

#### **Brief Description**

The ZSC31150 is a CMOS integrated circuit for highly accurate amplification and sensor-specific correction of bridge sensor signals. Digital compensation of sensor offset, sensitivity, temperature drift, and non-linearity is accomplished via an internal 16-bit RISC microcontroller running a correction algorithm, with calibration coefficients stored in an EEPROM.

The ZSC31150 is adjustable to nearly all bridge sensor types. Measured values are provided at the analog voltage output or at the digital ZACwire<sup>™</sup> and I<sup>2</sup>C<sup>™\*</sup> interface. The digital interface can be used for a simple PC-controlled calibration procedure in order to program a set of calibration coefficients into an on-chip EEPROM. A specific sensor and a ZSC31150 can be mated digitally: fast, precise, and without the cost overhead associated with trimming by external devices or a laser.

# Features

- Digital compensation of sensor offset, sensitivity, temperature drift, and non-linearity
- Adjustable to nearly all bridge sensor types
- Analog gain of up to 420
- Output options: ratiometric analog voltage output (5% to 95% maximum, 12.4-bit resolution) or ZACwire™ (digital one-wire-interface)
- Temperature compensation: internal or external diode, bridge resistance, thermistor
- · Sensor biasing by voltage or constant current
- Sample rate: up to 7.8kHz
- High voltage protection up to 33V
- Supply current: max. 5.5mA
- Reverse polarity and short-circuit protection
- Wide operation temperature depending on part number: up to -40 to +150°C
- Traceability by user-defined EEPROM entries
- Safety and diagnostic functions

#### **Benefits**

- No external trimming components required
- Only a few external protection devices needed
- PC-controlled configuration and single pass calibration via l<sup>2</sup>C<sup>TM</sup> or ZACwire<sup>TM</sup> interface: simple, cost efficient, quick, and precise
- End-of-line calibration via l<sup>2</sup>C<sup>™</sup> or ZACwire<sup>™</sup> interface
- High accuracy (0.25% FSO @ -25 to 85°C; 0.5% FSO @ -40 to 125°C)
- Excellent EMC/ESD robustness and AEC-Q100 qualification

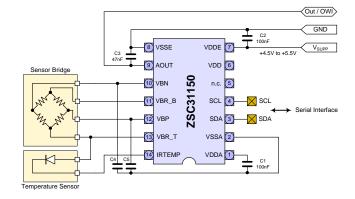
### **Available Support**

- Evaluation Kits
- Application Notes
- Mass Calibration System

### **Physical Characteristics**

- Supply voltage: 4.5 to 5.5 V
- Operation temperature: -40°C to 125°C (-40°C to +150°C extended temperature range)
- Available as DFN14 (5mm x 4mm; wettable flanks), SSOP14, and die

#### ZSC31150 Application Circuit

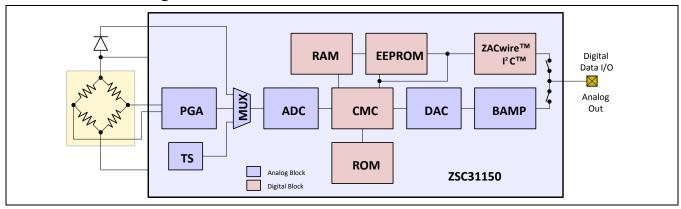


<sup>\*</sup> I<sup>2</sup>C<sup>™</sup> is a trademark of NXP.

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# ZSC31150 Block Diagram



### **Ordering Information**

Sales Code	Description	Package					
ZSC31150GE	ZSC31150 Die — Temperature range: -40°C to +150°C	Unsawn on Wafer: add "B" to sales code Sawn on Wafer Frame: add "C" Waffle Pack: add "D"					
ZSC31150GEG2-R	ZSC31150 DFN14 (5mmx4mm; wettable flanks) —Temperature range: -40°C to 150°C	Tape & Reel					
ZSC31150GAG2-R	ZSC31150 DFN14 (5mmx4mm; wettable flanks) —Temperature range: -40°C to 125°C	Tape & Reel					
ZSC31150GEG1	ZSC31150 SSOP14—Temperature range: -40°C to +150°C	Tube: add "-T" to sales code					
ZSC31150GLG1	ZSC31150 SSOP14—Temperature range: -40°C to +150°C (Long life: 5000h @150°C)	Tape & Reel: add "-R"					
ZSC31150GAG1	ZSC31150 SSOP14—Temperature range: -40°C to +125°C						
ZSC31150KIT Evaluation Kit V1.2	ZSC31150 SSC Evaluation Kit: three interconnecting boards, five ZSUSB cable (software can be downloaded from product page at www						
ZSC31150 Mass       Modular Mass Calibration System (MSC) for ZSC31150: MCS boards, cable, connectors         Calibration System V1.1       (software can be downloaded from product page)							



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# **1** Electrical Characteristics

### 1.1. Absolute Maximum Ratings

The absolute maximum ratings are stress ratings only. The ZSC31150 might not function or be operable above the recommended operating conditions. Stresses exceeding the absolute maximum ratings might also damage the device. In addition, extended exposure to stresses above the recommended operating conditions might affect device reliability. IDT does not recommend designing to the "Absolute Maximum Ratings."

Parameters apply in operation temperature range and without time limitations.

No.	Parameter	Symbol	Conditions	Min	Мах	Unit
1.1.1.	Supply voltage 1)	VDDE	To VSSE.	-33	33	VDC
1.1.2.	Potential at the AOUT pin <sup>1)</sup>	Vout	Relative to VSSE.	-33	33	VDC
1.1.3.	Analog supply voltage <sup>1)</sup>	VDDA	Relative to VSSA. VDDE - VDDA < 0.35 V	-0.3	6.5	VDC
1.1.4.	Voltage at all analog and digital IO pins	V <sub>A_IO</sub> V <sub>D_IO</sub>	Relative to VSSA.	-0.3	VDDA + 0.3	VDC
1.1.5.	Storage temperature	T <sub>STG</sub>		-55	150	°C
1) Refe	er to the ZSC31150 Technical Note	– High Voltage	Protection for specification and d	etailed conditions	for high voltage pro	otection.

Table 1.1 Absolute Maximum Ratings

# 1.2. Operating Conditions

All voltages are related to VSSA. See important table notes at the end of the table.

