## *High-Temperature Chopper-Stabilized Precision Hall-Effect Latches for 5 V Applications*

FEATURES AND BENEFITS

- Optimized for applications with regulated power rails
  □ Operation from 2.8 to 5.5 V
- AEC-Q100 automotive qualified
- Operation up to 175°C junction temperature
- Dynamic offset cancellation
  □ Resistant to physical stress
  □ Superior temperature stability
- Symmetrical latch switchpoints
- Output short-circuit protection
- Solid-state reliability
- Industry-standard packages and pinouts

PACKAGES:

Not to scale
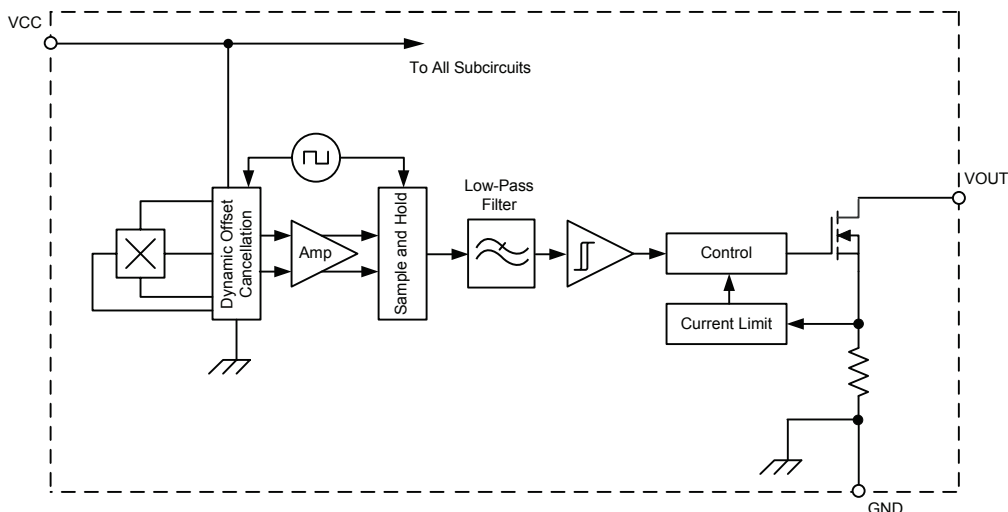
3-pin SOT23W
(suffix LH)

3-pin SIP
(suffix UA)
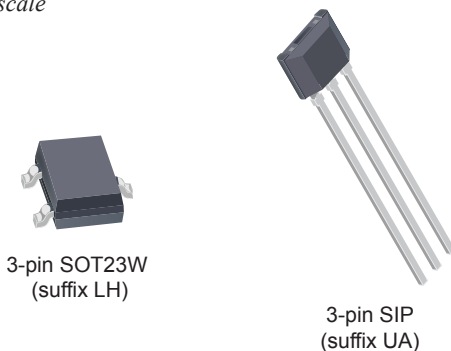
DESCRIPTION

The APS12205, APS12215, and APS12235 Hall-effect sensor ICs are extremely temperature-stable and stress-resistant devices especially suited for operation over extended junction temperature ranges up to 175°C. Superior high-temperature performance is made possible through dynamic offset cancellation, which reduces the residual offset voltage normally caused by device overmolding, temperature dependencies, and thermal stress.

The single silicon chip includes: a Hall plate, small signal amplifier, chopper stabilization, Schmitt trigger, and a short-circuit-protected open-drain output to sink up to 10 mA. A south pole of sufficient strength turns the output on; a north pole of sufficient strength is necessary to turn the output off. For applications requiring operation from greater than 5.5 V or operation directly from a battery, refer to the A1220, A1221, or A1223.

Two package styles provide a choice of through-hole or surface mounting. Package type LH is a modified 3-pin SOT23W surface-mount package, while UA is a three-pin ultramini SIP for through-hole mounting. Both packages are lead (Pb) free, with 100% matte-tin-plated leadframes.

