innovative power management and precision analog solutions. The company's products address some of the largest markets within the industrial and infrastructure, mobile computing and high-end consumer markets.

For the most updated datasheet, application notes, related documentation and related parts, please see the respective product information page found at <a href="https://www.intersil.com">www.intersil.com</a>.

You may report errors or suggestions for improving this datasheet by visiting www.intersil.com/ask.

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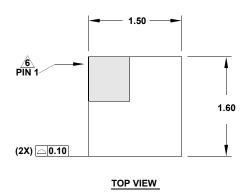


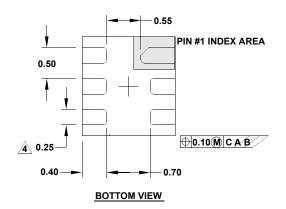
# **Package Outline Drawing**

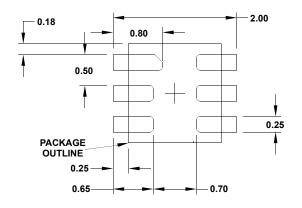
L6.1.5x1.6

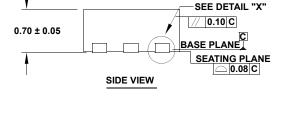
6 LEAD OPTICAL DUAL FLAT NO-LEAD PLASTIC PACKAGE (ODFN) Rev 1, 4/15

For the most recent package outline drawing, see <u>L6.1.5x1.6</u>.

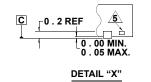












- 1. Dimensions are in millimeters. Dimensions in ( ) for Reference Only.
- 2. Dimensioning and tolerancing conform to ASME Y14.5m-1994.
- 3. Unless otherwise specified, tolerance : Decimal  $\pm 0.05$
- Dimension applies to the metallized terminal and is measured between 0.15mm and 0.30mm from the terminal tip.
- 5 Tiebar shown (if present) is a non-functional feature and may be located on any of the 4 sides (or ends)..
- 6. The configuration of the pin #1 identifier is optional, but must be located within the zone indicated. The pin #1 identifier may be either a mold or mark feature.

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