Table 1.2Operating Conditions

No.	Parameter	Symbol	Conditions	Min	Тур	Max	Unit
1.2.1.	1.2.1.1 TQE ambient temperature range for part numbers ZSC31150x <b>E</b> xx <sup>1)</sup>	T <sub>AMB_TQE</sub>	TQE	-40		150	°C
	1.2.1.2 TQA ambient temperature range for part numbers ZSC31150x <b>A</b> xx <sup>2)</sup>	T <sub>AMB_TQA</sub>	TQA	-40		125	°C
	1.2.1.3 TQI ambient temperature range for advanced performance <sup>2)</sup>	T <sub>AMB_TQI</sub>	ΤQΙ	-25		85	°C
1.2.2.	Supply voltage	VDDE		4.5	5.0	5.5	VDC
1.2.3.	Bridge resistance—Bridge Voltage Mode <sup>2), 3)</sup>	R <sub>BR_V</sub>		2		25	kΩ
1.2.4.	Bridge resistance—Bridge Current Excitation Mode <sup>2), 3)</sup>	R <sub>BR_C</sub>	See specification 1.2.6 for $I_{BR\_MAX}$			10	kΩ
1.2.5.	Current reference resistor <sup>2),4)</sup>	R <sub>IBR</sub>	$I_{BR} = VDDA / (16 * R_{IBR})$	0.07 * R <sub>BR</sub>			kΩ

Ν	lo.	Parameter	Symbol	Conditions	Min	Тур	Max	Unit
1.2.	6.	Maximum bridge current	I <sub>BR_MAX</sub>				2	mA
1.2.	7.	Maximum bridge top voltage	V <sub>BR_TOP</sub>				( <sup>15</sup> / <sub>16</sub> * VDDA) - 0.3	3 V
1.2.	8.	TC current reference resistor <sup>2)</sup>	TC R <sub>IBR</sub>	Behavior influences current generated		50		ppm/K
1)		to the temperature profile descriptio erature range > 125°C.	n in the ZSC	31150 Technical Note – Die a	nd Package S	Specificat	<i>ions</i> for operation in	
2)		easurement in mass production; para	motor is qua	aranteed by design and/or gus	ality observatio	'n		
3)	Symn the Z	netric behavior and identical electrica SC31150 are required. Unsymmetric C31150 can generate a failure in sig	l properties ( al conditions	especially with regard to the l of the sensor and/or external	ow pass chara	acteristic		
		application circuit components in Tab						

# **1.3. Electrical Parameters**

All parameter values are valid for operating conditions specified in section 1.2 except as noted. All voltages related to VSSA. See important table notes at the end of the table.

No.	Parameter	Symbol	Conditions	Min	Тур	Max	Unit						
1.3.1.	3.1. Supply Current and System Operation Conditions												
1.3.1.1.	Supply current	I <sub>S</sub>	Without bridge and load current; $T_{AMB_TQA}$ ; $f_{CLK} \le 3 MHz$			5.5	mA						
1.3.1.2.	Clock frequency <sup>1)</sup>	f <sub>osc</sub>	Guaranteed adjustment range (see the ZSC31150 Functional Description for details); T <sub>AMB_TQA</sub>	2	3	4	MHz						
1.3.2.	Analog Front-End (AFE) Cha	aracteristi	cs										
1.3.2.1.	Input span	$V_{\text{IN}\_\text{SP}}$	Analog gain: 420 to 2.8	1		275	mV/V						
1.3.2.2.	Analog offset compensation range		Depends on gain adjust; refer to section 2.3.1	-300		300	$\% \ V_{\text{IN}\_\text{SP}}$						
1.3.2.3.	Parasitic differential input	I <sub>IN_OFF</sub>	Within T <sub>AMB_TQE</sub>	-10		10	nA						
	offset current 1)		Within T <sub>AMB_TQI</sub>	-2		2	nA						
1.3.2.4.	Common mode input range	V <sub>IN_CM</sub>	Depends on gain adjustment; no XZC; see section 2.3.1	0.29 * VDDA		0.65 * VDDA	V						

#### Table 1.3 Electrical Parameters

No.	Parameter	Symbol	Conditions	Min	Тур	Max	Unit
1.3.3.	Temperature Measurement <sup>2)</sup>						
1.3.3.1.	External temperature diode channel gain	a <sub>TSED</sub>		300		1300	ppm FS / (mV/V)
1.3.3.2.	External temperature diode bias current	I <sub>TSE</sub>		6	10	20	μA
1.3.3.3.	External temperature diode input range <sup>1)</sup>			0		1.5	V
1.3.3.4.	External temperature resistor channel gain	a <sub>TSER</sub>		1200		3500	ppm FS / (mV/V)
1.3.3.5.	External temperature resistor/ input voltage range <sup>1)</sup>	$V_{\text{TSER}}$		0		600	mV/V
1.3.3.6.	Internal temperature diode sensitivity	ST <sub>⊺SI</sub>	Raw values – without conditioning	700		2700	ppm FS / K
1.3.4.	Analog-to-Digital Conversio	n (ADC)					
1.3.4.1.	ADC resolution <sup>1)</sup>	r <sub>ADC</sub>		13		16	Bit
1.3.4.2.	ADC differential nonlinearity (DNL) <sup>1)</sup>		r <sub>ADC</sub> =13-bit; f <sub>CLK</sub> =3MHz;			0.95	LSB
1.3.4.3.	ADC integral nonlinearity (INL) within TQA <sup>1)</sup>	INL <sub>ADC</sub>	best fit, 2nd order; complete AFE; with ADC input range specified in 1.3.4.5			4	LSB
1.3.4.4.	ADC INL within TQE		specified in 1.5.4.5			5	LSB
1.3.4.5.	ADC input range	Range		10		90	%VDDA
1.3.5.	Sensor Connection Check						
1.3.5.1.	Sensor connection loss detection threshold	$R_{SCC\_min}$		100			kΩ
1.3.5.2.	Sensor input short check	$R_{SSC\_short}$	Short detection guaranteed	0		50	Ω
1.3.5.3.	Sensor input no-short threshold	$R_{SSC_{pass}}$	A short is not indicated above this threshold	1000			Ω
1.3.6.	Digital-to-Analog Conversio	n (DAC) a	nd Analog Output (AOUT P	in)			
1.3.6.1.	DAC resolution	r <sub>DAC</sub>	Analog output, 10-90%		12		Bit
1.3.6.2.	Output current sink and	I <sub>SRC/SINK_OUT</sub>	$V_{OUT}$ : 5-95%, $R_{LOAD} \ge 2k\Omega$			2.5	mA
	source for VDDE=5V		$V_{OUT}$ : 10-90%, $R_{LOAD} \ge 1k\Omega$			5	mA
1.3.6.3.	Short circuit current	I <sub>OUT_max</sub>	To VSSE or VDDE 3)	-25		25	mA
1.3.6.4.	Addressable output signal	V <sub>SR_OUT95</sub>	$  R_{\text{LOAD}} \ge 2k\Omega $	0.05		0.95	VDDE
	range	V <sub>SR_OUT90</sub>	$@ R_{LOAD} \ge 1k\Omega$	0.1		0.9	VDDE