Functional Block Diagram

SPECIFICATIONS

SELECTION GUIDE

Part Number	Packing [1]	Mounting	Branding	Ambient, T_A	B_{RP} (Min)	B_{OP} (Max)
APS12205LLHALX	13-in. reel, 10000 pieces/reel	3-pin SOT23W surface mount	A04			
APS12205LLHALT [2]	7-in. reel, 3000 pieces/reel	3-pin SOT23W surface mount	A04	−40°C to 150°C	−40 G	40 G
APS12205LUAA	Bulk, 500 pieces/bag	3-pin SIP through hole	A18			
APS12215LLHALX	13-in. reel, 10000 pieces/reel	3-pin SOT23W surface mount	A01			
APS12215LLHALT [2]	7-in. reel, 3000 pieces/reel	3-pin SOT23W surface mount	A01	−40°C to 150°C	−90 G	90 G
APS12215LUAA	Bulk, 500 pieces/bag	3-pin SIP through hole	A03			
APS12235LLHALX [3]	13-in. reel, 10000 pieces/reel	3-pin SOT23W surface mount	A35			
APS12235LLHALT [2,3]	7-in. reel, 3000 pieces/reel	3-pin SOT23W surface mount	A35	−40°C to 150°C	−180 G	180 G
APS12235LUAA [3]	Bulk, 500 pieces/bag	3-pin SIP through hole	A36			

[1] Contact Allegro for additional packing options.
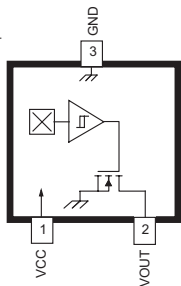[2] Available through authorized Allegro distributors only.
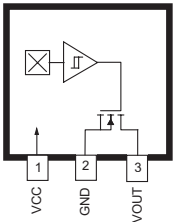[3] Contact Allegro for availability.

ABSOLUTE MAXIMUM RATINGS

Characteristic	Symbol	Notes	Rating	Units
Forward Supply Voltage	V_{CC}		6	V
Reverse Supply Voltage	V_{RCC}		−0.3	V
Output Off Voltage	V_{OUT}		6	V
Continuous Output Current	I_{OUT}		10	mA
Maximum Junction Temperature	T_J(max)		165	°C
		For 500 hours	175	°C
Storage Temperature	T_{stg}		−65 to 170	°C

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PINOUT DIAGRAMS AND TERMINAL LIST TABLE

Package LH

Package UA

Terminal List

Name	Description	Number	
		Package LH	Package UA
VCC	Connects power supply to chip	1	1
VOUT	Output from circuit	2	3
GND	Ground	3	2

ELECTRICAL CHARACTERISTICS: Valid over full operating voltage and T_A = –40°C to 150°C, unless otherwise noted