No.	Parameter	Symbol	Conditions	Min	Тур	Max	Unit
1.3.6.5.	Output slew rate <sup>1)</sup>	SROUT	C <sub>LOAD</sub> < 50nF	0.1			V/µs
1.3.6.6.	Output resistance in diagnostic mode	R <sub>OUT_DIA</sub>	$\begin{array}{l} \text{Diagnostic Range:} \\ <4 96>\%,\ R_{\text{LOAD}} \geq 2k\Omega \\ <8 92>\%,\ R_{\text{LOAD}} \geq 1k\Omega \end{array}$			82	Ω
1.3.6.7.	Load capacitance 1)	CLOAD	C3 (see section 3)			150	nF
1.3.6.8.	DNL (DAC)	DNLOUT		-1.5		1.5	LSB
1.3.6.9.	INL TQA (DAC) <sup>1)</sup>	INLOUT	Best fit, r <sub>DAC</sub> =12-bit	-5		5	LSB
1.3.6.10.	INL TQE (DAC)	INLOUT	Best fit, r <sub>DAC</sub> =12-bit	-8		8	LSB
1.3.6.11.	Output leak current @150°C	I <sub>LEAK_OUT</sub>	power or ground loss	-25		25	μA
1.3.7.	System Response	I					
1.3.7.1.	Startup time 4)	t <sub>STA</sub>	To 1 <sup>st</sup> output; f <sub>CLK</sub> =3MHz; no ROM check; ADC 14-bit and 2nd order			5	ms
1.3.7.2.	Response time (100% jump) <sup>1)</sup>	t <sub>RESP</sub>	f <sub>CLK</sub> =4MHz; 13-bit, 2nd order; refer to Table 2.3	256		512	μs
1.3.7.3.	Bandwidth <sup>1)</sup>		Comparable to analog SSCs		5		kHz
1.3.7.4.	Analog output noise peak-to-peak <sup>1)</sup>	V <sub>NOISE,PP</sub>	Shorted inputs; bandwidth ≤ 10kHz			10	mV
1.3.7.5.	Analog output noise RMS <sup>1)</sup>	V <sub>NOISE,RMS</sub>	Shorted inputs; bandwidth ≤ 10kHz			3	mV
1.3.7.6.	Ratiometricity error	RE <sub>OUT_5</sub>	Maximum error of VDDE=5V to 4.5/5.5V			1000	ppm
1.3.7.7.	Overall failure (deviation from	F <sub>ALL</sub> TQI	13-bit, 2 <sup>nd</sup> order ADC;		0.25 (0.1)		% FS
	ideal line including the INL, gain, offset and temperature	F <sub>ALL</sub> TQA	f <sub>CLK</sub> ≤3MHz; XZC=0		0.5 (0.25)		% FS
	errors) <sup>5)</sup>	F <sub>ALL</sub> TQE	No sensor caused effects; value in parentheses is the digital readout.		1.0 (0.5)		% FS

1) No measurement in mass production; parameter is guaranteed by design and/or quality observation.

2) Refer to section 2.4.

3) Minimum output voltage to VDDE or maximum output voltage to VSSE.

4) Depends on resolution and configuration - start routine begins approximately 0.8ms after power on.

5) XZC is active: additional overall failure of 25ppm/K for XZC=31 at maximum; failure decreases linearly for XZC adjustments lower than 31.

### 1.4. Interface Characteristics and EEPROM

### Table 1.4 Interface and EEPROM Characteristics

No.	Parameter	Symbol	Conditions	Min	Тур	Max	Unit
1.4.1.	I <sup>2</sup> C <sup>TM</sup> Interface <sup>1)</sup>		· · · · ·				
1.4.1.1	Input-high level 2)	V <sub>I2C_IN_H</sub>		0.8			VDDA
1.4.1.2	Input-low level 2),	V <sub>I2C_IN_L</sub>				0.2	VDDA
1.4.1.3	Output-low level 2)	V <sub>I2C_OUT_L</sub>	Open Drain, I <sub>OL</sub> <2mA			0.15	VDDA
1.4.1.4	SDA load capacitance <sup>2)</sup>	C <sub>SDA</sub>				400	pF
1.4.1.5	SCL clock frequency <sup>2)</sup>	f <sub>SCL</sub>				400	kHz
1.4.1.6	Internal pull-up resistor <sup>2)</sup>	R <sub>I2C</sub>		25		100	kΩ
1.4.2.	ZACwire™ One Wire Interfa	ice (OWI)					
1.4.2.1	Input-low level 2)	V <sub>OWI_IN_L</sub>				0.2	VDDA
1.4.2.2	Input-high level 2)	V <sub>OWI_IN_H</sub>		0.75			VDDA
1.4.2.3	Pull-up resistance master	R <sub>OWI_PUP</sub>		0.3	3.3		kΩ
1.4.2.4	OWI load capacitance	C <sub>OWI_LOAD</sub>	Summarized OWI line load			50	nF
1.4.2.5	Start window <sup>2)</sup>		Typ: @ f <sub>CLK</sub> =3MHz	96	175	455	ms
1.4.3.	EEPROM						
1.4.3.1	Ambient temperature EEPROM programming <sup>2)</sup>	T <sub>AMB_EEP</sub>		-40		150	°C
1.4.3.2	Write cycles <sup>2)</sup>	n <sub>wri_eep</sub>	Write temperature: <=85°C			100k	
			Write temperature: up to 150°C			100	
1.4.3.3	Read cycles <sup>2), 3)</sup>	n <sub>read_eep</sub>	Read temperature: <=175°C			8 * 10 <sup>8</sup>	
1.4.3.4	Data retention <sup>2), 4)</sup>	t <sub>RET_EEP</sub>	1300h at 175°C =100000h at 55°C; 27000h at 125°C; 3000h at 150°C)			15	years
1.4.3.5	Programming time <sup>2)</sup>	t <sub>WRI_EEP</sub>	Per written word, f <sub>CLK</sub> =3MHz		12		ms

2) No measurement in mass production; parameter is guaranteed by design and/or quality observation.

3) Note that the package and temperature versions cause additional restrictions.

4) Over lifetime; use calculation sheet SSC Temperature Profile Calculation Spreadsheet for temperature stress calculation; note additional restrictions are caused by different package and temperature versions.

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# 2 Circuit Description

Note: This data sheet provides specifications and a general overview of ZSC31150 operation. For details of operation, including configuration settings and related EEPROM registers, refer to the *ZSC31150 Functional Description*.

# 2.1. Signal Flow

The ZSC31150's signal path includes both analog (shown in blue in Figure 2.1) and digital (pink) sections. The analog path is differential; i.e., the differential bridge sensor signal is handled internally via two signal lines that are symmetrical around a common mode potential (analog ground = VDDA/2), which improves noise rejection.

Consequently, it is possible to amplify positive and negative input signals, which are located within the common mode range of the signal input.