Characteristics	Symbol	Test Conditions	Min.	Typ.[1]	Max.	Unit[2]
ELECTRICAL CHARACTERISTICS						
Forward Supply Voltage	V_{CC}	Operating, T_J < 175°C	2.8	–	5.5	V
Supply Current	$I_{CC(ON)}$	V_{CC} = 5.5 V, B > B_{OP}	–	–	4	mA
	$I_{CC(OFF)}$	V_{CC} = 5.5 V, B < B_{RP}	–	–	4	mA
Output Leakage Current	I_{OUTOFF}	V_{OUT} = 5.5 V, B < B_{RP}	–	–	10	µA
Output Saturation Voltage	$V_{OUT(SAT)}$	I_{OUT} = 5 mA, B > B_{OP}	–	–	500	V
Output Off Voltage	V_{OUTOFF}	B < B_{RP}	–	–	6	V
Output Short-Circuit Current Limit	I_{OM}		30	–	60	mA
Power-On Time [3]	t_{PO}	V_{CC} ≥ 2.8 V, B < B_{RP}(min) – 10 G, B > B_{OP}(max) + 10 G	–	–	25	µs
Power-On State, Output [3]	POS	V_{CC} ≥ V_{CC}(min), t < t_{ON}		Low		–
Chopping Frequency	f_C		–	800	–	kHz
Output Rise Time [3,4]	t_r	R_L = 1 kΩ, C_L = 20 pF	–	0.2	2	µs
Output Fall Time [3,4]	t_f	R_L = 1 kΩ, C_L = 20 pF	–	0.1	2	µs
MAGNETIC CHARACTERISTICS						
Operate Point	B_{OP}	APS12205	5	22	40	G
		APS12215	15	50	90	G
		APS12235	100	150	180	G
Release Point	B_{RP}	APS12205	–40	–22	–5	G
		APS12215	–90	–50	–15	G
		APS12235	–180	–150	–100	G
Hysteresis	B_{HYS}	APS12205	10	45	80	G
		APS12215 (B_{OP} – B_{RP})	30	100	180	G
		APS12235	200	300	360	G

[1] Typical data are are at T_A = 25°C and V_{CC} = 5 V, and are for initial design estimations only.
[2] 1 G (gauss) = 0.1 mT (millitesla).
[3] Guaranteed by device design and characterization.
[4] C_L = oscilloscope probe capacitance.

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THERMAL CHARACTERISTICS: May require derating at maximum conditions; see application information

Characteristic	Symbol	Test Conditions	Value	Units
Package Thermal Resistance	$R_{\theta JA}$	Package LH, 1-layer PCB with copper limited to solder pads	228	°C/W
		Package LH, 2-layer PCB with 0.463 in² of copper area each side connected by thermal vias	110	°C/W
		Package UA, 1-layer PCB with copper limited to solder pads	165	°C/W

Power Derating Curve

$T_{J(max)} = 175°C; I_{CC} = I_{CC(max)}, I_{OUT} = 0$ mA (Output Off)
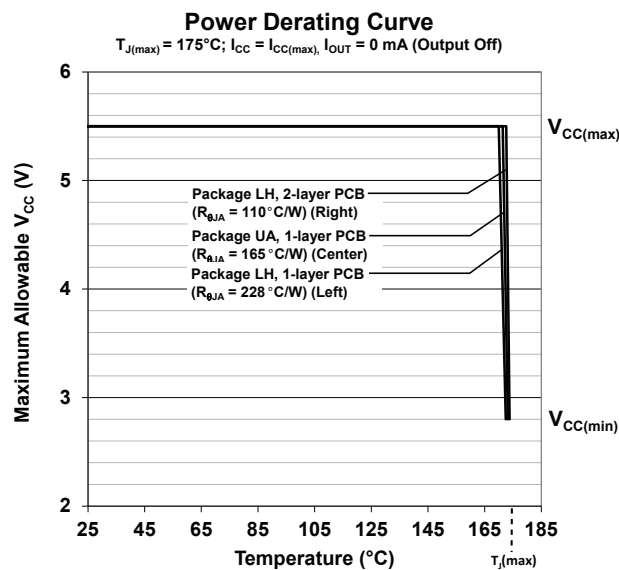
Package Power Dissipation versus Ambient Temperature
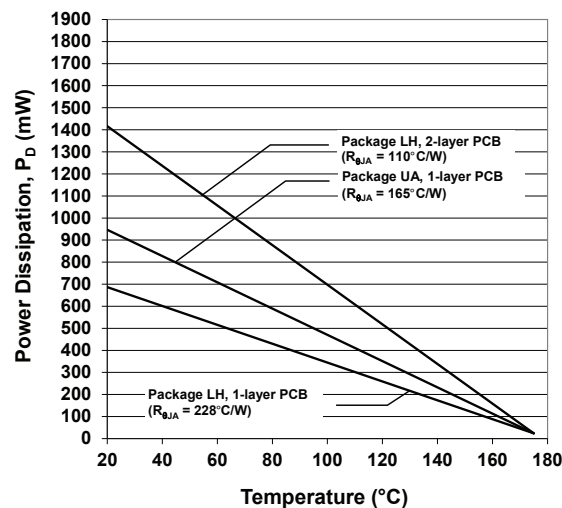
FUNCTIONAL DESCRIPTION

OPERATION

The output of these devices switches low (turns on) when a magnetic field perpendicular to the Hall element exceeds the operate point threshold, B_{OP} (see Figure 1). After turn-on, the output voltage is $V_{OUT(SAT)}$. The output transistor is capable of continuously sinking up to 10 mA, as noted in the Absolute Maximum Ratings specification table. When the magnetic field is reduced below the release point, B_{RP}, the device output goes high (turns off). The difference in the magnetic operate and release points is the hysteresis, B_{HYS}, of the device. This built-in hysteresis allows clean switching of the output even in the presence of external mechanical vibration and electrical noise.

Removal of the magnetic field will leave the device output latched on if the last crossed switchpoint is B_{OP}, or latched off if the last crossed switch point is B_{RP}.
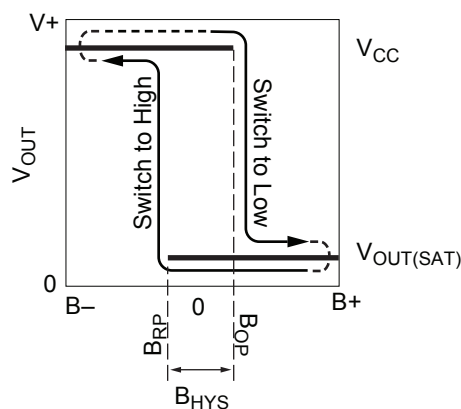

Figure 1: Switching Behavior of Latches

On the horizontal axis, the B+ direction indicates increasing south polarity magnetic field strength, and the B– direction indicates decreasing south polarity field strength (including the case of increasing north polarity.

POWER-ON BEHAVIOR

Device power-on occurs once t_{ON} has elapsed. During the time prior to t_{ON}, and after $V_{CC} > V_{CC}(min)$, the output state is $V_{OUT(SAT)}$(Low). After t_{ON} has elapsed, the output will correspond with the applied magnetic field for $B > B_{OP}$ or $B < B_{RP}$. See Figure 2 for an example.

Powering-on the device in the hysteresis range (less than B_{OP} and higher than B_{RP}) will give an output state of $V_{OUT(SAT)}$. The correct state is attained after the first excursion beyond B_{OP} or B_{RP}.
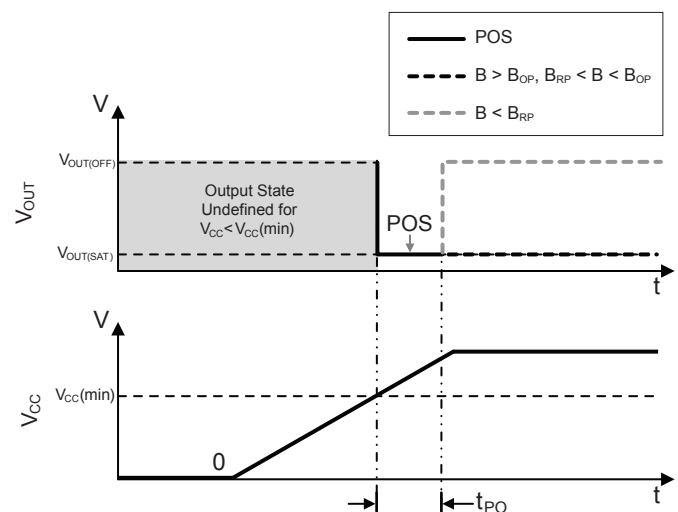

Figure 2: Power-On Timing Diagram

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APPLICATIONS

It is strongly recommended that an external bypass capacitor be connected (in close proximity to the Hall element) between the supply and ground of the device to reduce both external noise and noise generated by the chopper stabilization technique. As is shown in Figure 3, a 0.1 µF capacitor is typical.