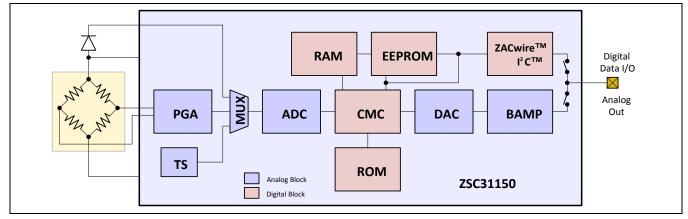


Figure 2.1 Block Diagram of the ZSC31150

PGA	Programmable Gain Amplifier

MUX M	Multiplexer
-------	-------------

- ADC Analog-to-Digital Converter
- CMC Calibration Microcontroller
- DAC Digital-to-Analog Converter
- BAMP Buffer Amplifier Output Buffer OPAMP
- EEPROM Non Volatile Memory for Calibration Parameters and Configuration
- TS On-Chip Temperature Sensor (pn-junction)
- ROM Memory for Correction Formula and Algorithm
- RAM Volatile Memory for Calibration Parameters and Configuration

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The differential signal from the bridge sensor is pre-amplified by the programmable gain amplifier (PGA). The multiplexer (MUX) transmits the signals from either the bridge sensor, the external diode, or the separate temperature sensor to the analog-to-digital converter (ADC) in a specific sequence (the internal pn-junction (TS) can be used instead of the external temperature diode). Next, the ADC converts these signals into digital values.

The digital signal correction takes place in the calibration microcontroller (CMC). It is based on a correction formula located in the ROM and sensor-specific coefficients stored in the EEPROM during calibration. Depending on the programmed output configuration, the corrected sensor signal is output as an analog value or in a digital format (I<sup>2</sup>C<sup>™</sup> or ZACwire<sup>™</sup>). The configuration data and the correction parameters can be programmed into the EEPROM via the digital interfaces.

### 2.2. Application Modes

For each application, a configuration set must be established (generally prior to calibration) by programming the on-chip EEPROM regarding to the following modes:

#### Sensor Channel

- Sensor mode: ratiometric bridge excitation in voltage or current supply mode.
- Input range: the gain adjustment of the AFE with respect to the maximum sensor signal span and the zero point of the ADC have to be chosen.
- An additional analog offset compensation, the Extended Zero-Point Compensation (XZC), must be enabled if required; e.g., if the sensor offset voltage is close to or larger than the sensor span.
- Resolution/response time: The ADC must be configured for resolution and conversion settings (1<sup>st</sup> or 2<sup>nd</sup> order). These settings influence the sampling rate, signal integration time, and, as a result, the noise immunity.

#### Temperature

• Temperature measurement: the source for the temperature correction must be chosen.

# 2.3. Analog Front End (AFE)

The analog front end (AFE) consists of the programmable gain amplifier (PGA), the multiplexer (MUX), and the analog-to-digital converter (ADC).

#### 2.3.1. Programmable Gain Amplifier (PGA)

Table 2.1 shows the adjustable gains, the sensor signal spans, and the allowed common mode range.

 Table 2.1
 Adjustable Gains, Resulting Sensor Signal Spans, and Common Mode Ranges

No.	Overall Gain a <sub>IN</sub>	Max. Span V <sub>IN_SP</sub>	Gain Amp1	Gain Amp2	Gain Amp3	Input common mode rang V <sub>IN_CM</sub> as % of VDDA <sup>2)</sup>	
		[mV/V] <sup>1)</sup>				XZC = Off	XZC = On
1	420	1.8	30	7	2	29 to 65	45 to 55
2	280	2.7	30	4.66	2	29 to 65	45 to 55
3	210	3.6	15	7	2	29 to 65	45 to 55
4	140	5.4	15	4.66	2	29 to 65	45 to 55
5	105	7.1	7.5	7	2	29 to 65	45 to 55
6	70	10.7	7.5	4.66	2	29 to 65	45 to 55
7	52.5	14.3	3.75	7	2	29 to 65	45 to 55
8	35	21.4	3.75	4.66	2	29 to 65	45 to 55
9	26.3	28.5	3.75	3.5	2	29 to 65	45 to 55
10	14	53.75	1	7	2	29 to 65	45 to 55
11	9.3	80	1	4.66	2	29 to 65	45 to 55
12	7	107	1	3.5	2	29 to 65	45 to 55
13	2.8	267	1	1.4	2	32 to 57	not applicable

 Recommended internal signal range maximum is 80% of the VDDA voltage. Span is calculated by the following formula: Span = 80% / gain.

2) Bridge in Voltage Mode with maximum input signal (with XZC = +300% Offset), 14-bit accuracy. Refer to the ZSC31150 Functional Description for usable input signal/common mode range at bridge in current mode. See section 2.3.2 for an explanation of the extended analog zero compensation (XZC).

#### 2.3.2. Offset Compensation

The ZSC31150 supports two methods of sensor offset compensation (zero shift):

- Digital offset correction
- XZC: analog compensation for large offset values (up to a maximum of approximately 300% of the span, depending on the gain adjustment)

The digital sensor offset correction will be processed during the digital signal correction/conditioning by the calibration microcontroller (CMC).

Analog sensor offset pre-compensation is needed for compensation of large offset values, which would overdrive the analog signal path by uncompensated gaining. For analog sensor offset pre-compensation, a compensation voltage is added in the analog pre-gaining signal path (coarse offset removal). The analog offset compensation in the AFE can be adjusted by 6 EEPROM bits (refer to the *ZSC31150 Functional Description* for details).

PGA gain a <sub>IN</sub>	Max. Span V <sub>IN_SP</sub> [mV/V]	Offset shift per step as % of full span	Approximate maximum offset shift [mV/V]	Approximate maximum shift [% V <sub>IN_SP</sub> ] (at ± 31)
420	1.8	12.5 %	7.8	388 %
280	2.7	7.6 %	7.1	237 %
210	3.6	12.5 %	15.5	388 %
140	5.4	7.6 %	14.2	237 %
105	7.1	12.5 %	31	388 %
70	10.7	7.6 %	28	237 %
52.5	14.3	12.5 %	32	388 %
35	21.4	7.6 %	57	237 %
26.3	28.5	5.2 %	52	161 %
14	53.75	12.5 %	194	388 %
9.3	80	7.6 %	189	237 %
7	107	5.2 %	161	161 %
2.8	267	0.83 %	72	26 %

Table 2.2 Analog Zero Point Shift Ranges (XZC)

### 2.3.3. Measurement Cycle

The complete measurement cycle is controlled by the CMC. Depending on EEPROM settings, the multiplexer (MUX) selects the following input signals in a defined sequence:

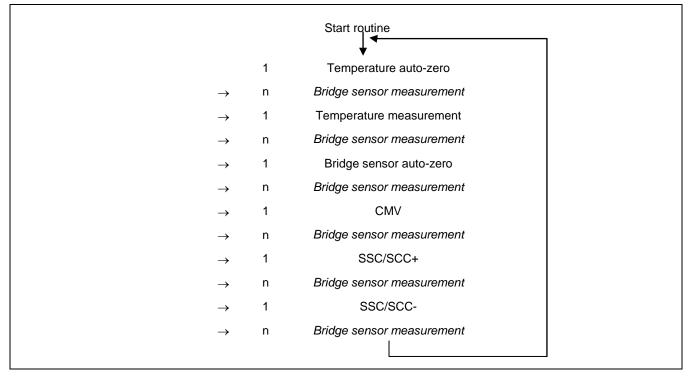
- Temperature measured by external diode or thermistor, internal pn-junction, or bridge
- Internal offset of the input channel (V<sub>OFF</sub>)
- Pre-amplified bridge sensor signal

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The cycle diagram in Figure 2.2 shows the basic structure of the measurement cycle. The bridge sensor measurement count can be configured in EEPROM for a value within n=<1,31>.