Extensive applications information on magnets and Hall-effect sensors is available in:

• *Hall-Effect IC Applications Guide, AN27701,*

• *Hall-Effect Devices: Guidelines for Designing Subassemblies Using Hall-Effect Devices AN27703.1*

• *Soldering Methods for Allegro's Products – SMD and Through-Hole, AN26009*

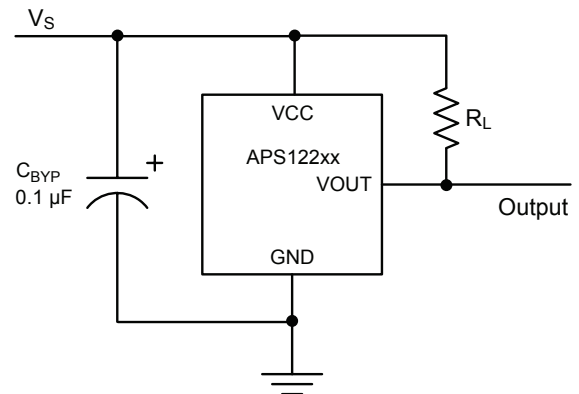All are provided on the Allegro website:

www.allegromicro.com

Figure 3: Typical Application Circuit

CHOPPER STABILIZATION

A limiting factor for switchpoint accuracy when using Hall-effect technology is the small signal voltage developed across the Hall plate. This voltage is proportionally small relative to the offset that can be produced at the output of the Hall sensor. This makes it difficult to process the signal and maintain an accurate, reliable output over the specified temperature and voltage range. Chopper stabilization is a proven approach used to minimize Hall offset.

The Allegro patented technique, dynamic quadrature offset cancellation, removes key sources of the output drift induced by temperature and package stress. This offset reduction technique is based on a signal modulation-demodulation process. Figure 4 illustrates how it is implemented.

The undesired offset signal is separated from the magnetically induced signal in the frequency domain through modulation. The subsequent demodulation acts as a modulation process for the offset, causing the magnetically induced signal to recover its original spectrum at baseband while the DC offset becomes a high-frequency signal. Then, using a low-pass filter, the signal passes while the modulated DC offset is suppressed. Allegro's innovative chopper stabilization technique uses a high-frequency clock. The high-frequency operation allows a greater sampling rate that produces higher accuracy, reduced jitter, and faster signal processing. Additionally, filtering is more effective and results in a lower noise analog signal at the sensor output. Devices such as the APS12205, APS12215, and APS12235 that use this approach have an extremely stable quiescent Hall output voltage, are immune to thermal stress, and have precise recoverability after temperature cycling. This technique is made possible through the use of a BiCMOS process which allows the use of low offset and low noise amplifiers in combination with high-density logic and sample-and-hold circuits.
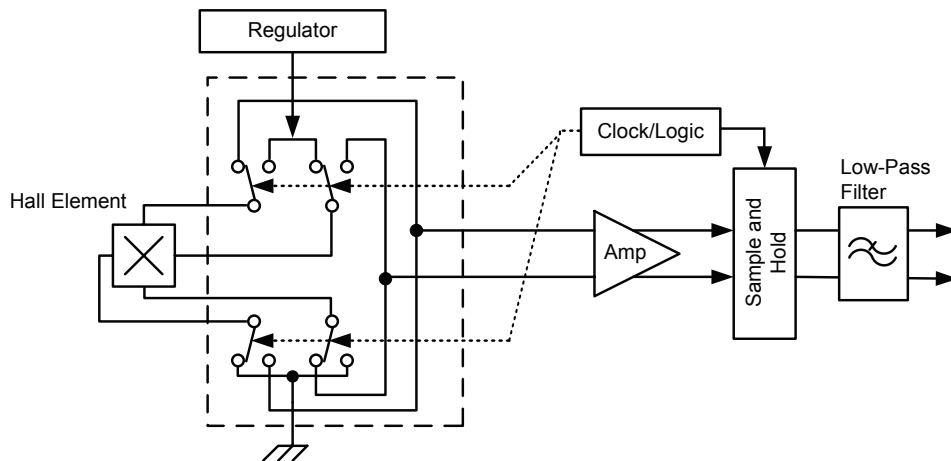

**Figure 4: Model of Chopper Stabilization
(Dynamic Offset Cancellation)**

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POWER DERATING

The device must be operated below the maximum junction temperature of the device, $T_J(max)$. Under certain combinations of peak conditions, reliable operation may require derating supplied power or improving the heat dissipation properties of the application. This section presents a procedure for correlating factors affecting operating T_J. (Thermal data is also available on the Allegro MicroSystems website.)

The Package Thermal Resistance, $R_{\theta JA}$, is a figure of merit summarizing the ability of the application and the device to dissipate heat from the junction (die), through all paths to the ambient air. Its primary component is the Effective Thermal Conductivity, K, of the printed circuit board, including adjacent devices and traces. Radiation from the die through the device case, $R_{\theta JC}$, is relatively small component of $R_{\theta JA}$. Ambient air temperature, T_A, and air motion are significant external factors, damped by overmolding.

The resulting power dissipation capability directly reflects upon the ability of the device to withstand extreme operating conditions. The junction temperature mission profile specified in the Absolute Maximum Ratings table designates a total operating life capability based on qualification for the most extreme conditions, where T_J may reach 175°C.

The silicon IC is heated internally when current is flowing into the VCC terminal. When the output is on, current sinking into the VOUT terminal generates additional heat. This may increase the junction temperature, T_J, above the surrounding ambient temperature. The APS12205, APS12215, and APS12235 are permitted to operate up to T_J = 175°C. As mentioned above, an operating device will increase T_J according to equations 1, 2, and 3 below. This allows an estimation of the maximum ambient operating temperature.

$$P_D = V_{IN} \times I_{IN} \qquad (1)$$

$$\Delta T = P_D \times R_{\theta JA} \qquad (2)$$

$$T_J = T_A + \Delta T \qquad (3)$$

For example, given common conditions such as: T_A = 25°C, V_{CC} = 5 V, I_{CC} = 2.5 mA, V_{OUT} = 185 mV, I_{OUT} = 2 mA (output on), and $R_{\theta JA}$ = 165°C/W, then:

$$\begin{aligned}
P_D &= (V_{CC} \times I_{CC}) + (V_{OUT} \times I_{OUT}) = \\
&\quad (5\ V \times 2.5\ mA) + (185\ mV \times 2\ mA) = \\
&\quad 12.5\ mW + 0.4\ mW = 12.9\ mW
\end{aligned}$$

$$\Delta T = P_D \times R_{\theta JA} = 12.9\ mW \times 165°C/W = 2.1°C$$

$$T_J = T_A + \Delta T = 25°C + 2.1°C = 27.1°C$$

A worst-case estimate, $P_D(max)$, represents the maximum allowable power level ($V_{CC}(max)$, $I_{CC}(max)$), without exceeding $T_J(max)$, at a selected $R_{\theta JA}$.

For example, given the conditions $R_{\theta JA}$ = 228°C/W, $T_J(max)$ = 175°C, $V_{CC}(max)$ = 5.5 V, $I_{CC}(max)$ = 4 mA, V_{OUT} = 500 mV, and I_{OUT} = 10 mA (output on), the maximum allowable operating ambient temperature can be determined.