After power-on, the startup routine is processed, which performs all measurements needed to acquire an initial valid conditioned sensor output. After the startup routine, the normal measurement cycle runs.

**Note:** The "CMV," "SSC/SCC+" and "SSC/SCC-" measurements are always performed in every cycle independent of the EEPROM configuration.



#### Figure 2.2 Measurement Cycle

#### 2.3.4. Analog-to-Digital Converter

The ADC is an integrating analog-to-digital converter in full differential switched capacitor technique.

Programmable ADC resolutions are  $r_{ADC}$ =<13, 14> or with segmentation,  $r_{ADC}$ =<15, 16> bit.

The ADC can be used as a first or second order converter. In the **first order** mode, it is inherently monotone and insensitive to short and long-term instability of the clock frequency. The conversion cycle time depends on the desired resolution and can be roughly calculated by the following equation where  $r_{ADC}$  is the ADC resolution and  $t_{ADC_1}$  is the conversion cycle time in seconds in first-order mode:

$$\mathbf{t}_{\mathrm{ADC}\_1} = \frac{2^{r_{\mathrm{ADC}}}}{\left(\frac{f_{OSC}}{2}\right)}$$

In the **second order** mode, two conversions are stacked with the advantage of a much shorter conversion cycle time but the drawback of a lower noise immunity caused by the shorter signal integration period. The approximate conversion cycle time  $t_{ADC_2}$  in second-order mode is calculated by the following equation:

$$t_{ADC_2} = \frac{2^{(r_{ADC} + 3)/2}}{\left(\frac{f_{OSC}}{2}\right)}$$

The calculation formulas for  $t_{ADC}$  give an overview of conversion time for one AD conversion. Refer to the *ZSC31150 Bandwidth Calculation Spreadsheet* for detailed calculations for sampling time and bandwidth.

The result of the AD conversion is a relative counter result corresponding to the following equation (see the *ZSC31150 Functional Description* for more detailed equations):

$Z_{ADC} = 2^{r_{ADC}} \left($	$\left(\frac{V_{\text{ADC\_DIFF}}}{V_{\text{ADC\_REF}}} + RS_{\text{ADC}}\right)$	
--------------------------------	---	--

Z <sub>ADC</sub>	Number of counts (result of the conversion)
r <sub>ADC</sub>	Selected ADC resolution in bits
$V_{\text{ADC}\_\text{DIFF}}$	Differential input voltage of the ADC
$V_{\text{ADC}\_\text{REF}}$	Reference voltage of the ADC
	Digital ADC range shift ( $RS_{ADC} = \frac{1}{16}, \frac{1}{8}, \frac{1}{4}, \frac{1}{2},$ controlled by the EEPROM setting)

The sensor input signal can be shifted to the optimal input range of the ADC with the RS<sub>ADC</sub> value.

AD Adjust			ited Output ition 2)	Sample Rate f <sub>CON</sub> <sup>2)</sup>		Averaged Bandwidth at f <sub>CLK</sub>	
ADC Order	r <sub>ADC</sub>	Digital	Analog	f <sub>CLK</sub> =3MHz	f <sub>CLK</sub> =3MHz f <sub>CLK</sub> =4MHz		f <sub>CLK</sub> =4MHz
-	[Bit]	[Bit]	[Bit]	[Hz]	[Hz]	[Hz]	[Hz]
	13	13	12	345	460	130	172
1	14	14	12	178	237	67	89
1	15	14	12	90	120	34	45
-	16	14	12	45	61	17	23
	13	13	12	5859	7813	2203	2937
2	14	14	12	3906	5208	1469	1958
2	15	14	12	2930	3906	1101	1468
	16	14	12	1953	2604	734	979

Table 2.3	Analog Output Resolution versus Sample Rate
10010 2.5	Analog Output Resolution Versus Gample Rate

1) The ADC resolution should be one bit higher than the required output resolution if the AFE gain is adjusted so that more than 50% of the input range is used. Otherwise the ADC resolution should be more than one bit higher than the required output resolution.

2) The sampling rate (A/D conversion time) is only a part of the whole cycle; refer to the *ZSC31150 Bandwidth Calculation Spreadsheet* for detailed information.

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**Note:** The ADC's reference voltage  $ADC_{VREF}$  is defined by the potential between  $\langle VBR_T \rangle$  and  $\langle VBR_B \rangle$  (or  $\langle VDDA \rangle$  to  $\langle VSSA \rangle$ , if selected in EEPROM by the bit CFGAPP:BREF=1). Theoretically, the input range  $ADC_{RANGE_INP}$  of the ADC is equivalent to the ADC's reference voltage.

In practice, the maximum ADC input range used should be from 10% to 90% of ADC<sub>RANGE\_INP</sub>, which is a necessary condition for ensuring the specified accuracy, stability, and nonlinearity parameters of the AFE. This condition is also valid for whole temperature range and all applicable sensor tolerances. The ZSC31150 does not have an internal failsafe function that verifies that the input meets this condition.

#### 2.4. Temperature Measurement

The ZSC31150 supports four different methods for acquiring the temperature data needed for calibration of the sensor signal in the specified temperature range.

Temperature data can be acquired using one of these temperature sensors:

- an internal pn-junction temperature sensor
- an external pn-junction temperature sensor connected to sensor top potential (V<sub>BRTOP</sub>)
- an external resistive half bridge temperature sensor
- the temperature coefficient of the sensor bridge at bridge current excitation

Refer to the ZSC31150 Functional Description for a detailed explanation of temperature sensor adaptation and adjustment.

#### 2.5. System Control and Conditioning Calculation

The system control supports the following tasks/features:

- Controlling the measurement cycle according to the EEPROM-stored configuration data
- Performing the16-bit correction calculation for each measurement signal using the EEPROM-stored calibration coefficients and ROM-based algorithms; i.e., the signal conditioning
- · Managing the start-up sequence and starting signal conditioning
- Handling communication requests received by the digital interface
- Managing failsafe tasks for the functions of the ZSC31150 and indicating detected errors with diagnostic states

Refer to the ZSC31150 Functional Description for a detailed description.