The power dissipation required for the output is shown below:

$$P_D(V_{OUT}) = V_{OUT} \times I_{OUT} = 500\ mV \times 10\ mA = 5\ mW$$

The power dissipation required for the IC supply is shown below:

$$P_D(V_{CC}) = V_{CC} \times I_{CC} = 5.5\ V \times 4\ mA = 22\ mW$$

Next, by inverting using equation 2:

$$\begin{aligned}
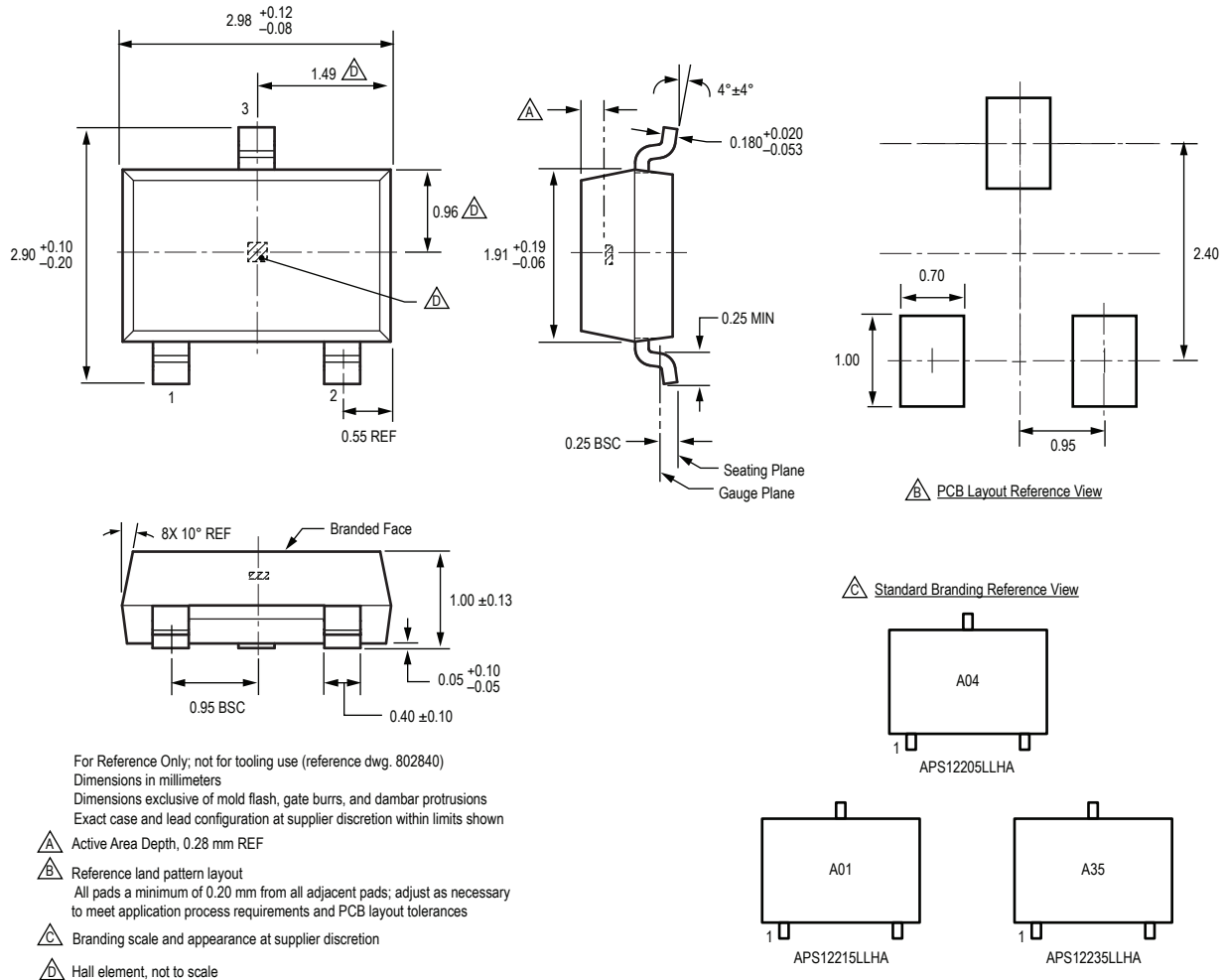\Delta T &= P_D \times R_{\theta JA} = [P_D(V_{OUT}) + P_D(V_{CC})] \times 228°C/W = \\
&\quad (5\ mW + 22\ mW) \times 228°C/W = \\
&\quad 27\ mW \times 228°C/W = 6.2°C
\end{aligned}$$

Finally, by inverting equation 3 with respect to voltage:

$$T_A(est) = T_J(max) - \Delta T = 175°C - 6.2°C = 168.8°C$$

In the above case there is only sufficient power dissipation capability to operate up to $T_A(est)$. This particular result indicates that, at $T_J(max)$, the application and device can only dissipate adequate amounts of heat at ambient temperatures $\leq T_A(est)$.

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Package LH, 3-Pin (SOT-23W)

2.98 +0.12 / −0.08

1.49 D

3

2.90 +0.10 / −0.20

0.96 D

D

1

2

0.55 REF

4°±4°

A

0.180 +0.020 / −0.053

1.91 +0.19 / −0.06

2

0.25 MIN

0.25 BSC

Seating Plane
Gauge Plane

2.40

0.70

1.00

0.95

B PCB Layout Reference View

8X 10° REF

Branded Face

1.00 ±0.13

0.05 +0.10 / −0.05

0.95 BSC

0.40 ±0.10

For Reference Only; not for tooling use (reference dwg. 802840)
Dimensions in millimeters
Dimensions exclusive of mold flash, gate burrs, and dambar protrusions
Exact case and lead configuration at supplier discretion within limits shown

A Active Area Depth, 0.28 mm REF

B Reference land pattern layout
All pads a minimum of 0.20 mm from all adjacent pads; adjust as necessary
to meet application process requirements and PCB layout tolerances

C Branding scale and appearance at supplier discretion

D Hall element, not to scale

C Standard Branding Reference View

A04

1

APS12205LLHA

A01

1

APS12215LLHA

A35

1

APS12235LLHA

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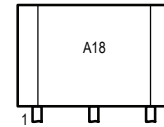1.508.853.5000; www.allegromicro.com
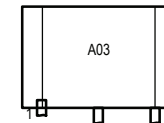
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Package UA, 3-Pin SIP

For Reference Only; not for tooling use (reference DWG-9065)
Dimensions in millimeters
Dimensions exclusive of mold flash, gate burrs, and dambar protrusions
Exact case and lead configuration at supplier discretion within limits shown

⚠️A Dambar removal protrusion (6X)

⚠️B Gate and tie bar burr area

⚠️C Active Area Depth, 0.50 mm REF

⚠️D Branding scale and appearance at supplier discretion
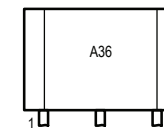
⚠️E Hall element (not to scale)

⚠️D Standard Branding Reference View

A18

APS12205LUAA

A03

APS12215LUAA

A36

APS12235LUAA

4.09 +0.08 −0.05

45°

⚠️B

⚠️E 2.04

3.02 +0.08 −0.05

1.44 ⚠️E

⚠️E

1.02 MAX

1 2 3

14.99 ±0.25

0.43 +0.05 −0.07

1.27 NOM

⚠️C

1.52 ±0.05

10°

Mold Ejector Pin Indent

Branded Face

45°

0.79 REF

0.41 +0.03 −0.06

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Revision History

Number	Date	Description
–	June 3, 2016	Initial release
1	June 20, 2016	Updated Functional Block Diagram (page 1); Updated Selection Guide (page 2) and package outline drawing brand information (pages 10-11).

For the latest version of this document, visit our website:
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