#### 2.5.1. Operation Modes

The internal state machine has three main states:

- The continuously running signal conditioning mode, which is called Normal Operation Mode (NOM)
- The calibration mode with access to all internal registers and states, which is called Command Mode (CM)
- The failure messaging mode, which is called Diagnostic Mode (DM)

### 2.5.2. Start Up Phase<sup>1</sup>

The start-up phase consists of following segments:

- 1. Internal supply voltage settling phase (i.e., the VDDA VSSA potential), which is ended when the reset signal is disabled through the power-on clear block (POR). Refer to the *ZSC31150 Technical Note High Voltage Protection* document, section 4 for power on/off thresholds.
- Time (from beginning with VDDA-VSSA=0V): 500µs to 2000µs; AOUT is in tri-state
  System start, EEPROM read out, and signature check (and ROM check if selected by setting EEPROM bit CFGAPP:CHKROM=1).
  - Time: ~200µs (~9000µs with ROM-check; i.e., 28180 clocks); AOUT is LOW (DM)
- 3. Processing the start routine for signal conditioning (all measurements and conditioning calculations). **Time:** 5 x A/D conversion time; AOUT behavior depends on selected OWI mode (refer to section 2.6):
  - OWIANA & OWIDIS => AOUT is LOW (DM)
  - OWIWIN & OWIENA => AOUT is in tri-state

The analog output AOUT will be activated at the end of the start-up phase depending on the adjusted output and communication mode (refer to section 2.6). If errors are detected, the Diagnostic Mode (DM) is activated and the diagnostic output signal is driven at the output.

After the start-up phase, the continuously running measurement and calibration cycle is started. Refer to *ZSC31150 Bandwidth Calculation Spreadsheet* for detailed information about output update rate.

#### 2.5.3. Conditioning Calculation

The digitalized value for the bridge sensor measurement (acquired raw data) is processed with the correction formula to remove offset and temperature dependency and to compensate nonlinearity up to 3rd order. The result of the correction calculation is a non-negative 15-bit value for the bridge sensor in the range [0; 1). This value P is clipped with programmed limitation coefficients and continuously written to the output register of the digital serial interface and the output DAC.

**Note:** The conditioning includes up to third-order nonlinearity sensor input correction. The available adjustment ranges depend on the specific calibration parameters; for a detailed description, refer to *ZSC31150 Functional Description*. Basically, offset compensation and linear correction are only limited by the loss of resolution they will cause. The second-order correction is possible up to approximately 30% of the full scale difference from a straight line; third order is possible up to approximately 20% (ADC resolution = 13-bit). The calibration principle used is able to reduce existing nonlinearity errors of the sensor up to 90%. The temperature calibration includes first and second order correction and should be fairly sufficient in all relevant cases. ADC resolution also influences calibration possibilities; e.g., 1 additional bit of resolution reduces the calibration range by approximately 50%. The maximum calculation input data width is 14-bit. The 15 or 16 bit ADC resolution mode uses only a 14-bit segment of the ADC range.

<sup>1</sup> All timings described are roughly estimated values and are affected by the internal clock frequency. Timings are estimated for f<sub>CLK</sub>=3MHz.

### 2.6. Analog Output AOUT

The analog output is used for outputting the analog signal conditioning result and for "end of line" communication via the ZACwire<sup>™</sup> interface one-wire communication interface (OWI). The ZSC31150 supports four different modes of the analog output in combination with the OWI behavior:

- OWIENA: Analog output is deactivated; OWI communication is enabled.
- OWIDIS: Analog output is active (~2ms after power-on); OWI communication is disabled.
- OWIWIN: Analog output will be activated after the time window; OWI communication is enabled in a time window of ~500ms (maximum); transmission of the "START\_CM" command must be finished during the time window.
- OWIANA: Analog output will be activated after a ~2ms power on time; OWI communication is enabled in a time window of ~500ms (maximum); transmission of the START\_CM" command must be finished during time window; to communicate, the internal driven potential at AOUT must be overwritten by the external communication master (AOUT drive capability is current limited).

The analog output potential is driven by a unity gain output buffer for which the input signal is generated by a 12.4-bit resistor-string DAC. The output buffer (BAMP), which is a rail-to-rail op amp, is offset compensated and current limited. Therefore, a short-circuit of the analog output to ground or the power supply does not damage the ZSC31150.

# 2.7. Serial Digital Interface

The ZSC31150 includes a serial digital interface (SIF), which is used for communication with the circuit to calibrate the sensor module. The serial interface is able to communicate with two communication protocols:  $I^2C^{TM}$  and the ZACwire<sup>TM</sup> one-wire communication interface (OWI). The OWI can be used to for an "end of line" calibration via the analog output AOUT of the complete assembled sensor module.

Refer to the ZSC31150 Functional Description for a detailed description of the serial interfaces and communication protocols.

### 2.8. Failsafe Features, Watchdog and Error Detection

The ZSC31150 detects various possible errors. A detected error is indicated by a change in the internal status in Diagnostic Mode (DM). In this case, the analog output is set to LOW (minimum possible output value; i.e., the lower diagnostic range LDR) and the output registers of the digital serial interface are set to a significant error code.

A watchdog oversees the continuous operation of the CMC and the running measurement loop. The operation of the internal clock oscillator is verified continuously by the oscillator failure detection.

A check of the sensor bridge for broken wires is done continuously by two comparators watching the input voltage of each input (sensor connection and short check). Additionally, the common mode voltages of the sensor and sensor input short are watched continuously (sensor aging).

Different functions and blocks in the digital section, e.g. the RAM, ROM, EEPROM, and register content, are watched continuously. Refer to the *ZSC31150 Functional Description* for a detailed description of safety features and methods of error indication.

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### 2.9. High Voltage, Reverse Polarity, and Short Circuit Protection

The ZSC31150 is designed for 5V power supply operation.

The ZSC31150 and the connected sensor are protected from overvoltage and reverse polarity damage by an internal supply voltage limiter. The analog output AOUT can be connected with all potentials (short circuit, overvoltage, and reverse voltage) in the protection range under all potential conditions at the VDDE and VSSE pins.

All external components (see section 3) are required to guarantee this operation. The protection is not time limited. Refer the *ZSC31150 Technical Note – High Voltage Protection* for a detailed description of protection cases and conditions.

# 3 Application Circuit Examples

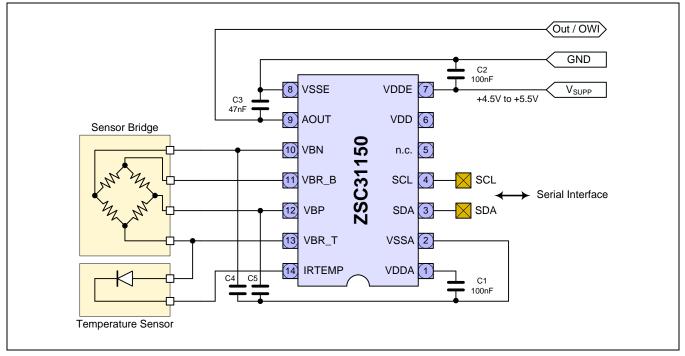
The application circuits contain external components that are needed for over-voltage, reverse polarity, and short circuit protection.

**Recommendation:** Check the ZSC31150 product page <u>www.IDT.com/ZSC31150</u> for other application examples given in application notes. Note: Some application notes require a customer login—see section 9 for details.

Symbol	Parameter	Min	Тур	Мах	Unit	Notes
C1	С	100		470	nF	
C2	С	100			nF	
C3 <sup>1)</sup>	С	4	47	160	nF	The value of C3 is the sum of the load capacitor and the cable capacitance
C4, C5 <sup>1)</sup>	С	0		10	nF	Recommended to increase EMC immunity.
R1			10		kΩ	
R <sub>IBR</sub>	R	Ref	er to section	1.2.	Ω	
1) Higher values for C3, C4, and C5 increase EMC immunity.						

 Table 3.1
 Application Circuit Parameters





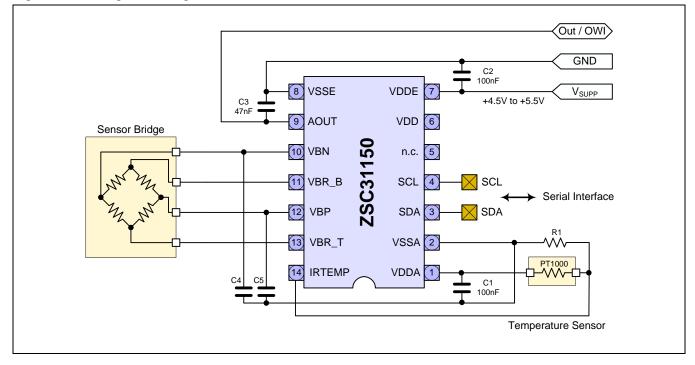
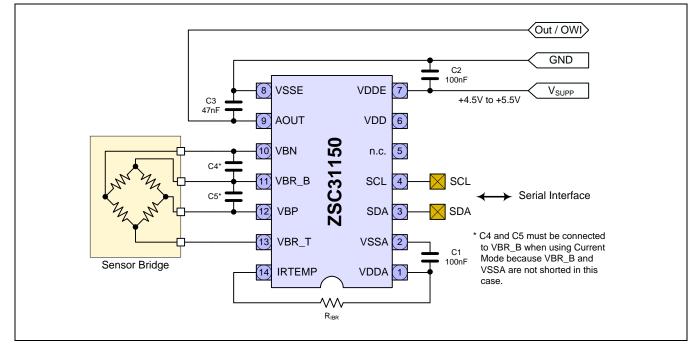


Figure 3.2 Bridge in Voltage Mode, External Thermistor

Figure 3.3 Bridge in Current Mode, Temperature Measurement via Bridge TC



#### 4 Pin Configuration, Latch-Up and ESD Protection

#### Pin Configuration and Latch-up Conditions 4.1.

Table 4.1	Pin Configuration and Latch-Up Conditions
-----------	---

Pin	Name	Description	Notes	Usage/ Connection <sup>1)</sup>	Latch-up Related Application Circuit Restrictions and/or Notes			
1	VDDA	Positive analog supply voltage	Analog IO	Required/-				
2	VSSA	Negative analog supply voltage	Analog IO	Required/-				
3	SDA	I²C <sup>™</sup> data IO	Digital IO, pull-up	-/VDDA	Trigger Current/Voltage to VDDA/VSSA:			
4	SCL	I²C™ clock	Digital IN, pull-up	-/VDDA	+/-100mA or 8/-4V			
5	N.C.	No connection						
6	VDD	Positive digital supply voltage	Analog IO	Required or open/-	Only capacitor to VSSA is allowed, otherwise no application access			
7	VDDE	Positive external supply voltage	Supply	Required/-	Trigger Current/Voltage: -100mA/33V			
8	VSSE	Negative external supply voltage	Ground	Required/-				
9	AOUT	Analog output & one wire IF IO	Ю	Required/-	Trigger Current/Voltage: -100mA/33V			
10	VBN	Negative input sensor bridge	Analog IN	Required/-				
11	VBR_B	Bridge bottom potential	Analog IO	Required/VSSA	Depending on application circuit, short to VDDA/VSSA possible			
12	VBP	Positive input sensor bridge	Analog IN	Required/-				
13	VBR_T	Bridge top potential	Analog IO	Required/VDDA				
14	IRTEMP	Temp sensor & current source resistor	Analog IO	-/VDDA, VSSA	Depending on application circuit			
	<ol> <li>Usage: If "Required" is specified, an electrical connection is necessary; refer to the application circuits in section 3.</li> <li>Connection: To be connected to this potential if not used or if no application/configuration-related constraints are given.</li> </ol>							

Connection: To be connected to this potential if not used or if no application/configuration-related constraints are given.

### 4.2. ESD Protection

All pins have an ESD protection of >2000V. Additionally, the pins VDDE, VSSE and AOUT have an ESD protection of >4000V.

ESD protection referenced to the Human Body Model is tested with devices during product qualification. The ESD test follows the Human Body Model with  $1.5k\Omega/100pF$  based on MIL 883, Method 3015.7.

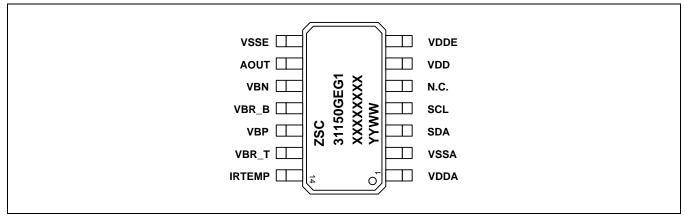
# 5 Package

### 5.1. SSOP14 Package

The standard packages of the ZSC31150 are the SSOP14 green package (5.3mm body width) with a lead pitch of 0.65mm and the DFN14 (4mmx5mm) package with a lead pitch of 0.5mm.

For the SSOP14 package markings shown in Figure 5.1, YYWW refers to the last two digits of the year (YY) and two digits for the work-week designation (WW). XXXXXXX refers to the lot number.

#### Figure 5.1 SSOP14 Pin Diagram

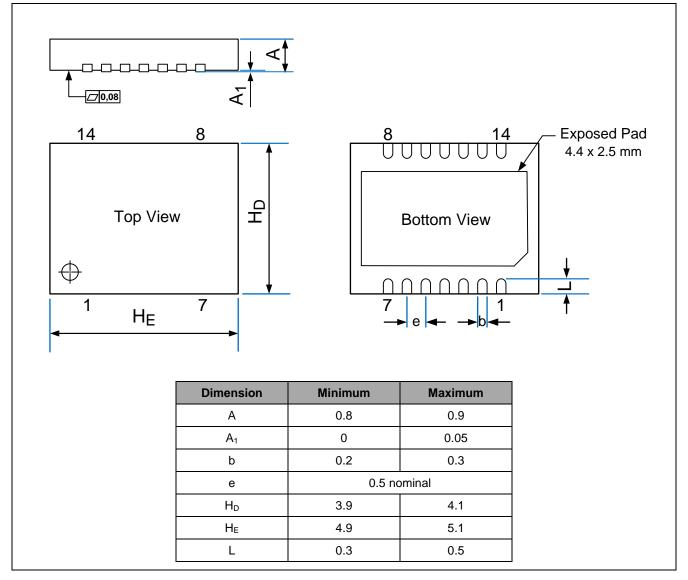


### 5.2. DFN14 Package

For the DFN14 package, the pin assignment is the same as in SSOP14. Refer to the ZSC31150 Technical Note – Die and Package Specifications for a description of package markings.

Figure 5.2 provides the dimensions for the DFN14 package option, which are based on JEDEC MO-229. The DFN14 package has wettable flanks.





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# 6 Quality and Reliability

The ZSC31150 is qualified according to the AEC-Q100 standard, operating temperature grade 0. A fit rate < 5fit (temperature =55°C, S=60%) is guaranteed. A typical fit rate of the C7D technology, which is used for ZSC31150, is 2.5fit.

# 7 Customization

For high-volume applications, which require an upgraded or downgraded functionality compared to the standard ZSC31150, IDT can customize the circuit design by adding or removing certain functional blocks.

For this purpose, IDT has a considerable library of sensor-dedicated circuitry blocks. As a result, IDT can provide a custom solution quickly. Please contact IDT for further information.

# 8 Ordering Information

Product Sales Code	Description	Package		
ZSC31150GEB	ZSC31150 Die — Temperature range: -40°C to +150°C	Unsawn on Wafer		
ZSC31150GEC	ZSC31150 Die — Temperature range: -40°C to +150°C	Sawn on Wafer Frame		
ZSC31150GED	ZSC31150 Die — Temperature range: -40°C to +150°C	Waffle Pack		
ZSC31150GEG2-R	ZSC31150 DFN14, 5x4mm with wettable flank—Temperature range: -40°C to 150°C	Tape & Reel		
ZSC31150GAG2-R	ZSC31150 DFN14, 5x4mm with wettable flank —Temperature range: -40°C to 125°C	Tape & Reel		
ZSC31150GEG1	ZSC31150 SSOP14—Temperature range: -40°C to +150°C	Tube: add "-T" to sales code Tape & Reel: add "-R"		
ZSC31150GLG1	ZSC31150 SSOP14—Temperature range: -40°C to +150°C (Long life: 5000h @150°C)	Tube: add "-T" to sales code Tape & Reel: add "-R"		
ZSC31150GAG1	ZSC31150 SSOP14—Temperature range: -40°C to +125°C	Tube: add "-T" to sales code Tape & Reel: add "-R"		
ZSC31150KIT Evaluation Kit V1.2	SC31150 SSC Evaluation Kit: three interconnecting boards, five ZSC31150 SSOP14 samples, ISB cable (software can be downloaded from product page at <u>www.IDT.com/ZSC31150</u> )			
ZSC31150 Mass Calibration System V1.1	Modular Mass Calibration System (MSC) for ZSC31150: MCS boards, cable, connectors (software can be downloaded from product page at <a href="http://www.IDT.com/ZSC31150">www.IDT.com/ZSC31150</a> )			

# 9 Related Documents and Tools

Note: Documents marked with an asterisk (\*) require a login account for access on the web. Note: Documents marked with a double asterisk (\*\*) are only available upon request.

Document/Tool		
ZSC31150 Feature Sheet		
ZSC31150 Functional Description		
ZSC31150 Evaluation Kit Description		
SSC AN - Single Ended Input		
ZSC31150 Technical Note – High Voltage Protection *		
SSC Temperature Profile Calculation Spreadsheet *		
ZSC31150 Technical Note – Die and Package Specifications **		
ZSC31150 Bandwidth Calculation Spreadsheet **		
CAD Model Library Files for IDT SSC ICs		

Visit the ZSC31150 product page <u>www.IDT.com/ZSC31150</u> or contact your nearest sales office for the latest version of these documents.

# 10 Glossary

Term	Description		
ADC	Analog-to-Digital Converter		
AEC	Automotive Electronics Council		
AFE	Analog Front End		
AOUT	Analog Output		
BAMP	Buffer Amplifier		
СМ	Command Mode		
CMC	Calibration Microcontroller		
CMV	Common Mode Voltage		
CMOS	Complementary Metal Oxide Semiconductor		
DAC	Digital-to-Analog Converter		
DM	Diagnostic Mode		
EEPROM	Electrically Erasable Programmable Read Only Memory		
ESD	Electrostatic Device		

Term	Description		
LDR	Lower Diagnostic Range		
MUX	Multiplexer		
NOM	Normal Operation Mode		
OWI	One Wire Interface		
Ρ	Bridge Sensor Measurement; e.g., Pressure Sensor		
PGA	Programmable Gain Amplifier		
POC	Power on Clear		
RAM	Random-Access Memory		
RISC	Reduced Instruction Set Computer		
RMS	Root-Mean-Square		
ROM	Read Only Memory		
SCC	Sensor Connection Check		
SIF	Serial Interface		
SSC+	Positive-biased Sensor Short Check		
SSC-	Negative-biased Sensor Short Check		
TS	Temperature Sensor		
XZC	eXtended Zero Compensation		

# 11 Document Revision History

Revision	Date	Description
0.46	June 12, 2008	First release after format update
0.47	July 20, 2008	Update after review
1.01	September 20, 2008	Section 6: fit rate added. Section 1.5.2: ROM check time revised/corrected. Section 5.3.4.3: SC – no detection limit added
1.02	September 20, 2009	Update to new ZMDI template
1.03	October 2, 2009	Update to ZMDI denotation
1.04	October 22, 2009	Formatting and linking issues solved
1.05	February 26, 2010	Update for ZMDI template, including ZSC31150 Feature Sheet at page 2&3 Added ordering codes for ZSC31150 and evaluation kits. Extended glossary. Update for contact information.
1.06	July 29, 2010	Correct "Offset shift per step" and "Approx. maximum offset shift" in Table 2.2 for PGA gain = 105 and 52.5. Moved 1.4.1.6 "Internal pull-up resistor" into section 1.4.1 in Table 1.2. Redrew of Sensor Bridge in Figure 3.1, Figure 3.2 and Figure 3.3. Added comment for C4 and C5 in Figure 3.3. Renamed ZMD31150 as ZSC31150.

Revision	Date	Description
1.07	August 31, 2010	Connection of R <sub>IBR</sub> in Figure 3.3 corrected
1.08	August 15, 2011	Update ordering information with "Long Life Automotive" in "Ordering Information" on page 3 and section 8)
2.00	December 15, 2012	Update for part numbers and IDT contact information. Minor edits.
2.10	March 31, 2014	Revision of specifications in section 1.4.2. Recommended internal signal range revised to 80%. OWI interface parameters list extended. ADC formula corrected. DFN14 package added. Minor edits for clarity. Updated contact information. Updated imagery for cover and headings.
2.20	April 30, 2014	Added notation that DFN14 package has wettable flanks. Update for contact information and addition of CAD model files to section 9.
2.30	August 27, 2014	Minor edits on page 2. Minor edits for die description in part code tables.
2.40	December 3, 2014	Corrected connection of temperature PTC sensor in Figure 3.2 Update for contact information.
2.41	July 27, 2015	Update for order code for ZSC31150 SSC Evaluation Kit order code. Update for contact information.
	January 29, 2016	Changed to IDT branding.



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